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Application of Differential Evolution to Power System Stabilizer Design

PREPARED BY:
TSHINA FA MULUMBA



This thesis is submitted to the University of Cape Town in full fulfilment of the academic requirements for the Master of Science degree in Electrical Engineering

SUPERVISOR:
PROF. K.A FOLLY



UNIVERSITY OF CAPE TOWN
IDYUNIVESITHI YASEKAPA • UNIVERSITEIT VAN KAAPSTAD

DEPARTMENT OF ELECTRICAL ENGINEERING
UNIVERSITY OF CAPE TOWN
CAPE TOWN

Date: November 2012

Declaration

I hereby declare that this is my own work. All alternative sources used have been identified and referenced. This thesis has not been submitted before for any degree at this or any other institution for any degree or examination.

Signature:

Mr. Tshina Fa Mulumba

Signed at the University of Cape Town

Date: 21st of November 2012

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Synopsis

Low frequency oscillations in the range of 0.2 to 3 Hz are inherent to power systems. They appear when there are power exchanges between large areas of interconnected power systems or when power is transferred over long distances under medium to heavy conditions. The use of fast acting high gain Automatic Voltage Regulator (AVR), although improves the transient stability, has a detrimental effect on the small-signal stability. For the last four decades, low frequency oscillations arising from the lack of sufficient damping in the system have been frequently encountered in power systems. The recent introduction of the deregulation and the unbundling of generation, transmission and distribution as well as the large amount of Distributed Generation connected to the power system have exacerbated the problem of low-frequency oscillations.

For many years, Power System Stabilizers (PSSs) have been used to add damping to electromechanical oscillations. Conventional Power System Stabilizers (CPSSs) have been widely accepted by the power utilities due to their simplicity, moreover they are the most cost effective damping control. Traditionally, CPSSs were designed using classical control techniques such as root-locus, phase compensation, eigenvalue analysis, etc. These stabilizers are mainly designed around the nominal operating condition. However, the main disadvantage with CPSSs is that they cannot guarantee the stability of power systems due to their nonlinear nature and varying operating conditions.

Over the past 30 years, there have been increasing interests in the optimization of the parameters of Power System Stabilizers (PSSs) to provide adequate performance of the PSS over varying operating conditions. Several approaches such as adaptive control, robust optimal control, etc., have been proposed. However, adaptive controllers are difficult to design and susceptible to problems like non-convergence of parameters and numerical instability. Robust controllers based on H_∞ optimal control theory have their drawbacks as well. These include the selection of appropriate weighting functions, and issues related to practical implementation due to the high dimension of the controllers.

In recent years, many Evolutionary Algorithms (EAs) such as Genetic Algorithms (GAs) have been proposed to optimally tune the parameters of the PSS. GAs are population based search methods inspired by the mechanism of evolution and natural genetic. Despite the fact that GAs are robust and have given promising results in many applications, they still have some drawbacks. Some of these drawbacks are related to the problem of genetic drift in GA which restricts the diversity in the population. As a result, GAs may converge to suboptimal solutions. In addition, GAs are computationally time consuming and require large computer storage when dealing with difficult problems that have many variables. To cope with the above mentioned drawbacks, many variants of GAs have been proposed often tailored to a particular problem. Recently, several simpler and yet effective heuristic algorithms such as Population Based Incremental Learning (PBIL) and Differential Evolution (DE), etc., have received increasing attention.

PBIL is a method that combines genetic algorithms and competitive learning for function optimization. PBIL is an extension to the Evolutionary Genetic Algorithm (EGA). PBIL's algorithm is achieved through the re-examination of the performance of the EGA in terms of competitive learning. However, PBIL drawbacks reside in the slow convergence due to the learning process involved.

DE is a stochastic optimizer based on the differential mutation technique, used as a search mechanism, and applies the greedy selection to direct the search toward the prospective solutions in the search space. DE employs a "one-to-one" survivor selection which consists of comparing each trial vector to its corresponding target vector. This process ensures that the best vector at each index is retained. Furthermore, this also guarantees that the very best-so-far solution is kept. In the last few years, DE has grown in reputation among researchers due to its simplicity, efficiency, and robustness in function optimization. Recently, DE has been used in various real-world problem solving applications, especially in engineering.

Differential Evolution has been shown to be simple yet powerful algorithm especially for optimally tuning Power System Stabilizers. However like other EAs, DE's performance is closely dependent on its intrinsic control parameters such as the mutation factor, the

crossover probability and population size. Inappropriate choice of these control parameter values may result in significant deterioration of the algorithm performance and reliability to effectively explore the search space for the global maximum or minimum. Very often, these parameters are selected by means of trial-and-error. Once selected, these parameters remain fixed throughout the search which may present limitations in DE's performance.

Recently, the self- updating of the control parameters based on the feedback from the search has been developed to overcome DE's drawback and therefore enhancing the robustness of the algorithm. The self-adaptive DE, often referred to as jDE because of the adaptive scheme used, is similar to the DE scheme. During the optimization, jDE has a fixed number of population whilst adapting the control parameters F_i and CR_i associated with each individual. Better values of the parameters lead to better individuals hence better solutions.

In this thesis, first DE is used to optimally tune the parameters of the PSS to provide adequate damping over a wide range of operating conditions. Then the self-adaptive DE is applied. The PSS parameters optimization is achieved by maximizing an eigenvalue based objective function. This consists of maximizing the lowest damping ratio of the electromechanical modes of the system.

The resulting PSS was assessed using modal analysis and validated with time domain simulations. For the small signal simulations, the system was subject to a 10 percent disturbance in the voltage reference whereas for the transient analysis, the system was subjected to 5-cycle three phase fault applied on the line.

The performances of DE-PSS were compared to those of PBIL-PSS, and to those of the CPSS. The three PSSs were designed and tested on two power system models, namely, the Single Machine Infinite bus (SMIB) and the Two-Area Multimachine system.

In the SMIB, both DE and PBIL PSSs were designed using five operating conditions. The fitness curve revealed that DE was able to maximize the lowest damping from 2.7% to 26.59%, whereas PBIL was able to maximize the lowest damping to 23.25%. The modal

analysis showed that DE-PSS performed better than PBIL-PSS for all the cases. As expected, CPSS yielded good performance for the nominal case but its performance degraded when the operating conditions changed. The time domain simulation results validated the modal analysis results. It is shown that DE-PSS settled faster than PBIL-PSS whereas CPSS was the slowest to settle down. The transient stability analysis, both DE-PSS and PBIL-PSSs had similar performances. However DE-PSS showed slightly higher overshoots and undershoots than PBIL.

In the Two-Area system, similar trends to SMIB were observed whereby DE-PSS displayed better performance than PBIL and CPSS. The frequency domain findings were validated using the time domain simulation for both small and large disturbances.

Furthermore, Self-Adaptive DE algorithm was also tested by applying it to tune the PSS implemented in SMIB system. Its performance was evaluated by means of comparison to the performance obtained with DE. Results showed that the Self-Adaptive algorithm performed better than DE both in frequency domain and time domain.

Further investigations were conducted with regards to the effect of the population size in the DE's performance when applied to Two-Area Multimachine system. Results revealed that larger population increased diversity hence explored better the search space. However, there is a trade-off between efficiency and robustness. In PSSs, despite the considerable improvement of the lowest damping ratio, larger populations tend to converge slowly to the optimum value; whereas a relatively small population converges quickly possibly to a local optimum.

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Nomenclature

A	State matrix
P_e	Electrical power
P_m	Mechanical power
T_e	Electrical torque
T_m	Mechanical torque
Δ	Small perturbation
T'_{do}, T'_{qo}	d-axis, q-axis transient open circuit time constants
T''_{do}, T''_{qo}	d-axis, q-axis sub-transient open circuit time constants
e_d, e_q	d-axis, q-axis terminal voltage component
i_d, i_q	d-axis, q-axis terminal current component
H	Inertia constant
K_D	Damping torque coefficient
L_{fd}	Field winding leakage inductance
L_l	Stator leakage inductance
L_{ad}, L_{aq}	d-axis, q-axis stator to rotor mutual inductance
L_d, L_q	d-axis, q axis synchronous inductance
L'_d, L'_q	d-axis, q axis transient inductance
L''_d, L''_q	d-axis, q axis subtransient inductance
R_{fd}	Field resistance
λ	Eigenvalue
ζ	Damping ratio
δ	Rotor angle
ω	Rotor angular velocity
ω_0	Base angular velocity
ψ_d, ψ_q	d-axis, q axis stator flux linkage
ψ_{fd}	Field flux linkage
pu	per unit
AVR	Automatic Voltage Regulator

SMIB	Single machine infinite bus
PSS	Power System Stabilizer
CPSS	Conventional Power System Stabilizer
DE-PSS	Differential Evolution Power System Stabilizer
jDE -PSS	Self-Adaptive Differential Evolution Power System Stabilizer
PBIL-PSS	Population Based Incremental Learning Power System Stabilizer
CL	Competitive Learning
PV	Probability Vector
F	Mutation Factor
CR	Crossover Probability
N_p	Population number
EA	Evolutionary Algorithm
AI	Artificial Intelligence
SVC	Static Var Compensator
STATCOM	Static Compensators
FACTS	Flexible Alternating Current Transmission Systems

Remarks: when any of the abbreviations is used in plural, then an (s) is added at the end of the listed abbreviations. Also pu and per unit have been used interchangeably within the thesis, but they are the same.

The notation for the complex numbers used in this thesis is j and not i.

Chapter 1

Introduction

This chapter introduces the concept of small signal stability problems with emphasis on low frequency oscillations. Emphasis is given to techniques applied to limit and mitigate these oscillations. Furthermore, a proposed method, which is the main focus of this research, is briefly discussed. The chapter ends with the objectives and the scope of the research.

Small signal stability is defined in [1] as “*the ability of a power system to maintain synchronism under small disturbances*”. These disturbances are mainly due to external faults and the constant change of operating conditions. As a result, low frequency oscillations in the range of 0.2 – 3 Hz are often observed in the system. For the purpose of analysis, these oscillations are often categorized into 5 groups: interplant mode, local mode, inter-area mode, control mode, and torsional mode [2]. This thesis will however focus on only the inter-area and local modes because the system models that have been considered only exhibit these oscillations.

Inter-area oscillations characterized by frequencies ranging from 0.2 to 0.8 Hz. They occur as a result of a group of generators in one area of a network oscillating against a group of generators in another area. This is typical of an interconnected network with many groups of generators [1], [3].

Local area oscillations on the other hand, have frequencies ranging from 0.8 to 2.0 Hz. They are mostly associated with a single generator oscillating against the rest of the network [1].

Over the years, these inherent oscillations have received a great deal of attention. Since the development of interconnection between synchronous generators and the introduction of deregulation of power systems, these oscillations have become apparent especially

during and after small and large disturbances, [2], [4], [5]. Several factors contribute to the rise of these oscillations, commonly known as electromechanical modes. The use of automatic controls, necessary to maintain the stability during transient faults have adverse effects on the system damping due to their negative feedback nature [4]. For instance, the rapid Automatic Voltage Regulator (AVR) and fast acting excitation system tend to reduce the damping torque component on the rotor which is necessary to damp oscillations [1], [5]. Moreover, the recent exponential increase in power demand which has led to bulk power transfer over weak transmission lines was also found to cause oscillations that limit the transfer capability of the system. If no adequate damping is provided, these oscillations might grow in magnitude with time to create system separation [1], [4] [5]. To mitigate these oscillations, various controllers have been developed and implemented over the years. Power Systems Stabilizers (PSS) have been extensively used as supplementary excitation controllers that provide additional damping to eliminate electromechanical oscillations and enhance the overall system stability. To achieve this, the PSS is often designed at a particular operating condition using conventional methods such as phase compensation, root locus, etc. However, due to the nonlinearity characteristics of power systems and the varying operating conditions, the resulting Conventional PSS (CPSS) performance deteriorates as the operating conditions change and therefore require re-tuning [6], [7].

Recently, new controllers that can provide adequate damping over a wide range of operating conditions have been investigated to compensate for the shortcoming of the CPSS. Modern control theory have been applied to design robust PSSs such as Adaptive Control [8], [9], H_∞ [10]- [11], and variable structure control [12] have been successfully applied and tested in laboratory and online. However, power utilities still remain cautious over their implementation [13].

In the recent years, increasing interests have been focused on the optimization of stabilizers parameters to provide adequate performance for a wide range of operating conditions.

Consequently, many optimization techniques and Computational Intelligence (CI) techniques have been used to find an optimal set of parameters that guarantee robust performance under varying operating conditions.

Genetic Algorithms (GAs) have received particular attention in the last few decades. GAs are heuristic population based search methods inspired by the mechanism of evolution and natural genetics. They can be used to solve optimization problems related to engineering. For instance, in [14] GA was successfully applied to design PSSs for multi-machines system. Despite the successful implementation of GAs, recent analyses have revealed some drawbacks [3]. The problem of genetic drift in GA restricts the population diversity and the searching space for solutions [3], [6], [15]. When the solutions in the population are similar, the crossover operator becomes ineffective in exploring different portion of the searching space. Consequently, the population may converge toward local optima. There are also difficulties in selecting the genetic operator variables. In the last few years, many GA variants have been developed to cope with the above limitations and increase the search space for the global maxima. Recently, Differential Evolution (DE) and Population Based Incremental Learning (PBIL) have shown their potential in global optimization problems to overcome the deficiencies of GAs in exploring wider space for the global maxima [16]- [17], especially in power system stabilizer tuning. Like GAs, DE is a population based algorithm that uses similar operators; crossover, mutation and selection. However DE search methods differ from GAs in many aspects. The main differences between the two search methods are as follows

- ❖ GAs rely on the crossover to escape from local optima and search in different zones of the search space. Whereas, DE relies on the mutation parameters as a search mechanism and selection operation to direct the search toward the prospective regions in the search space [16].

- ❖ In DE, all solutions have the same chance of being selected as parents regardless of their fitness value.

- ❖ DE encodes parameters in floating – point regardless of their type, whereas GA encoding is mainly binary although floating, gray, etc. Real-value encoding for GA has also been proposed recently.

Some of the features of DE are [16], [16], [18] :

- ❖ Low computational complexity, and efficient in memory utilization due to its one-to-one selection method.
- ❖ DE has a faster convergence, and greater freedom in designing mutation distribution than GAs.
- ❖ Because DE only has three control parameters, mutation factor, crossover probability, and population size, it makes it fairly easy to use.

Tuning these intrinsic control parameters often presents a challenge. Some guidelines have been provided in [16] to help choose appropriate mutation factor and crossover probability. Because these parameters are dependent on the nature of problem being analysed, the guidelines may not guarantee good performance of DE. Hence a trial-and-error approach is often used. In this thesis, the impact of the population size on the performance of DE is also investigated. Recently, self-adaptive DE has been used to address this drawback [19].

PBIL, on the other hand, is an extension to the Evolutionary Genetic Algorithm (EGA) achieved through the re-examination of the performance of the EGA in terms of competitive learning [20], [17]. PBIL has the following features [20]:

- ❖ It has no crossover and fitness proportional operators.
- ❖ It works with probability vector (number in range 0-1). This probability vector controls the random bitstrings generated by PBIL and is used to create other individuals through learning.

- ❖ Unlike GAs and DE, in PBIL, there is no need to store all solutions in the population. Only the current best solution and the solution being evaluated are stored. Hence, the “best” individual is used to update the probability vector so as to produce solutions similar to the current best individuals.

- ❖ PBIL only has two intrinsic parameters, namely population size and the learning rate factor.

As a result, PBIL is simpler, faster and more effective than the standard GA. However, PBIL requires a large number of generations to converge toward the optimal solution.

Following the prominence of DE in the global optimization scene, this thesis aims to present an extensive analysis of DE in PSS tuning. The results are compared to those of PBIL, an equally renowned optimizer, to assess the performance of DE. The comparison also includes the CPSS which has been designed for different system configurations.

1.1 Power System Stabilizer

For many years PSSs have been used to add damping to electromechanical oscillations. They were first introduced in the late 1960s to compensate for the AVR's adverse effect on the damping torque by means of a positive feedback loop to provide additional damping in the system [4].

PSSs essentially use the power amplification capability of the generators to generate a damping torque in phase with the speed change of the generator rotor. This is achieved by injecting a stabilizing signal into the excitation system voltage reference in such a way that a component of electrical torque proportional to the rotor speed deviation is produced [21], [22]. This stabilizing signal is, in most cases, the deviations in generator rotor speed which are fed through a compensation circuit to compensate for the phase lag between the exciter voltage reference and generator electrical torque [1], [21].

1.1.1 Power System Stabilizer Structure

The basic objective of power system stabilizer is to modulate the generator's excitation in order to produce an electrical torque at the generator proportional to the rotor speed [1], [21]. In order to achieve that, the PSS uses a simple lead-lag compensator circuit to adjust the input signal and correct the phase lag between the exciter input and the electrical torque. The PSS can use various inputs, such as the speed deviation of the generator shaft, the change in electrical power or accelerating power, or even the terminal bus frequency. However in many instances the preferred signal input to the PSS is the speed deviation.

Figure 1.1 below illustrates the block diagram of a typical PSS. The PSS structure generally consists of a washout, lead-lag networks, a gain and a limiter stages. . Each stage performs a specific function.

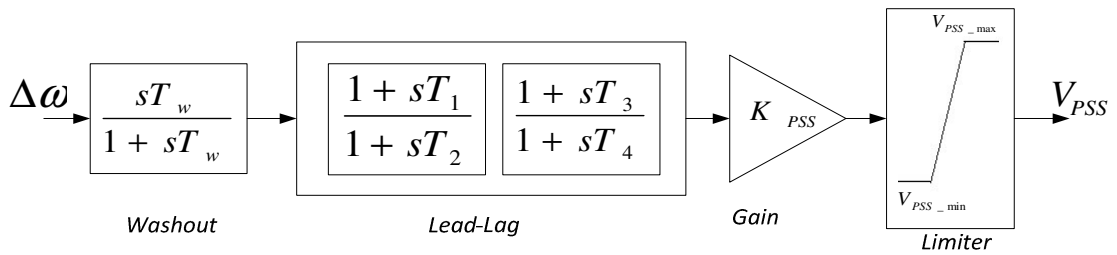


Figure 1.1: Block Diagram of typical PSS

The washout stage is a high pass filter whose purpose is to filter out undesirable signals and let through signals with frequencies in the range of 0.2 – 2 Hz. This stage prevents any change in the terminal voltage. The value associated with the time constant T_w is not critical and may be in the range of 1 to 20 seconds. However it must be long enough to pass stabilizing signals at the frequencies of interest relatively unchanged, but not so long that it leads to undesirable generator voltage excursions. For local mode oscillations, a washout of 1 to 2 s is satisfactory. Whereas for the inter-area oscillations a washout time constant of 10 s or higher may be required in order to reduce phase lead at low frequencies [23], [24].

The compensation stage consists of a combination of lead and lag circuits that produce an appropriate phase lead characteristic to compensate for the phase lag between the exciter input and the generator's electrical torque. However the phase lag changes with the operating condition. Therefore, a compromise must be made when determining the phase lead. Hence a characteristic that is satisfactory for both the range of frequencies between 0.1 to 2.0 Hz and for different system conditions must be selected. This may result in less than optimum damping at any one frequency. Generally, slight under-compensation is preferable to over-compensation so that both damping and synchronizing torque components are increased [1], [4], [23],

The gain stage determines the amount of damping introduced by the stabilizer. Hence, increasing the gain can move unstable oscillatory modes into the left – hand complex s-plane. Ideally, the gain should be set to a value corresponding to a maximum damping. However, in practice the gain K_{PSS} is set to a value satisfactory to damp the critical mode without compromising the stability of other modes [4], [23], [24].

The limiter prevents conflicts with AVR actions during transient fault. The positive and negative limits should be set around the AVR set point to avoid any counteraction. The positive limit of the PSS contributes to improve the transient stability in the first swing during a fault whereas the negative limit acts during the back swing of the rotor.

Designing a power system stabilizer is a complex task, particularly in a multimachine system environment where several machines are involved. Parameters of the stabilizer need to be appropriately tuned so that the damping of the electromechanical modes is increased without adversely affecting the other oscillatory modes. Several design methods have been investigated however the most common one is the Conventional Power System Stabilizer (CPSS).

1.1.2 Conventional Power System Stabilizer design method

For many years conventional control methods have been applied to design PSSs. These approaches consist of first linearizing the system at the nominal operating condition to be able to extract the dynamic characteristics of the power system and its frequency response. Once the phase lag is identified, the phase lead can be obtained by tuning the

time constants of the lead-lag circuit. Ideally a phase lead, equal and opposite to the phase lag, is required to produce an electrical torque with a component proportional to the speed. However in practice this cannot be achieved but can be closely matched over the frequency range [4].

The gain on the other hand is obtained by applying the root locus method. The gain must be carefully selected to stabilize the electromechanical mode without adversely affecting the other modes such as the exciter mode [4], [24].

It is important to choose an appropriate value for the washout T_w . It would be adequate to choose the time constant between 1 and 2 seconds if the damping of the local mode is the only concern. However a T_w of 10 seconds or higher when inter – area is considered [1], [24], [25].

Generally, determining the stabilizer's parameters in systems with both local and inter – area modes has a more complex approach. For the most case this situation is encountered in a multimachine system. Therefore PSSs must be tuned one at a time through off-line analysis, and tuned further during commissioning. The validity of the model used in the off-line studies should be checked on commissioning. Setting power system stabilizers to typical values is particularly dangerous for systems in which inter – area modes are of concern. It is very easy for the stabilizer to have a destabilizing effect at low frequencies that cannot be observed during on-line commissioning test [23], [26].

The performance of the CPSS often deteriorates over time due to nonlinearity and changes of operating conditions. Over the years, several approaches of controllers design have been investigated and implemented to overcome the shortcomings of the CPSSs. Some of these methods are reviewed in the next section.

1.2 Modern Control Design Methods

In the last few decades, new stabilizers able to provide adequate damping across a wide range of system operating conditions have been developed and successfully tested. These

designs are based on modern control theories which cater for the nonlinearity characteristic of power systems.

1.2.1 Adaptive Control

Adaptive control can be described as the changing of controller parameters based on the changes in system operating conditions [8]. The idea is to constantly update the controller parameters according to recent measurement [3]. Power systems are inherently nonlinear with varying operating conditions, hence adaptive control technique is well suited to track the operating conditions and changes in the system. The resulting adaptive stabilizer uses an identification algorithm that tracks the actual system operating condition, which then adjusts its parameters on-line according to the environment in which it works. This method can provide good damping over a wide range of operating condition [9], [27]-[28].

Despite the good performance of the stabilizer, adaptive controllers are difficult to design and susceptible to problems like non-convergence of parameters and numerical instability. The response time of the controller is the key factor to a good closed-loop performance. The adaptive power system stabilizer (APSS) employs complicated algorithms for parameter identification and optimization which require significant amount of computing time. The higher the order of the discrete model of the controlled system used in identification, the more computing time is needed. To develop a quick response PSS, it is necessary to investigate alternative techniques such as neural network and Fuzzy logic based adaptive [22], [27], [29]

1.2.2 H_{∞} Controller

H_{∞} is a control technique that addresses the issue of the worst – case controller design for systems subject to unknown disturbances, including problems of disturbance attenuation, model matching, and tracking. The objective is to minimize the maximum norm of an input-output operator, where the maximum is often taken over the unknowns such as disturbances [30]. Following the nonlinearity characteristic of power systems and unpredictable change of operating conditions, H_{∞} theory has been applied to overcome the CPSS shortcomings. The method provides a theoretical mechanism to deal with

uncertainty in a system control design problem. The resulting controller minimizes the effect of external disturbances on system output in terms of a H_∞ norm which can easily put the various types of disturbances into a single framework by using a frequency weighting function to emphasise the interesting noise band. The mismatch between the physical system and its mathematical description has been taken into account in the control design process to cope with the stability problem in the presence of system uncertainty [10]. H_∞ has been successfully applied the designing PSSs [10], [31], [32], [33]. However these controllers have drawbacks. Which includes the selection of appropriate weighting functions, and issues related to practical implementation due to the high dimension of the controllers [32]. In addition, the mixed sensitivity approach produces closed loop poles whose damping is directly dependent on the open loop system. This effect is caused by the pole-zero cancellation phenomena associated with such an approach. This problem was solved in [34] using Bilinear approach. However the H_∞ based PSS tends to have the same order as the plant, therefore resulting in highly complex stabilizers [3].

1.2.3 State-Feedback Optimal PSS

This method is based on eigenvalue shifting technique used to determine the weighing matrix in the performance index. The dominant eigenvalue is shifted to the left side of the s-plane until satisfactory shift is achieved or the controller's practical limit is reached. Computational programs are used to shift the poles in the complex s-plane. It allows for the shaping of the dynamic response of the system. This method works properly, but tends to have some problems, especially as the state matrix of the system grows. It makes the process to be complex and computationally intensive [3], [35], [36].

Several compensations devices, such as Static Var Compensators (SVC), Thyristor Controlled Series Capacitors (TCSC), and other Flexible AC Transmission System (FACTS) have also been implemented to damp electromechanical modes [4], [37], [38]. However power utilities still prefer the conventional lead-lag PSS because they are the most cost-effective oscillation damping controllers. This has led researchers to focus on the optimization of the PSS parameters to effectively damp electromechanical oscillations over a wide range of operating conditions.

1.3 Optimization Based Power System Stabilizers

Optimization is defined in [18] as the attempt to maximize a system's desirable properties whilst simultaneously minimizing its undesirable characteristic. In the recent years, there has been increasing interests in applying optimization techniques to design power system stabilizers. The approach consists of converting the problem of selecting PSS parameters into a simple optimization problem. Thereafter finds the optimal set of PSS' parameters that will guarantee adequate damping across the entire range of operating conditions. Over the years, several optimization algorithms have been investigated and few are reviewed below.

1.3.1 Gradient based optimization

Gradient methods are classic optimization techniques that are based on a single point derivative and iterative procedures. They use quadratic approximation of the function of interest around one initial point. Then, the value of the point is then adjusted to a new approximation or increased by a small step of the gradient until optimum value is reached. These methods often require the existence of first derivative of the objective function or even higher order in order to work properly [39]. Several gradient based methods have been used for optimization such as Newton-Raphson, Quasi-Newton and Gradient descent. However these methods are complex and computationally demanding. They require excessive restrictions on parameters, objective functions, and constraint functions. Convexity, continuity, and differentiability of objective functions are necessary for these algorithms. The key disadvantage is their convergence to local minima. Most real-world applications, such as PSS tuning, are represented as nonlinear and discontinued functions. Therefore, the gradient techniques are not appropriate for the PSS tuning. Hence, people have increasingly turned their attention to stochastic optimization algorithms, especially evolution algorithms.

1.3.2 Evolution Algorithms

Evolution Algorithms (EAs) are population-based optimizer inspired by the mechanism of evolution and natural selection. They attack the starting point problem by sampling the objective function at multiple random initial points and explore the search space by iteratively generating new point that are perturbation of the existing ones [16]. This

approach is convenient in locating the global maximum/minimum instead of local. EAs are simple, robust, efficient, and versatile algorithms that can be applied to any types of problems irrespective the function type, such as nonlinearity, or discontinuity or even the complex.

In the last few years, increasing number of researches have proposed EAs to optimally tune the parameters of the PSS to guarantee a robust performance over a wide range of system conditions. Genetic Algorithms (GAs) have received particular attention in the last few decades. GAs are heuristic population based search methods inspired by the mechanism of evolution and natural genetic. They can be used to solve optimization problems related to Engineering. For instance, in [14] GA was successfully applied to design PSSs for multi-machines system. Despite GAs' performance and promising results in numerous applications, recent analyses have revealed some drawbacks [40]. The problem of genetic drift in GA restricts the population diversity and the searching space for solutions. As a result, GAs may converge to suboptimal solutions [6], [7], [40]. In addition, GA is computationally time consuming and require large computer storage when dealing with difficult problems such as tuning PSSs in a multi-machine environment [41], [42].

To cope with the above drawbacks, many variants of GAs have been proposed often tailored to a particular problem. Recently, several simpler and yet effective heuristic algorithms have received increasing attention. They have shown their potential in global optimization problems to overcome the deficiencies of GAs in exploring wider space for the global maxima [16], [43], [44] especially in power system stabilizer tuning. Some of these algorithms have been reviewed in the subsequent paragraphs.

Breeder Genetic Algorithm (BGA): was first introduced by Mühlenbein [44]. BGA uses the concept of artificial breeding, whereby the top $T\%$ (say 15%) best solutions are saved as “breeding pool” and the rest discarded. This is a similar concept used in animal breeding. This algorithm use real-valued encoding. In [3], both BGA and PBIL were applied to the design of PSS. They all displayed similar performance with each other and

outperformed GA. However BGA major disadvantage is its tendency quickly converge to a solution that may be local due to the quick loss of diversity.

Population Based Incremental Learning (PBIL): is a method that combines genetic algorithms and competitive learning for function optimization. PBIL is an extension to the Evolutionary Genetic Algorithm (EGA) algorithm achieved through the re-examination of the performance of the EGA in terms of competitive learning [20]. In [3], [7] and [41], PBIL was used to tune the PSS' parameters. The results showed considerable enhancement in the PSS performance as opposed to those of GA. However, PBIL drawbacks reside in the slow convergence due to the learning process involved. This drawback can be solved by optimally tuning the learning rate (LR) parameter by trial-and-error. Despite the tuning the LR, PBIL still requires a large number of generation to obtain a solution.

Differential Evolution: is a stochastic optimizer based on the differential mutation technique, used as a search mechanism, and applies the greedy selection to direct the search toward the prospective solution in the search space [16], [45], [46]. DE employs a "one-to-one survivor selection" which consists on comparing each trial vector to its corresponding target vector. This process ensures that both the best vector at each index is retained. Furthermore this also guarantees that the very best-so-far solution is kept. In DE, all candidates have the same probability of being accepted. This algorithm is very simple yet very powerful and robust optimizer.

DE has been the subject of intensive performance evaluation since its inception. Many comparisons have been carried out with other optimization algorithms on benchmark functions and several other applications. Most often DE outperforms its counterparts in efficiency and robustness [45]. DE has thus earned its reputation and recognition among researchers for its simplicity, efficiency, and robustness for function optimization. Recently, DE has been implicitly used in a variety of problems, especially in engineering [45]. As a result, in this thesis DE' ability will be examined in optimally tune the PSS in different system configurations. A systematic approach is used to test DE's performance. This approach consists of investigating the effect of the intrinsic control parameters on

the DE ability to tune the PSS. To address the drawbacks of DE, this investigation further extends to the application of a self-adaptive scheme to PSS tuning whereby the mutation and crossover parameters are changing during the optimization.

Several researches have applied DE in tuning the PSS. In [47], DE was successfully applied in designing PSS. The resulting PSS was then compared to the CPSS through a series of tests whereby DE based PSS outperformed the CPSS. In [46], DE is compared to GA when used to simultaneously tune PSSs. The results have proven that DE outperformed GA.

Several other PSS design methods Particle Swarm Optimization (PSO) [48], and Simulated Annealing (SA) [49] that have been successfully applied over the years, but have not been discussed in this thesis.

1.4 Objectives of the Thesis

The objective of this thesis is to design a power system stabilizer based on Differential Evolution (DE). The purpose of the applied optimization algorithm is to solve the problem of conventional power system stabilizer design in the context of nonlinear plant and varying operating conditions. With the versatility and effectiveness of DE in the global optimization arena, it is hoped that this work will make a contribution to the development and application of DE.

In order to develop a robust DE based PSS, the following topics are addressed in this thesis:

- ❖ Find the optimal set of PSS' parameters that ensures robust performance and stability of the system over a wide range of operating conditions.
- ❖ Assess the performance of DE's optimization ability in comparison to that of Population-Based Incremental Learning (PBIL), a prominent optimization technique.
- ❖ Simultaneously tune the power system stabilizers using DE and PBIL for different system configurations such as SMIB and the Two-Area Multimachine.

- ❖ Evaluate the resulting PSSs (DE & PBIL) by comparing their performances including those of the CPSS by means of modal analysis which are validated using time domain simulation.
- ❖ Investigate the effect of the intrinsic control parameters of DE in the PSS tuning. Address the resulting drawbacks and propose alternatives.
- ❖ Draw conclusions and recommendations on the DE's performance.

1.5 Scope of the research

This research investigates the application of Differential Evolution (DE) to design Power System Stabilizers over a wide range of operating conditions. This approach consists of determining a set of PSS' parameters that ensures optimal performance, stability and robustness across all operating conditions. The PSS tuning is achieved by maximizing the system minimum damping ratio. This process is also known as the objective function which is presented in Chapter 5.

The scope of the research also includes the design of PSSs using conventional methods and Population Based Incremental Learning, a prominent optimization algorithm. A comparison analysis is carried out to assess the performances DE based PSS. The resulting PSSs are applied to different power system models such as Single Machine Infinite Bus (SMIB) and Two-area multimachine. The investigations in this thesis are limited to the analysis of the electromechanical modes which were conducted by means of modal analysis. The results of the modal analysis were validated with time domain simulations using both small and large disturbances. Further investigation is carried out to assess the effect of DE's control parameters on the performance of the DE based PSS. This has led to the investigation of Self-Adaptive DE which was implemented, tested and compared with the DE on an SMIB system. Conclusions and recommendations based on the results obtained are presented.

1.6 Research Contribution

This research addresses the performance of DE in comparison with PBIL and CPSS when applied Power System Stabilizers tuning. The performance is determined by the

convergence rate, frequency and time domains analysis. The major contribution are summarized as follows

First, experimental analysis has been conducted to investigate DE's convergence behaviour for a PSS by understanding the effect of mutation factor and crossover probability on the dynamic of the algorithm. This underlines the necessity to use self-adaptive control parameters to tune PSSs.

Second, a self-adaptive algorithm is implemented to tune the PSS. This process prevents the users' manual tuning of the control parameters suitable to the characteristics of the optimization problems. The resulting controller is tested and compared with the classic DE from a performance point of view.

Third, the effects of the population size on the algorithm performance to tune the PSS have been investigated.

1.7 Thesis Outline

The thesis report is organized as follows:

- Chapter 1** Gives a background theory on small signal stability concept with emphasis on the low frequency oscillation, its causes and mitigation techniques. A review of the relevant work in the area of PSS tuning is presented which lays down the motivation and the objectives of this work. Differential Evolution is also briefly discussed.
- Chapter 2** Reviews the mathematical concept of linearization applied to small signal stability. The state space representation is also discussed.
- Chapter 3** Presents DE and its optimization algorithm. Emphasis is put on the differential mutation which is a process used to guide DE's search toward the global optimum. The crossover and selection criteria are also discussed.
- Chapter 4** Presents an overview of Population Based Incremental Learning (PBIL), which is also an Evolutionary Algorithm. PBIL is based on a concept of learning that update a probability vector (PV) to direct the search to the optimum solution.

- Chapter 5** Discusses the design procedure of a Power System Stabilizer (PSS). The following consists of using the conventional method and the application of the Differential Evolution (DE) and Population-Based Incremental Learning (PBIL) to find the optimum set of parameters that guarantees the robustness over a wide range of operating conditions. In addition, test cases for various system configurations such as SMIB and multimachine are also discussed in this section.
- Chapter 6** Presents the optimized power System Stabilizers parameters. It also discusses the results obtained for the SMIB. The eigenvalues analysis is conducted as well as the time domain simulations are discussion thereof.
- Chapter 7** Presents the application of DE and PBIL to simultaneously tune PSSs in a Two-Area Multimachine system. Results are discussed by conducting a modal analysis and time domain simulations which compares DE-PSS, PBIL-PSS to CPSS.
- Chapter 8** Tests the effects of DE control parameters on the optimization. Preliminary work on Self-Adaptive Differential Evolution applied to power system tuning, and the effect of population size in the optimization are discussed.
- Chapter 9** Concludes the thesis and give recommendations as well as giving suggestions for future work.

Chapter 2

Small Signal Stability Analysis

2.1 Introduction

As defined in Chapter 1, small signal stability is the ability of a power system to maintain stability when subject to small disturbances [1]. The disturbances are considered small only if linearization of the system is possible. Hence, this chapter reviews linear techniques used to analyse small signal oscillations and extract information about the system dynamic characteristic. Techniques such as modal analysis, eigenvectors, eigenvalues' sensitivity and participation factors are elaborated in subsequent sections.

2.2 State-Space Representation

The State-Space representation is often used to describe the behaviour of a dynamic system. Hence, to model this particular behaviour, a set of n first order nonlinear differential equations are used. The notations are borrowed from [1] and are as follows

$$\dot{x}_i = f_i(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_r; t) \quad i = 1, 2, 3, 4 \dots n \quad (2.1)$$

Where n is the order of the system and r is the number of inputs. That can be written in vector representation form:

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}; \mathbf{u}; t) \quad (2.2)$$

Where:

$$\mathbf{x} = \begin{bmatrix} x_1 \\ \vdots \\ x_n \end{bmatrix} \quad \mathbf{u} = \begin{bmatrix} u_1 \\ \vdots \\ u_n \end{bmatrix} \quad \mathbf{f} = \begin{bmatrix} f_1 \\ \vdots \\ f_n \end{bmatrix}$$

\mathbf{x} : the state vector containing the state variables

$\dot{\mathbf{x}}$: contains the derivative of the state variables

\mathbf{f} : the vector of the non-linear functions

\mathbf{u} : the vector of the inputs

t : the time

n : the order of the system

r : the number of inputs

If the derivatives of the states variables are not explicit functions of time, the system is referred to as autonomous systems of the form:

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}) \quad (2.3)$$

The equation relating the outputs to the inputs and state variables can be written as

$$\mathbf{y} = \mathbf{g}(\mathbf{x}, \mathbf{u}) \quad (2.4)$$

Where

$$\mathbf{y} = \begin{bmatrix} y_1 \\ \vdots \\ y_n \end{bmatrix} \quad \mathbf{g} = \begin{bmatrix} g_1 \\ \vdots \\ g_n \end{bmatrix}$$

\mathbf{y} : is the vector of outputs

\mathbf{g} : is the vector of non-linear equations relating the state and inputs variables to the outputs variables.

2.3 Linearization

The linearization process consists of initializing the state vector \mathbf{x}_0 and \mathbf{u}_0 the input vector around the equilibrium point. Therefore, the following equation should be true.

$$\dot{\mathbf{x}}_0 = \mathbf{f}(\mathbf{x}_0, \mathbf{u}_0) = 0 \quad (2.5)$$

Following a small perturbation, $\Delta\mathbf{x}$ and $\Delta\mathbf{u}$, in the system state and input variables, we get

$$\dot{\mathbf{x}} = \dot{\mathbf{x}}_0 + \Delta\dot{\mathbf{x}} \quad \mathbf{u} = \mathbf{u}_0 + \Delta\mathbf{u} \quad (2.6)$$

The above must satisfy equation 6.3 as follows

$$\dot{\mathbf{x}} = \dot{\mathbf{x}}_0 + \Delta\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}_0 + \Delta\mathbf{x}, \mathbf{u}_0 + \Delta\mathbf{u}) \quad (2.7)$$

If the perturbations in the system are assumed to be small, the system equations $\mathbf{f}(\mathbf{x}, \mathbf{u})$ can be expressed in Taylor's series expansion terms as in [1]. With few simplification, by neglecting the terms associated with the second and higher order of powers of $\Delta\mathbf{x}$ and $\Delta\mathbf{u}$, equations (2.3) and (2.4) are as follows.

$$\Delta\dot{x}_i = \frac{\partial f_i}{\partial x_1} \Delta x_1 + \dots + \frac{\partial f_i}{\partial x_n} \Delta x_n + \frac{\partial f_i}{\partial u_1} \Delta u_1 + \dots + \frac{\partial f_i}{\partial u_r} \Delta u_r \quad (2.8)$$

Where $i = 1, 2, 3 \dots n$ and

$$\Delta y_i = \frac{\partial g_j}{\partial x_1} \Delta x_1 + \dots + \frac{\partial g_j}{\partial x_n} \Delta x_n + \frac{\partial g_j}{\partial u_1} \Delta u_1 + \dots + \frac{\partial g_j}{\partial u_r} \Delta u_r \quad (2.9)$$

Where $j = 1, 2, 3 \dots m$

For the general state space system, the linearization of (2.3) and (2.4) around the operating point \mathbf{x}_0 and \mathbf{u}_0 are [1]:

$$\Delta\dot{\mathbf{x}} = \mathbf{A}\Delta\mathbf{x} + \mathbf{B}\Delta\mathbf{u} \quad (2.10)$$

$$\Delta\mathbf{y} = \mathbf{C}\Delta\mathbf{x} + \mathbf{D}\Delta\mathbf{u} \quad (2.11)$$

where:

$$\mathbf{A} = \begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \dots & \frac{\partial f_1}{\partial x_n} \\ \vdots & \ddots & \vdots \\ \frac{\partial f_n}{\partial x_1} & \dots & \frac{\partial f_n}{\partial x_n} \end{bmatrix} \quad \mathbf{B} = \begin{bmatrix} \frac{\partial f_1}{\partial u_1} & \dots & \frac{\partial f_1}{\partial u_r} \\ \vdots & \ddots & \vdots \\ \frac{\partial f_n}{\partial u_1} & \dots & \frac{\partial f_n}{\partial u_r} \end{bmatrix} \quad (2.12)$$

$$\mathbf{C} = \begin{bmatrix} \frac{\partial g_1}{\partial x_1} & \dots & \frac{\partial g_1}{\partial x_n} \\ \vdots & \ddots & \vdots \\ \frac{\partial g_m}{\partial x_1} & \dots & \frac{\partial g_m}{\partial x_n} \end{bmatrix} \quad \mathbf{D} = \begin{bmatrix} \frac{\partial g_1}{\partial u_1} & \dots & \frac{\partial g_1}{\partial u_r} \\ \vdots & \ddots & \vdots \\ \frac{\partial g_m}{\partial u_1} & \dots & \frac{\partial g_m}{\partial u_r} \end{bmatrix}$$

$\Delta \mathbf{x}$ is the linearized state vector of dimension n

$\Delta \mathbf{y}$ is the linearized output vector of dimension m

$\Delta \mathbf{u}$ is the linearized input vector of dimension r

\mathbf{A} is the state matrix of size $(n \times n)$

\mathbf{B} is the input matrix, size $(n \times r)$

\mathbf{C} is the output matrix, size $(m \times n)$

\mathbf{D} is the feed forward matrix, size $(m \times r)$

After linearization, the system stability can be analysed using the modal analysis as follows.

2.4 Modal Analysis

Once the state space has been established, the stability of the system can be extracted and analysed by means of eigenvalues, eigenvector properties and participation factor.

2.4.1 Eigenvalues

The eigenvalues are the non-trivial solutions of a the matrix \mathbf{A} , obtained by solving the following equation

$$\mathbf{A}\Phi = \lambda\Phi; \quad (2.13)$$

Where Φ is $n \times 1$ vector and \mathbf{A} is $n \times n$ state matrix.

By rearranging equation (2.11), λ value can be obtained by solving the following equation

$$\det(\mathbf{A} - \lambda\mathbf{I}) = 0 \quad (2.14)$$

Where \mathbf{I} is the identity matrix

The n solutions of the above equation are the eigenvalues $(\lambda_1, \lambda_2, \lambda_3, \dots, \lambda_n)$ of the $n \times n$ matrix \mathbf{A} . These eigenvalues can be real or complex of the form $\lambda = \sigma \pm j\omega$.

For a real \mathbf{A} matrix, the complex eigenvalues occurs in conjugate pairs.

The stability of the system at an operating point (x_0, u_0) can be determined by analysing the eigenvalues. Therefore, a system is said to be stable at a particular operating point if all the eigenvalues are on the left-hand side of imaginary axis of the complex plane, and unstable otherwise. A real eigenvalue corresponds to a non-oscillatory mode, whilst a complex eigenvalue corresponds to an oscillatory mode. If the real eigenvalue is negative, the mode decays over time; and if positive, the mode is said to have an aperiodic instability. Whilst, for the conjugate complex eigenvalues $(\sigma \pm j\omega)$, the real component σ determines whether the mode is oscillatory stable or unstable by either being positive or negative, whereas the imaginary component, ω , gives the oscillatory frequency in rad/s. The frequency of oscillation in Hertz is calculated as follows

$$f = \frac{\omega}{2\pi} \quad (2.15)$$

where the damping ratio ζ , which determines the rate of decay of the amplitude of the oscillations and is given by the following equation:

$$\zeta = \frac{-\sigma}{\sqrt{\sigma^2 + \omega^2}} \quad (2.16)$$

For power systems, a damping ratio of 5% and above is considered adequate, but a damping ratio of 20% and above is often preferred, especially for electromechanical oscillations [4], [3].

2.4.2 Eigenvectors

Despite the fact that the accurate evaluation of the frequency and damping of oscillations provide useful information about the stability, further information regarding the nature of oscillations can be extracted by using eigenvectors. The contribution of each mode to a particular state can be determined.

Hence, there is one and only one eigenvector associated with each eigenvalue. Such that for the i^{th} eigenvalue, the eigenvector u_i satisfies the equation

$$A\phi_i = \lambda_i \phi_i; \quad i = 1, 2, 3 \dots n; \quad (2.17)$$

The eigenvector has the form

$$\phi_i = \begin{bmatrix} \phi_{1i} \\ \vdots \\ \phi_{ni} \end{bmatrix}$$

Where ϕ_i is a column vector also known as the right eigenvector of dimension equal to the number of state variables. Eigenvectors are not unique and each remains a valid eigenvector when scaled by a constant. The right eigenvector describes how each mode of oscillation is distributed among the system states [4].

Whereas, left eigenvectors are row vectors ψ_i that satisfy;

$$\psi_i A = \lambda_i \psi_i; \quad i = 1, 2, 3, \dots, n \quad (2.18)$$

The left eigenvector describes the contribution of the activity of a state variable in a mode.

The left and right eigenvectors are orthogonal, therefore;

$$\psi_i \phi_i = 1 \quad (2.19)$$

To express the eigenproperties of the matrix A , modal matrices are introduced as follows

$$\begin{aligned} \Phi &= [\phi_1 \quad \dots \quad \phi_n] \\ \Psi &= [\psi_1^T \quad \dots \quad \psi_n^T]^T \\ \Lambda &= \text{diagonal matrix of eigenvalues as elements } (\lambda_1, \lambda_2, \lambda_3, \dots, \lambda_n) \end{aligned}$$

Such that;

$$\begin{aligned} A\Phi &= \Phi\Lambda & (2.20) \\ \Psi\Phi &= 1 \quad \text{hence } \Psi = \Phi^{-1} \end{aligned}$$

Equation 2.17 can be re-written as follows

$$\begin{aligned}\Phi^{-1}A\Phi &= \Lambda \\ \Psi A\Phi &= \Lambda\end{aligned}\quad (2.21)$$

2.4.3 Eigenvalue sensitivity

Eigenvalue sensitivity is a property useful for control analysis which helps determine the sensitivity to changes in the elements of the state matrix. To achieve that, equation 2.17 which defines the i^{th} eigenvalue and i^{th} eigenvectors are differentiated with respect to the element a_{rs} of the state matrix in the r^{th} row and s^{th} column. The following is obtained

$$\frac{\partial A}{\partial a_{rs}}\Phi_i + A\frac{\partial \Phi_i}{\partial a_{rs}} = \frac{\partial \lambda_i}{\partial a_{rs}}\Phi_i + \lambda_i\frac{\partial \Phi_i}{\partial a_{rs}}\quad (2.22)$$

Multiplying both sides with the i^{th} left eigenvector ψ_i

$$\psi_i\frac{\partial A}{\partial a_{rs}}\Phi_i + \psi_i(A - \lambda_i I)\frac{\partial \Phi_i}{\partial a_{rs}} = \psi_i\frac{\partial \lambda_i}{\partial a_{rs}}\Phi_i\quad (2.23)$$

Using the definition of left eigenvector above, $\psi_i\Phi_i = 1$ and $\psi_i(A - \lambda_i I) = 0$, the equation will simplify to

$$\frac{(\partial \lambda_i)}{\partial a_{rs}} = \frac{\psi_i\frac{\partial A}{\partial a_{rs}}\Phi_i}{\psi_i\Phi_i} = \psi_i\frac{\partial A}{\partial a_{rs}}\Phi_i\quad (2.24)$$

Thus the sensitivity of the eigenvalue λ_i to the element a_{rs} of the state matrix is equal to the product of the left eigenvector element ψ_i and the right eigenvector element ϕ_i .

2.4.4 Participation factors

Despite the fact that eigenvectors are a good indication of the relative activity of the states within a mode, it is not a good indicator of the importance of states to the mode from a control aspect. To overcome this shortcoming, a matrix containing the measures of

the association between the state variables and the modes is formed by combining the right and left eigenvectors. This matrix is known as participation matrix \mathbf{P} [1], [4].

$$\mathbf{P} = [\mathbf{p}_1 \quad \dots \quad \mathbf{p}_n] \quad \text{with} \quad \mathbf{p}_i = \begin{bmatrix} p_{1i} \\ \vdots \\ p_{ni} \end{bmatrix} = \begin{bmatrix} \phi_{1i} \psi_{1i} \\ \vdots \\ \phi_{ni} \psi_{ni} \end{bmatrix} \quad (2.25)$$

Let a certain element p_{ki} , also known as the participation factor be defined as the measure of the relative participation of the k_{th} state variable in the i_{th} mode and given as follows

$$p_{ki} = \frac{\partial \lambda_i}{\partial a_{kk}} \quad (2.22)$$

In power system, participation factors give good indications for power system stabilizers placement [4]. As previously mentioned, PSS is a device use to provide supplementary signal to add damping at the generator shaft. Hence, the participation factor is use to identify the states that correspond have the highest participation in the mode (in general rotor speed and rotor angle). If the corresponding rotor angle and/or rotor speed participation factor of a generator in a mode is zero, then that particular generator state does not contribute to the damping of the mode. However if the participation factor is real positive, adding damping at the generator will increase the damping of the mode whereas if negative, it will have adverse effects [4].

2.5 Summary

This Chapter reviews the mathematical modelling of the small signal stability. The linearization and the modal analysis have been discussed. The relationship between the eigenvalues and the system states were also established.

Chapter 3

Differential Evolution Algorithm

This chapter provides a background theory to the Differential Evolution technique. Furthermore, an in-depth analysis on its searching mechanism, used to find the optimum value of a function, is presented.

3.1 Overview

DE was originally developed by Price and Storn in 1995 in an effort to overcome Genetic Annealing shortcomings in solving the Chebyshev polynomial fitting problem [16], [18]. This algorithm is also a stochastic technique classified as an Evolutionary Algorithm (EA). DE is more accurately defined as a parallel direct search method that uses a population of points to search for a global minimum or maximum of a function over a wide search space [16], [50]. Similar to most EAs, DE simulates the Darwinian evolution theory to direct its search toward prospective areas. However, unlike Genetic Annealing and traditional Genetic Algorithm (GA) which uses “bit-string” encoding, DE encodes all parameters as floating-point regardless of their type. The floating-point representation offers efficient memory utilization and its one-to-one selection strategy lowers computational complexity which subsequently increases the speed of convergence. Whereas, GAs are slower at converging to an optimum value and difficult to effectively choose the control parameters [16].

DE is designed to efficiently solve non-differentiable and nonlinear functions and yet retains its simplicity and good convergence to a global optimum [50]. The algorithm starts by sampling the search space at various initial points randomly chosen, and then generates new points that are perturbations (or mutations) of existing points. The perturbation is achieved by adding the scaled difference between two randomly selected vectors to a third. The resultant mutation vector is crossed over with the corresponding parent to generate a trial or offspring vector [16], [50], [51]. Finally, the offspring competes against its parent in a one-to-one selection process based on their fitness value.

The one with the better fitness value survives and enters the next generation. The procedure is repeated for each point (or vector) in the search space to form the new generation in the evolutionary cycle. The search stops when either the solution converges to the true optimum or following a termination criterion such as maximum generation number is reached [16].

3.2 Population Structure

In DE, the population, P_x , is composed of N_p candidates solution or points denoted as $X_{i,g}$. Each candidate is D dimensional real – valued vector and ‘ D ’ is the number of parameters to be optimized [16].

$$P_{x,g} = (X_{i,g}), \quad i = 0,1, \dots, N_p - 1, \quad g = 0,1, \dots, g_{max} \quad (3.1)$$

$$X_{i,g} = (x_{j,i,g}), \quad j = 0,1, \dots, D - 1$$

The index, $g = 0,1, \dots, g_{max}$ indicates the generation to which the vector belongs. Whereas $i = 0,1, \dots, N_p - 1$, indicates the individual within the population. The parameters to be optimised are indexed by j .

Figure 3.1 best illustrates the population and the vector candidates.

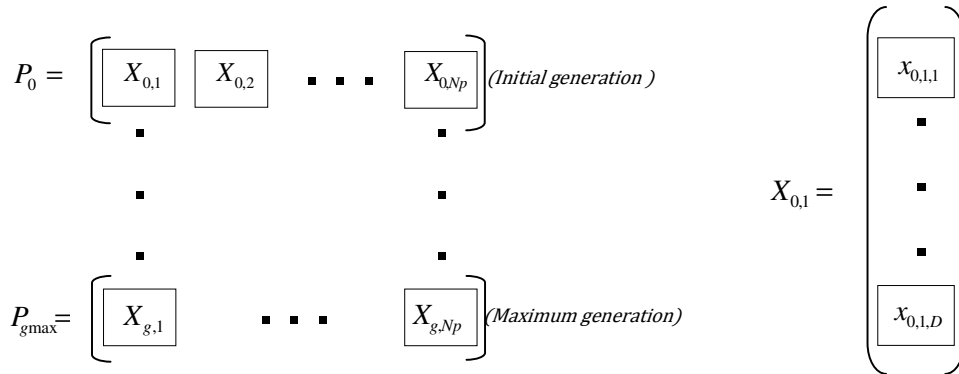


Figure 3.1: Population and Candidate structure

Figure 3.1 shows the population within the generation constituted of N_p individuals. Each individual is encoded with D parameters to be optimized.

3.3 Initialization

DE starts the optimization process by generating an initial population P_o of N_p points or vectors $X_{i,g}$ encoded with D parameters $x_{j,i,D}$. Every vector's parameter is initialized within the specified upper and lower bound of each parameter [16], [18], [50].

$$x_j^L < x_{j,i,g} < x_j^U \quad (3.2)$$

For example, the initial value, at generation ($g=0$), of the j -th parameter of the i -th vector is as follows

$$x_{j,i,0} = rand_j(0,1) \cdot (x_j^U - x_j^L) + x_j^L \quad (3.3)$$

$Rand_j$ is a random number generated within the $0 \leq rand_j < 1$. The index j refers to the vector's parameter being generated [16], [50].

In order for DE to efficiently optimize, the initial population must be distributed throughout the search space. Consequently, either a uniform or non-uniform distribution may be used depending on how much is known about the optimum location. However, non-uniform distribution or Gaussian distribution increases the chances of premature convergence [16]. Therefore, uniform distribution is preferred since they best reflect the lack of knowledge of the optimum location [16]. Hence, the uniform distribution is used for the purpose of this research.

After initialization, the population is mutated. This process is further discussed in the next section.

3.4 Mutation

Mutation is defined in [52] as a change in an organism's gene which results in an altered effect of its expression. This gene's alteration allows the organism to better adapt to its environment. In DE context, 'Mutation' is defined as a process of sampling four random vectors from the current population and manipulating them with simple arithmetic operations such as subtraction, addition and multiplication to form mutant vectors [51].

Equation 3.4 shows how the operation is done to obtain the mutant vector $V_{i,g}$. This process is also known as “differential mutation”.

$$V_{i,g} = X_{r0,g} + F \cdot (X_{r1,g} - X_{r2,g}), \quad r0 \neq r1 \neq r2 \quad (3.4)$$

The mutation scale factor F is a positive real number, $\in [0,2]$, that controls the rate at which the population evolves [16], [18], [50]. As mentioned before, to form one mutant vector, DE randomly selects four vectors from the current population. One of which is the parent, also known as “target vector”. This vector is used in the crossover and selection stages, and will be discussed in section 3.5 and 3.6. As for the remaining three vectors, namely the “base vector”, indexed with $r0$, and the “difference vectors”, indexed with $r1$ & $r2$, they are combined using equation 3.4 to create one vector mutation, as illustrated in Figure 3.2.

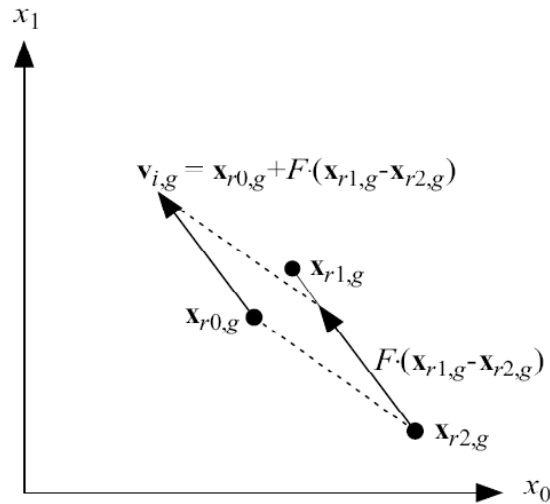


Figure 3.2: Differential mutation

The mutant vector, $V_{i,g}$, is obtained by adding the weighted differential, $F \cdot (X_{r1,g} - X_{r2,g})$, to the base vector $X_{r0,g}$.

The above process is repeated N_p -times to constitute a mutant population [16].

In order for DE to be effective, the following restrictions must be enforced:

- All selected vectors must be distinct from one another, $r0 \neq r1 \neq r2$;

- The base vector, \mathbf{b} , must only serve once per generation to avoid repetitive pick and omission of other vectors which might lead the algorithm to converge to a local optimum and lack of diversity in the population.

To ensure that the selected “base vector” only serves once, DE uses “stochastic universal sampling” [16]. This method replaces the traditional roulette wheel, see Figure 3.3, used in GA with a stationary pointer and slots size based on the objective function value. This wheel is spun N_p –times. Whereas the stochastic universal sampling consists of N_p equally spaced pointers and the roulette wheel only spins once [53]. Furthermore, in DE, all candidates have the same probability of being accepted. Hence, slots on the wheel, as in Figure 3.4, are equal size [16].

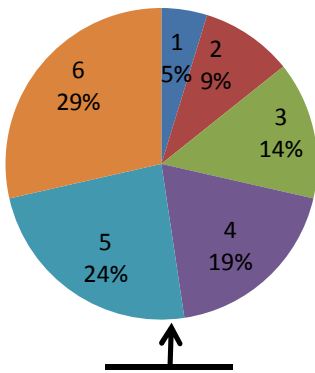


Figure 3.3: Traditional Roulette wheel spun N_p – times with fraction of allotted space dependent on vector objective function value

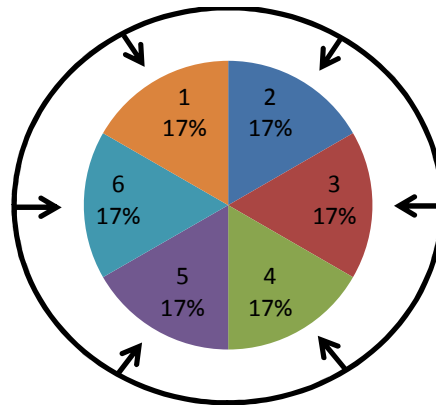


Figure 3.4: DE Stochastic universal sampling with equal N_p spaced pointers and equal slots size. The wheel is spun only once.

Several mutation strategies are used in literature to improve the global optimum search [16], [51], [54], [55] These strategies consist of:

3.4.1 Strategy DE/rand/1

This strategy is the classical version of DE, aforementioned. It is summarized in equation 3.4. Despite the fact of being widely use, this approach is particularly slow in converging to the optimum value [16], [56].

3.4.2 Strategy DE/best/1

This strategy is similar to DE/rand/1 with exception that the mutant vector is obtained as follows

$$V_{i,g} = X_{best,g} + F \cdot (X_{r1,g} - X_{r2,g}) \quad (3.5)$$

The best individual, $X_{best,g}$, is perturbed to find the mutant vector. This approach has a fast convergence and often to the local optimum [56].

3.4.3 Strategy DE/best/2

This strategy uses two mutation differences to create a mutant vector as shown in equation 3.6.

$$V_{i,g} = X_{best,g} + F \cdot (X_{r1,g} - X_{r2,g}) + F \cdot (X_{r3,g} - X_{r4,g}) \quad (3.6)$$

where $X_{r1}, X_{r2}, X_{r3}, X_{r4}$ are random mutually different vectors and $X_{best,g}$, the best individual of the current population. This approach attempts to balance between convergence speed and robustness [16]. However, it might converge to local optimum due to the fact that $X_{best,g}$ pulls the population toward its direction very quickly.

3.4.4 Strategy DE/local-to-best/2

Similar to DE/best/2, this approach uses two mutation differences in the mutation process as follows

$$V_{i,g} = X_{ro,g} + F \cdot (X_{best,g} - X_{r1,g}) + F \cdot (X_{r2,g} - X_{r3,g}) \quad (3.7)$$

Here, the mutant vector is obtained by randomly sampling the base vector to be perturbed by the scaled difference between the current best $X_{best,g}$ & X_{r1} and the scaled difference of 2 randomly sampled vectors. This approach often misses some of the best values due to the effect of the current best which influences the search direction [16].

3.4.5 Strategy DE/rand/2

This strategy is similar to both DE/local-to-best/2 and DE/best/2. However, all vectors are randomly selected and mutually different from each other.

$$V_{i,g} = X_{r0,g} + F \cdot (X_{r1,g} - X_{r2,g}) + F \cdot (X_{r3,g} - X_{r4,g}) \quad (3.8)$$

Where, $r0 \neq r1 \neq r2 \neq r3 \neq r4$. This approach is a little slower convergence. However, it is very robust [16], [56].

In this thesis, DE/rand/2 is used due to the aim set to appropriately tune the PSS with optimum time constants values for a robust performance. However, six vectors are randomly selected as opposed to four vectors previously mentioned.

3.5 Crossover

In biology, crossover is defined as the process of forming offsprings by genetically combining two different parents [57]. Whereas in DE, crossover refers to the process of combining a mutant vector with its ‘target vector’ to create a ‘trial vector’ [16], [50]. As described in section 3.4, each vector $X_{i,g}$ of the current population has a corresponding mutant vector $V_{i,g}$. Hence, by applying equation 3.9, a *D-dimensional* trial vector $U_{i,g} = [u_{1,i,g}, u_{2,i,g}, \dots, u_{D,i,g}]$, is obtained.

$$u_{j,i,g} = \begin{cases} v_{j,i,g} & \text{if } (rand_j(0,1) \leq CR \text{ or } j = j_{rand}), \quad j = 1, 2, \dots, D \\ x_{j,i,g} & \text{otherwise} \end{cases}$$

(3.9)

where $CR \in [0,1]$ is the crossover probability that controls the fraction of the parameter values that are copied from both mutant and trial vector [16]. To determine whether the parameter to be copied is from the mutant or trial vector, a random number, $rand_j$ between $[0, 1]$ is generated and compared to the predefined value of CR . If the generated number is less than or equal to CR , the parameter will be copied from $V_{i,g}$; else, the parameter is inherited from $X_{i,g}$. Whereas, j_{rand} is the random mutant parameter selected to be copied first to ensure that the trial vector is not a duplicate of target vector.

DE employs two forms of crossover, namely exponential crossover and uniform (binomial) crossover.

3.5.1 Exponential crossover

The exponential crossover consists of initially choosing a parameter from the mutant vector at random, to be copied to the corresponding trial vector. Thereafter, a random number $rand_j$ is generated and compared to the user defined crossover probability CR to determine the source of the subsequent parameter. i.e., if $rand_j(0,1) \leq CR$, parameters will be copied from the mutant vector. However, the first time that $rand_j(0,1) > CR$, the current and all the remaining parameters are copied from the target vector [16]. Figure 3.5 illustrates the exponential crossover.

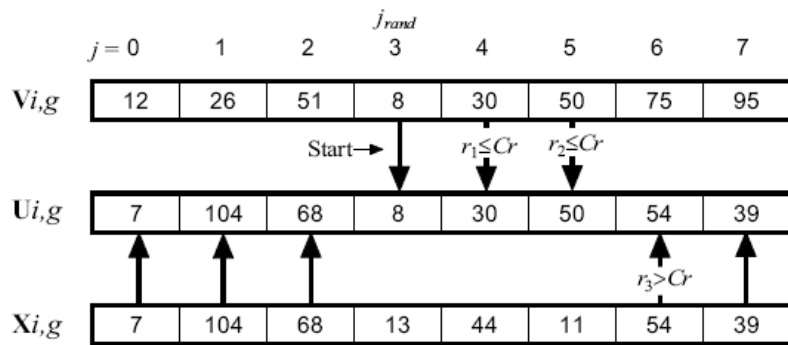


Figure 3.5: Exponential Crossover

The initial random parameter index, $J_{rand} = 3$, of the mutant vector which is copied to the corresponding location of the trial vector. As long as $rand_j(0,1) \leq CR$, parameters are copied from the mutant ($J_{rand} = 4,5$). However, when $rand_j(0,1) > CR$, the remaining parameters are copied from the target vector, $J_{rand} = 6,7,0,1$ and 2. This method presents some drawbacks. The trial vector has a greater probability of resembling the target vector. Hence, reduces the diversity of population [16]. To overcome this shortcoming, uniform crossover is used.

3.5.2 Uniform crossover

Similar to the exponential crossover, the initial random parameter index, J_{rand} , is copied from the mutant vector to the corresponding trial vector location to limit the risk of duplicating the target vector. However, in uniform crossover, all remaining parameters from both mutant and target vectors have an equal probability to be selected and copied [16]. Expressly, the trial parameter source is determined by whether $rand_j(0,1) \leq CR$ or

$rand_j(0,1) > CR$, to come from the mutant or target vector respectively. The uniform crossover is illustrated in Figure 3.6

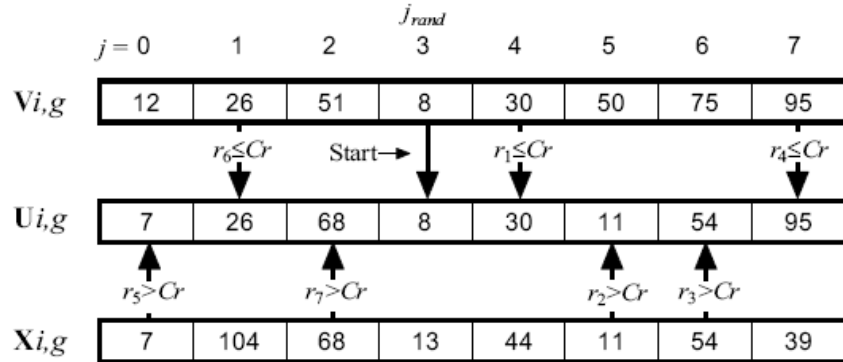


Figure 3.6: Uniform crossover

The uniform crossover has been used in this research due to the fact that it yield better results and maintain the population diversity [18].

3.6 Selection

This process consists of choosing individuals that will enter the next generation. In most EAs, selection pressure is applied when either choosing the vectors to recombine or when choosing survivors [16].

Traditionally, GAs employed parent selection based on the fitness values, whereby individuals with the highest fitness value are kept to undergo recombination or breeding to produce offsprings with superior characteristics [58].

GAs bias selection in favour of better individuals whereas DE employs selection pressure when picking the survivor to enter the next generation [16], [18], [50]. This type of selection can cause the search to prematurely converge to a local optimum [16]. To overcome this drawback, DE employs a “one-to-one survivor selection” which consists of comparing each trial vector to its corresponding target vector. This process ensures that the best vector at each index is retained. Furthermore, this also guarantees that the very best-so-far solution is kept. Even so, a trial vector that is better than most individuals in the current population will be rejected if its target is better, hence preserving the best solution [16]. However trial vectors that have the worst fitness are never accepted.

DE selection is done by applying equation 3.10 as follows

$$X_{i,g+1} = \begin{cases} U_{i,g} & \text{if } f(U_{i,g}) \leq f(X_{i,g}) \\ X_{i,g} & \text{otherwise} \end{cases} \quad (3.10)$$

The vector that enters the next generation, $X_{i,g+1}$, is the result of the comparison between the target vector $X_{i,g}$ and the trial vector $U_{i,g}$. Hence, for minimization, if $f(U_{i,g}) \leq f(X_{i,g})$, the trial vector enters the next generation; but if otherwise, the target vector is retained.

Once this stage is completed and a new population is formed, the processes of mutation, crossover and selection are repeated until a termination criterion is satisfied [18].

3.7 Termination

There are several ways to terminate an optimization process. The appropriate criterion is dependent on the function to be optimized. The following briefly describes different ways of termination.

3.7.1 Objective met

This criterion is used for function with known minimum or maximum value. Thus, the algorithm will compare whether it is within the specified tolerance and halt.

3.7.2 Maximum generation

The optimization is terminated when the maximum number of generations is attained. Thus, the best known solution is reported. This criterion is best suited for function with unknown minimum or maximum.

For the purpose of this research, this criterion best suit the termination due to the fact that the global maximum location is unknown, therefore investigated.

3.7.3 Population statistics

An optimization can be terminated when the difference between individuals with the worst fitness value and best fitness value is below a predetermined limit [16]. This is an indication that population is no longer diverse.

3.7.4 Limited time

Some applications are constrained by the time. In such case, manual intervention is required regardless of the number of generation.

3.8 Summary

The theory behind the Differential Evolution (DE) technique is presented in this chapter. DE is a parallel direct method that stochastically searches for the global optimum. DE relies on the differential mutation and the crossover operations to guide its search toward the prospect region.

Chapter 4

Population Based Incremental Learning (PBIL)

This chapter provides a background theory to Population-Based Incremental Learning (PBIL). An analysis on PBIL optimization approach is also provided.

4.1 Overview

Like most EAs, PBIL is an optimization technique that stochastically searches for the optimum value of a function by utilising some aspects of GAs combined with competitive learning to guide its search toward prospective areas [3], [7], [15], [17], [20], [59], [60]. Developed by Shumeet Baluja in 1994 [20], PBIL borrows its competitive learning concept from Artificial Neural Networks, and applies it to re-examine the performance and update the population of prospect solutions [3], [15], [20]. Furthermore, PBIL combines the aforementioned concept to that of GAs, such as initialization, mutation and binary encoded representation [20]. However, the crossover operator has been subtracted by redefining the role of the population [3], [7]. Hence, PBIL uses real-valued Probability Vectors (PV) to control the random generated bit strings and create other vectors through learning. PBIL then uses the current probability distribution to create N_p individuals which in turns, are evaluated with an objective function [3], [7], [15]. Using the best individuals of the current population, the probability vector is updated by shifting the likelihood of producing solutions corresponding to the best [3], [61]. There are other variants of PBIL such as using the best and the worst elements to update the probability vector (towards the best element and away from the worst element). One could also use more than one element to update the PV. However in this thesis only the best individual is used to update the probability vector.

4.2 Probability Vector (PV) and Population in PBIL

As mentioned in section 4.1, the Probability Vector (PV) is constituted of real-valued probabilities that are used to generate individuals of a population [17]. The probability vector size corresponds to the number of parameters, “ n ”, of an individual multiplied by the parameter binary encoding-length “ l ”. As given by equation 4.1.

$$PV = (p_1, p_2, \dots, p_m), \text{ where } m = n * l \quad (4.1)$$

For example, if an individual λ has 4 parameters encoded and each parameter is 2-bits representation in binary, PV will be constituted of 8 probability values, $PV = (p_1, p_2, \dots, p_8)$, where each entry represents a bit string.

To start off, the probability vector is initialized to 0.5 which is the equal probability to generate 1-or-0, bit string of given individual. This is illustrated in Figure 4.1.

Population #1	Population #2
0011	1010
1100	0101
1100	1010
0011	0101
Representation	Representation
0.5, 0.5, 0.5, 0.5	0.5 0.5 0.5 0.5

Figure 4.1: Probability vectors representation of 2 small population of 4

Notice that both PVs are similar, however the individuals are entirely different. Therefore, sampling from this vector will generate N_p random individuals. The population is a fundamental aspect in EAs; it provides the optimization with the ability to explore the search space in parallel from multiple points [3], [7], [15], [20]. However as the search progresses, the values in the probability vector will shift toward either 0.0 or 1.0, to represent the best individuals or solutions of a given problem [17], [20]. Hence, through the aforementioned process, PV has redefined the role of the population and the crossover operator [7]. In Figure 4.2, PV has been updated. The current probability of generating a “1” for the first element of individuals in the population is 1.0 whereas the second element is 0.75.

Population #2
1 0 1 0
1 1 0 0
1 1 0 0
1 1 0 0
Representation
1.0 0.75 0.25 0.0

Figure 4.2; Probability vectors representation after update

4.3 Competitive Learning

Competitive learning is a concept borrowed from Artificial Neural Network (ANN). It consists of grouping points based on their similarities with respect to the characteristics of the study. This process of clustering is achieved without prior knowledge of the number of groups or their features. Hence, Competitive Learning is given freedom to determine the most relevant features for each group and cluster each point to their respective groups based on these features [20].

Furthermore competitive learning applies the winner-take-all rule used by neurons that are competing within a layer to stay active and the rest are shut down [20]. In the same manner, PBIL only keeps the best solution to update the probability vector in order to represent individuals with similar characteristics [17], [20].

4.4 Mutation

Similar to most EAs, PBIL apply mutation on the updated probability vector with small perturbations on the position in the probability to maintain the diversity and prevent the probability vector from converging quickly to an extreme value (either 0.0 or 1.0), hence preventing premature convergence to a local optimum [20].

The mutation can be applied to the population or to the PV. In [20], the mutation operator was applied to the PV whereby the vector is shifted by a value in a randomly chosen

direction. However in this research, a forgetting factor has been included to relax the probability vector toward a neutral value of 0.5, [3], [62].

The mutation process is applied as follows.

For each locus $i = \{1, \dots, l\}$, if a random number $r = rand(0,1) < p_m$ (where p_m is the mutation probability), then mutate P_i as follows.

$$P'_i = P_i - p_m * (P_i - 0.5) \quad (4.2)$$

where p_m is the mutation shift that controls the amount by which a mutation can alter the value in each bit position. Following this operation, the cycle is repeated until N_p individual is reached.

4.5 Learning rate

During the probability vector (PV) update, the learning rate (LR) determines the distance by which PV is pushed for each iteration. In other words, LR determines how fast or slow PV is shifted towards the best individuals [3]. Equation 4.3 shows how LR is applied.

$$P_i(t + 1) = (1 - \alpha) * P_i(t) + \alpha * B_i(t), \quad i = \{1, \dots, l\} \quad (4.3)$$

where α is the learning rate that defines the amount by which PV is pushed. B_i is the best solution in generation t .

The magnitude of the learning rate plays a major role and its significance cannot be undermined. A larger rate speeds up convergence, but it reduces the function space to be searched, while a smaller rate will slow down the convergence, even though it increases the exploration of a bigger search space, thereby increasing the likelihood of better optimal solutions [3], [20].

4.6 Termination

Due to the fact that PBIL only stores the current best solution and the current solution being evaluated, and therefore it runs as long as the current best solution keeps being

updated [3]. For that reason, the program is terminated when the number of iterations reaches a specified maximum number.

4.7 Summary

The theory behind PBIL techniques has been presented in this chapter. In PBIL, the roles of the population and the crossover operator have been substituted with a probability vector (PV). This vector controls the random generated bit strings to create individuals. PBIL uses PV to guide its search toward prospective solution by updating the vector through a competitive learning process which in turn, is used to produce individuals with the same likelihood as the best individual or solution.

Chapter 5

Power System Stabilizer Design

This chapter discusses the design procedure of a Power System Stabilizer (PSS). The following consists of using the conventional method and the application of the two aforementioned optimization techniques, namely Differential Evolution (DE) and Population-Based Incremental Learning (PBIL). In addition, test cases for various system configurations are also discussed in this section.

5.1 Introduction

As mentioned in chapter 1, PSS is a device that aids in subsiding low frequency oscillations observed in a system by providing supplementary signal. Generally, conventional control methods are used to tune the PSS' parameters by linearizing the system at a particular operating condition. However, its response becomes inadequate when operating conditions change. As a result, the PSS requires constant re-tuning. Alternatively, find parameters that will guarantee adequate performance of the PSS over a wide range of operating conditions. Hence, the problem of selecting the PSS' parameters which simultaneously stabilizes a set of plants was converted to an optimization problem solved by Differential Evolution (DE) and Population-Based Incremental Learning (PBIL) using an eigenvalue based objective function [3], [6]. These computational methods are used to find the optimum parameters. The resulting PSSs are tested under various conditions for robust performance. However several factors affecting the overall system performance, such as the objective function and controller parameters, must be considered to achieve the desired performance. The CPSS design and the objective function are discussed in subsequent sections.

5.2 Conventional PSS design

The conventional PSS consists of a washout block, a phase compensator block (lead/lag) as well as the gain. These parameters, as described in equation 5.1, determine the

performance of the PSS. The block diagram of the equation 5.1 was shown in chapter 1 and reproduced here for easy reference.

$$V_{pss}(s) = K_{pss} \frac{sT_w}{1+sT_w} \frac{1+sT_1}{1+sT_2} \frac{1+sT_3}{1+sT_4} \Delta\omega(s) \quad (5.1)$$

K_{pss} is the gain of the PSS, T_1 to T_4 are lead/lag time constants, T_w is the washout time constant. V_{pss} is the output signal of the PSS in volts, whereas $\Delta\omega$, the speed deviation, is the input signal.

Conventionally, phase compensation and root locus are used to tune the PSS' parameters. These methods consist of identifying the system phase lag between the reference voltage and the electrical torque, which is due to the automatic voltage regulators (AVRs) and fast acting exciters [25], [4] as shown in Figure 5.1 to 5.2. Then determine the PSS phase lead by fine tuning the lead/lag parameters, $T_1 - T_4$, needed to compensate the lag angle. Once the time constants have been determined, the system root locus must be plotted. The PSS gain is chosen such that the rotor mode eigenvalues is maximized without adversely affecting other modes.

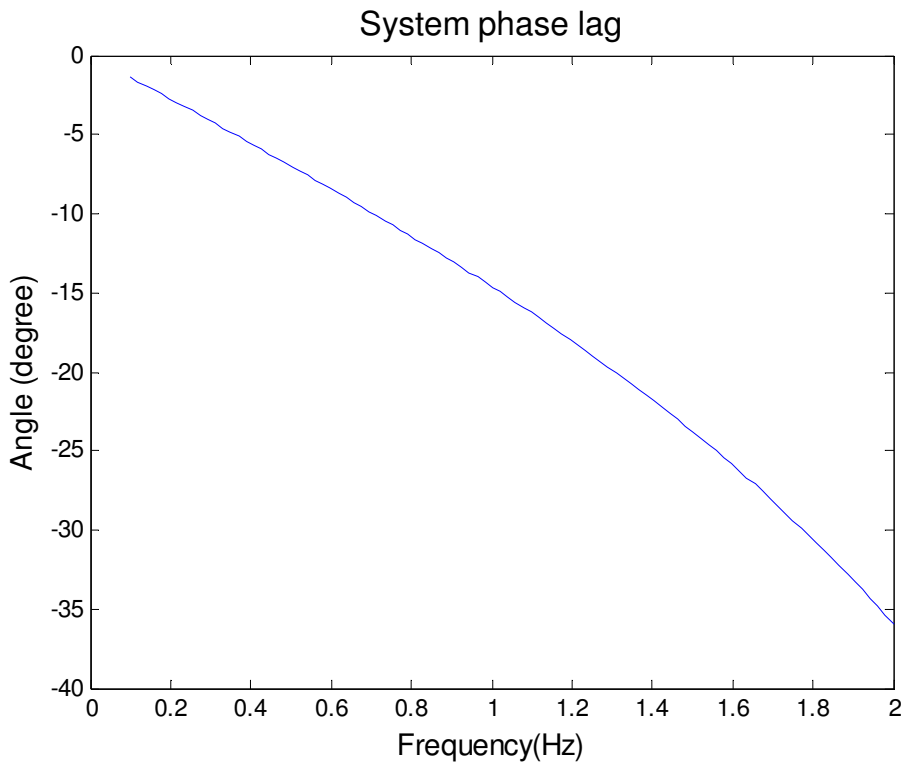


Figure 5.1: Phase Lag in an SMIB system

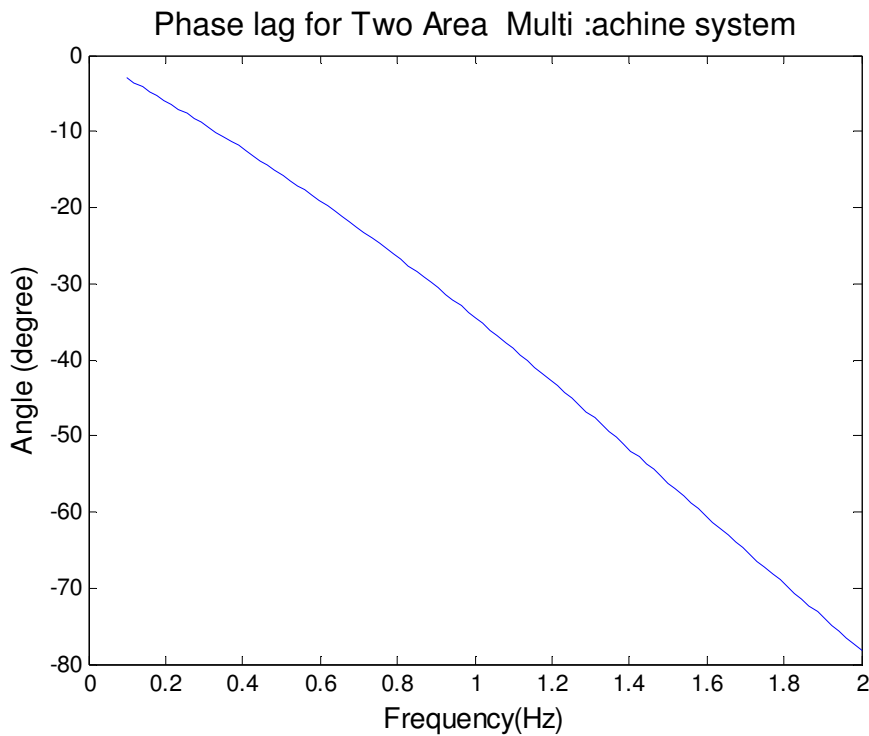


Figure 5.2: Phase lag in a Two Area Multi Machine system

Figure 5.1 and 5.2 show the different phase lags in a single machine infinite bus (SMIB) and a two-area multi machine (2AMM). In the SMIB, the phase lag is found to be 10° at a frequency of 0.7 Hz whereas the two area multimachine has phase lag of 20° at a frequency of 0.65 Hz for the inter-area mode and 40° at a frequency of 1.15 Hz for the local modes. These phase lags are corrected with the application of PSSs. However tuning these parameters is not an easy task, especially in a multi machine system. The selection of PSS parameters and the location of the PSS in the system are very critical. There are many modes of consideration in multi-machine as previously discussed. The PSS might improve the damping of one electromechanical mode, but may have adverse effect on other modes, such as the control modes. Whereas in single machine to infinite bus (SMIB), the process is relatively simpler since the focus is to find optimal values of the PSS that will help improve the stability of local modes only. Hence, EAs are applied to simultaneously find optimum sets of parameters for different system setups. However EAs, (PBIL and DE), both require an evaluation function, also known as objective function, to find these optimum parameters.

5.3 Objective function

To get a better understanding of an objective function, let define the term optimization. In simple terms, an optimization can be defined as an attempt to maximize a system's desirable properties whilst simultaneously minimizing its undesirable characteristics [16]. These properties and their effectiveness are dependent on the problem. Often, the problem is represented as a function. For example, in this thesis, tuning the PSS' parameters in order to add maximum damping ratio to the least damped mode is the optimization goal. Mathematically, this can be represented by equation 5.2, also known as objective function. In EA, the objective function can be defined as a performance measure of individuals within a population. Moreover, it gives the proximity of a point to the global maximum/minimum.

In this instance, the problem of finding the optimum set of PSS' parameters to stabilize the system over a wide range of operating conditions have been converted to an optimization problem with the objective of improving the system lowest damping ratio. Subsequently, this will ensure that all closed-loop poles remain on the negative side of the s-plane despite the changes in operating condition [3], [6].

Therefore the objective function is formulated as follows [3]:

$$Val = \max(\min(\zeta_{i,j})), \text{ where } i = 1,2, \dots, n \ \& \ j = 1,2, \dots, m \quad (5.2)$$

Where ζ is the lowest damping ratio. The index 'i' represents eigenvalue of the 'jth' operating conditions.

$\zeta_i = \frac{-\sigma_i}{\sigma_i^2 + \omega_i^2}$ is the damping ratio of the ith eigenvalue

σ_i, ω_i are the real part and imaginary part (frequency of oscillation) of the ith eigenvalue respectively.

The objective function can be stated as:

Maximize the lowest damping ratio

Subject to:

$$K_j^{min} \leq K_j \leq K_j^{max} \quad \text{where } j = 1, 2, \dots, m$$

$$T_{i,j}^{min} \leq T_{i,j} \leq T_{i,j}^{max} \quad \text{where } i = 1, 2, \dots, n \text{ \& } j = 1, \dots, m \quad (5.3)$$

K and T_i denote the optimized controller gain and the lead lag time constants respectively within their respective boundaries. The index ‘ i ’ denotes the lead lag parameter number whereas ‘ j ’ is the machine number. In this thesis the following values were used:

$$K_j^{max} = 20; T_{i,j}^{max} = 5 \text{ seconds}$$

$$K_j^{min} = 1; T_{i,j}^{min} = 0.001 \text{ seconds.}$$

The objective function was applied to various system configurations, such as SMIB and two-area multi machine systems. These systems are described in section 5.5.

5.4 Application of DE & PBIL to PSS design

In this section, DE & PBIL are applied to design the PSS based on the mathematical model aforementioned. Once the parameters are obtained, they are tested for the robustness in the following chapter. However, the method of tuning essentially involves the following steps for all system configurations.

Step 1. Define the range of operating conditions.

Step 2. Test them by running the load flow. Eliminate those conditions for which the load flow does not converge.

Step 3. Linearize the system at each operating condition and store the state-space A, B, C, D matrices for each condition.

Step 4. Design the PSSs by solving the constrained optimization problem given by equation 5.2 using DE and PBIL. This is applied to all operating conditions

Step 5. Once the optimum parameters are found, they are tested for robustness.

Step 6. In the case where the resulting PSSs performances are not good, go to step1.

These steps are summarized by the flow charts in Figure 5.4 and 5.5 representing the application of DE and PBIL respectively.

Following the obtainment of PSSs’ parameters based on DE and PBIL, their performances are evaluated by conducting a modal analysis. The results are then validated through time domain simulations. Furthermore, a comparison analysis is carried out for different designs and CPSS.

However, to obtain the optimum PSSs' parameters, both algorithms had to be configured. DE requires user-define parameters inputs such as the mutation factor ' F ', the crossover probability ' CR ', the population number ' N_p ' and the maximum number of generations ' g '. These parameters greatly influence DE's ability to find the global minimum/maximum. Despite the recommendations provided by the authors in [16], F & CR values were determined by trial-and-errors. The population also affects the optimization [16], [63], [64]. In [16], it is recommended that the population should be 10 times the dimensionality (or parameters to be optimized), 10D, however in this thesis 50, was found to give best results. Some preliminary work with regards to PSS tuning are presented in chapter 8 together with self-adaptive mutation and crossover parameters. Whereas PBIL has fewer user-define inputs such as Learning rate (LR), forgetting factor (FF) and population. Similar to [3], LR was set to 0.1 and FF to 0.005 within a population of 50 for comparison sake. The generation however, was set to 500 to allow for enough learning within the optimization [3], [6], [15]. Figure 5.3 summarizes the parameters used for both algorithms.

<p><u>DE PARAMETERS</u> Population: 50 Generation: 200 Mutation F: 0.95 Crossover CR: 0.95 Termination: Max gen</p>	<p><u>PBIL PARAMETERS</u> Population: 50 Length of encoded solution (15 bits) Generation: 500 Learning rate: 0.1 Forgetting factor: 0.005 Termination: Max gen</p>
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Figure 5.3: DE and PBIL parameters

Application of Differential Evolution to Power System Stabilizer design

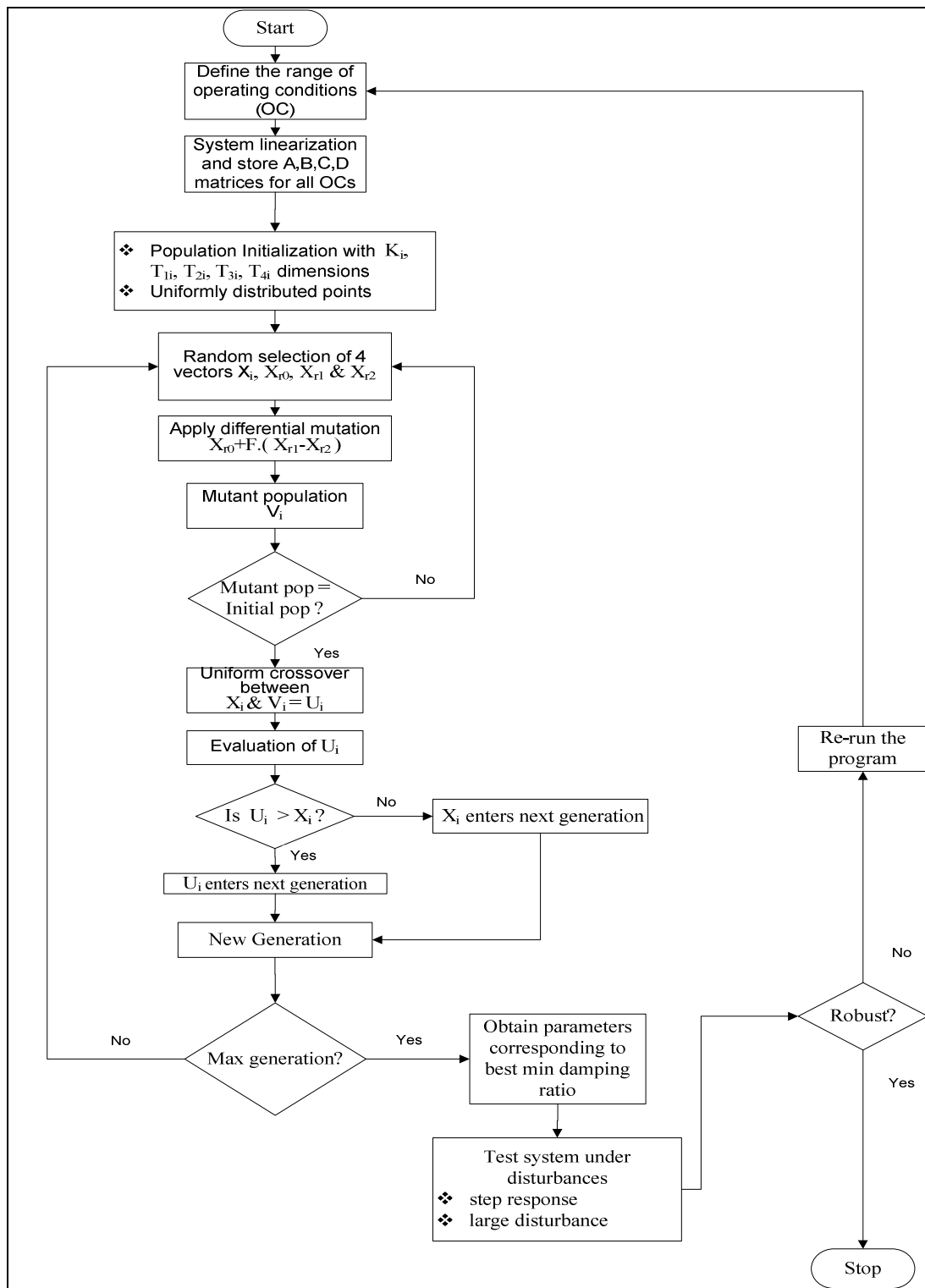


Figure 5.4: chart for the PSS design using DE

Application of Differential Evolution to Power System Stabilizer design

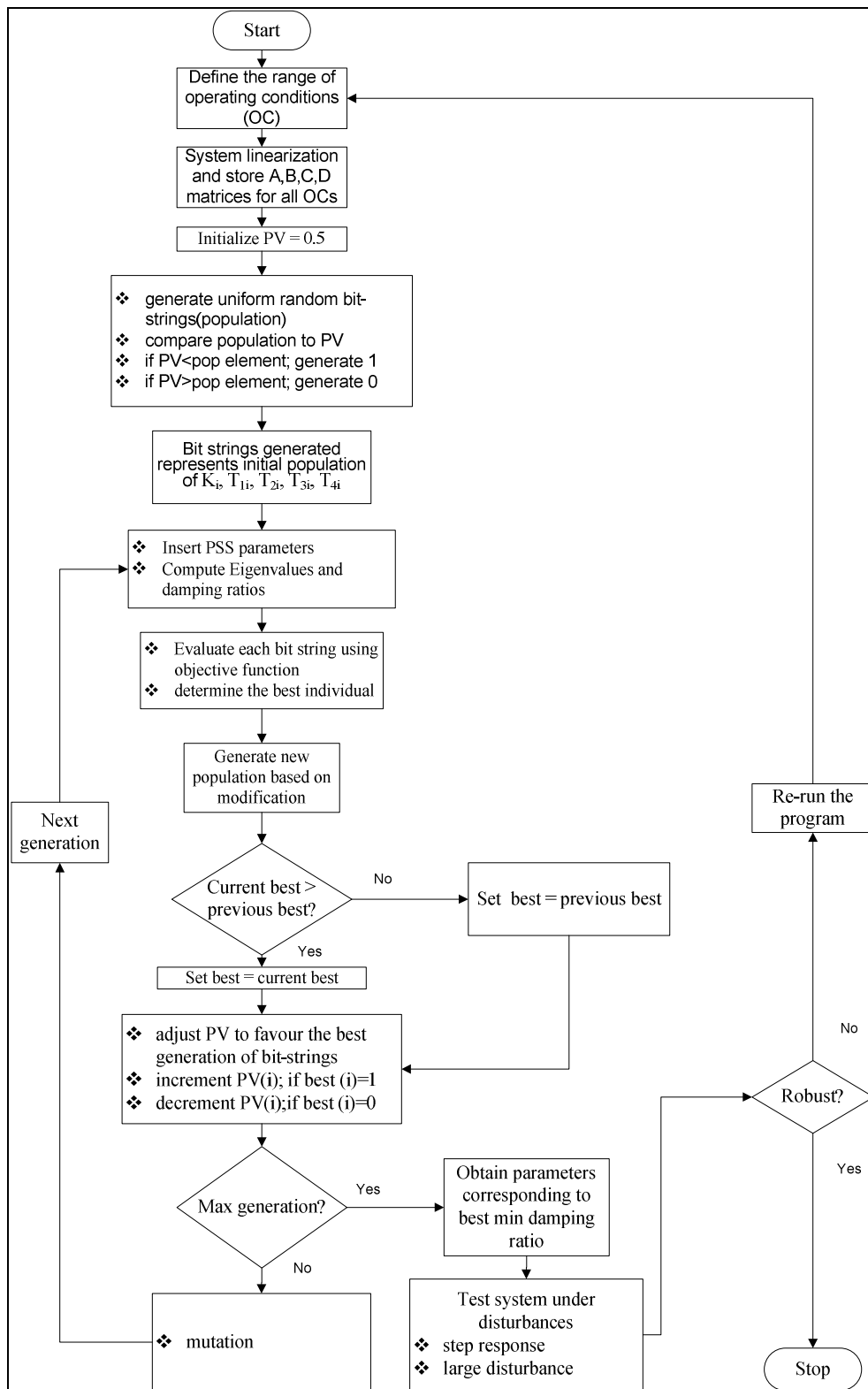


Figure 5.5: chart for the PSS design using PBIL

5.5 System configurations

In this thesis, three system configurations were considered for the PSS design. The Single Machine to Infinite Bus (SMIB), the Two-area Multi-Machine and the New England 10-Machines systems. These set-ups are further discussed in subsequent sections.

5.5.1 Single Machine to Infinite Bus system

This system consists of a synchronous generator connected to the infinite bus through a double transmission line as illustrated by Figure 5.6. The generator was modelled using a detailed 6th order differential equations and is equipped with a simple exciter which was modelled using a first order equation and a turbine governor. The parameters are given in Appendix B. The operating conditions of this system have been obtained by simulating variations of generator output and transmission line reactance. Hence, by varying the generated power from 0.5pu to 1.0 pu, it simulates a lightly loaded system to a heavily loaded one whereas the variation of the line reactance, from 0.5pu to 1.0pu, simulates the behaviour of a system with a strong tie-lines to weak tie-lines. The operating conditions used in the design of the PSS are given in Table 5.1 together with the eigenvalues and damping ratios.

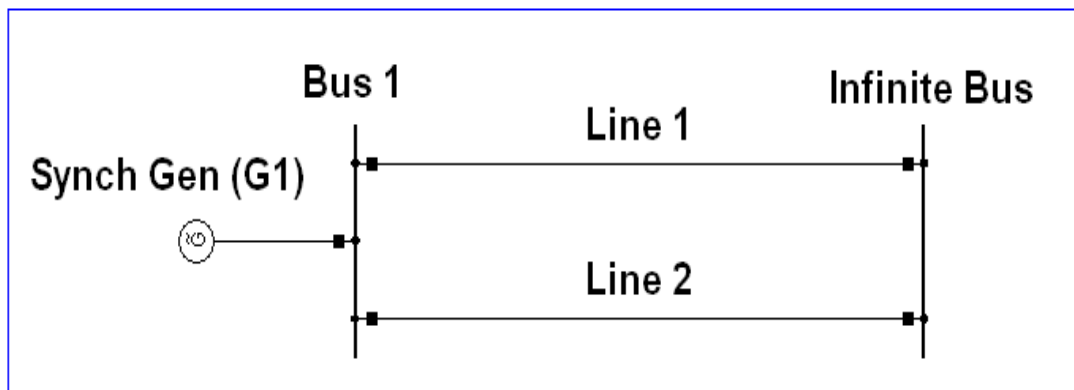


Figure 5.6: Single machine to infinite bus system

The operating condition range consists of five cases used to tune the PSS. Table 5.1 lists the open-loop system (or without PSS). Case 1 corresponds to the nominal operating condition with an active power of 1.0pu and line reactance of 0.5pu. In this case, the system is poorly damped with 4.7% which is below the power system acceptable margin of 5% [4]. The system oscillates at a frequency of 0.59 Hz and decay away in about 23 seconds.

As the line reactance was increased by 50% to 0.7 and active power decreased by 50% to 0.5 pu for Case 2, the damping ratio reduced to 4.4%, oscillating at a frequency of 0.52 Hz which decay after 27 seconds. In Cases 3 and the line reactance was kept constant at 0.7 pu whereas the active power was increased to a maximum loading of 1.0pu. The frequency of oscillation for the electromechanical modes is 0.51 Hz with damping ratio of 3.7 %. In case 4, the power was reduced by 50% whereas the line reactance was increased by 100% from the nominal condition to 0.5pu and 1.0pu respectively. The system oscillates at a frequency of 0.47 Hz with a damping of 3.6% with a settling time of 37 seconds. In Cases 5, the line reactance was kept constant at 1.0pu and the power was set to maximum loading of 1.0pu. This case simulates the maximum power transfer on weak tie-line. The frequency of oscillation for case 5 was 0.44 Hz with damping ratio of 2.7%. These oscillations settled after 53 seconds.

Table 5.1: SMIB Open Loop operating conditions used in the PSS design

Case	Active Power (pu)	Reactive Power (pu)	Line reactance (pu)	Eigenvalues	Damping ratio
Case 1	1.0	0.32	0.5	-0.177 ± j3.764	0.047
Case 2	0.5	0.17	0.7	-0.151 ± j3.395	0.044
Case 3	1.0	0.39	0.7	-0.119 ± j3.245	0.037
Case 4	0.5	0.16	0.9	-0.108 ± j2.94	0.036
Case 5	1.0	0.47	0.9	-0.076 ± j2.781	0.027

5.5.2 Two Area Multi-machine System

The system shown in Figure 5.7 consists of two areas, each with identical generating units and similar ratings [1], [3]. All the generators are equipped with simple exciters (see Appendix C) and similar turbine governors. The system's parameters are given in Appendix C. All the generators are 6th order whereas the AVRs are first order differential equations. In addition, the system has two loads connected to bus 4 and 14 as illustrated in Figure 5.7. The two areas are connected by two tie-lines.

The operating conditions were obtained by varying the loads' demand which subsequently varied the power transferred over the tie-lines. Hence, six cases (or operating conditions), given in Table 5.2, were used to design and test the PSSs. Furthermore, eigenvalues and damping ratio related to both inter area mode and local modes are listed on the same table.

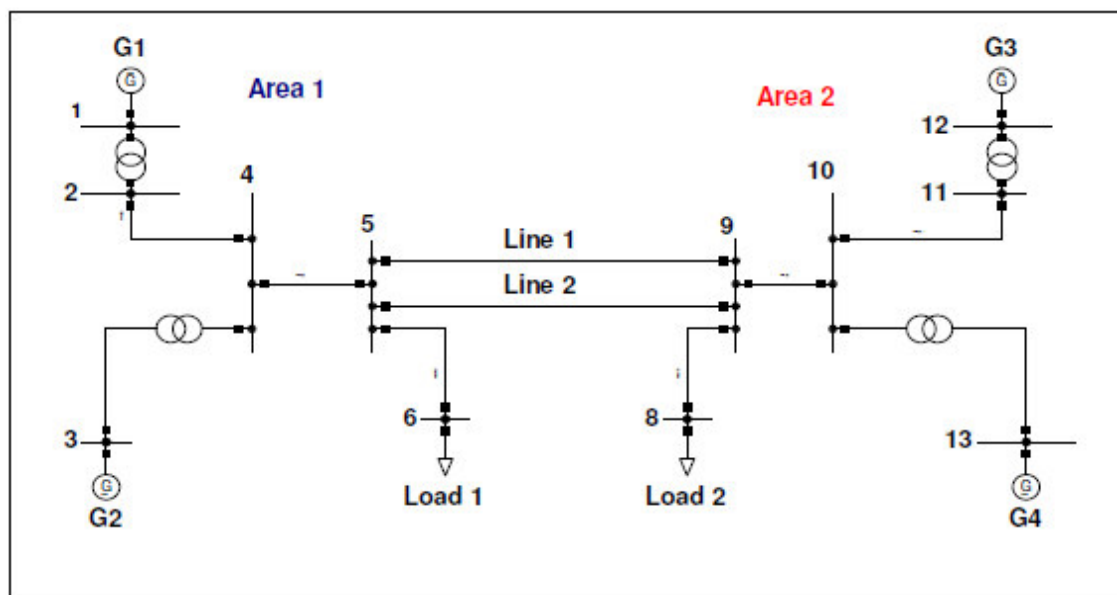


Figure 5.7: Two Area multi-machine system line diagram

Case 1 is the nominal operating condition. The load at bus 4 was set at 1167MW, whereas the one at bus 14 was set to 1467MW. The total generation in Area 1 was 1330 MW whereas 1230MW was generated by Area 2. As a result, the power transferred from Area 1 to Area 2 was 100 MW. The inter-area mode oscillates at a frequency of 0.78 Hz and a damping ratio of 0.1%. Hence, these oscillations sustain for a long period. Both local modes are marginally damped with 8.3% and 6.5 % where generators oscillate

against each other at frequencies of 1.15 Hz and 1.18Hz for Area 1 and Area 2 respectively. For case 2, load 1 and 2 were set to 1167MW and 1567MW respectively. The total generation in Area 1 was 1405MW whilst 1400MW in Area 2. The power transmitted from Area 1 to Area 2 was observed to be 200MW. The inter-area mode was found to be unstable characterized by a negative damping ratio of -0.9% hence growing oscillation of frequency of 0.77 Hz. Furthermore, the local modes oscillate at a frequency of 1.16Hz and 1.18Hz. They have similar damping ratios of 6.7% and settle in 9 seconds. In case 3, 300 MW was flowing through the tie-lines. The loads were set to 1067MW and 1667MW. Consequently, the eigenvalue corresponding to inter-area mode was further destabilized as the corresponding poles were moving further into the right hand side of the s-plane.

In case 4 & 5, the power transmitted was 400MW and 500MW respectively, from Area 1 to Area 2. Both cases are unstable, indicated by the negative damping of -1.68% and -1.64%. Their frequency of oscillation has been further reduced from nominal case to 0.76Hz and 0.74Hz respectively. The local mode in Area 1 is relatively well damped with 7.07% in both cases. However, the local mode in Area 2 is poorly damped with 5.0% and 4.8% for case 4 and 5. Table 5.2 shows that the variation of the power transmitted between the two areas negatively affect the electromechanical modes, both inter-area and local modes, by reducing their relative damping ratios. Hence it can be observed that the damping ratio related to the inter-area is poorly damped and in some cases negatively damped. Whereas the damping related to the local modes have low damping but still enough for the modes to be classified stable and in a good state [4]. In Case 6, the power transmitted was 500MW over one-tie line from Area 1 to Area 2. It can be observed that the inter-area frequency of oscillation has reduced to 0.544Hz and the damping ratio slightly decreased to -1.7%.

Table 5.2: Two-Area Open loop poles for the selected design operating conditions

Case	Active Power	Tie-lines	Inter-area mode	Local area mode 1	Local area mode 2
1	100MW	2	$-0.006 \pm j4.962$ (0.0012)	$-0.606 \pm j7.28$ (0.0829)	$-0.488 \pm j7.42$ (0.0656)
2	200MW	2	$0.0461 \pm j4.937$ (-0.0093)	$-0.4916 \pm j7.290$ (0.0735)	$-0.4990 \pm j7.420$ (0.0670)
3	300MW	2	$0.0428 \pm j4.894$ (-0.0087)	$-0.4955 \pm j07.280$ (0.0707)	$-0.5069 \pm j7.405$ (0.0683)
4	400MW	2	$0.0806 \pm j4.792$ (-0.0168)	$-0.5178 \pm j07.300$ (0.0678)	$-0.3721 \pm j7.344$ (0.0506)
5	500MW	2	$0.0771 \pm j4.697$ (-0.0168)	$-0.5357 \pm j7.269$ (0.0673)	$-0.3591 \pm j7.335$ (0.0488)
6	500MW	1	$0.0589 \pm j3.419$ (-0.0170)	$-0.4550 \pm j7.330$ (0.0617)	$-0.3375 \pm j7.224$ (0.0460)

5.6 Summary

This chapter presented the different procedures used to design and tune the conventional PSS, the PBIL-PSS and the DE-PSS optimization tuning approaches. The two EAs procedures use an objective function based on maximizing the lowest damping ratio to find the optimum PSS parameters. These methods were applied to two different system models. On the SMIB, the PSS was designed over 5 operating conditions to improve the open-loop eigenvalues. In the two-area multi machine system, the PSS was design over six operating conditions to improve the dynamic of the inter-area mode.

Chapter 6

Single Machine Infinite Bus System (SMIB): Simulation Results

6.1 Introduction

This chapter presents the simulation results for the single machine connected to the infinite bus system. The PSS' parameters are tuned based on the system minimum damping ratio using DE and PBIL. The performance of the PSSs is assessed via the modal analysis and validated using time domain simulations. The effectiveness of the resulting PSSs to damp the low frequency oscillations is tested under various operating conditions. The results are validated through simulations of the system's response for five different operating conditions. The comparison is carried out between the system equipped with CPSS, DE-PSS and PBIL-PSS.

6.2 Optimized PSS

Following the application of DE and PBIL to tune the PSS, Figures 6.1 and 6.2 show the optimization fitness curves of both algorithms. The optimization consisted of finding a set of parameters that would give the best damping for the most dominant poles of the system. Figure 6.1 shows DE converging to an optimum damping ratio value of 0.2659 corresponding to the electromechanical damping ratio of case 5. On the other hand, Figure 6.2 shows that PBIL converged to an optimum damping ratio value of 0.2325. These results indicate that DE performs slightly better than PBIL in finding the optimum value.

It is deduced from Figure 6.1 that DE is able to explore the search space to only converge after 100 generations for a maximum of 200 generations. PBIL, on the other hand, explores the search space in a parallelized manner and starts settling after 150 for a maximum of 300 generations. The difference in the maximum number of generations is to allow PBIL to settle since its algorithm is based on learning.

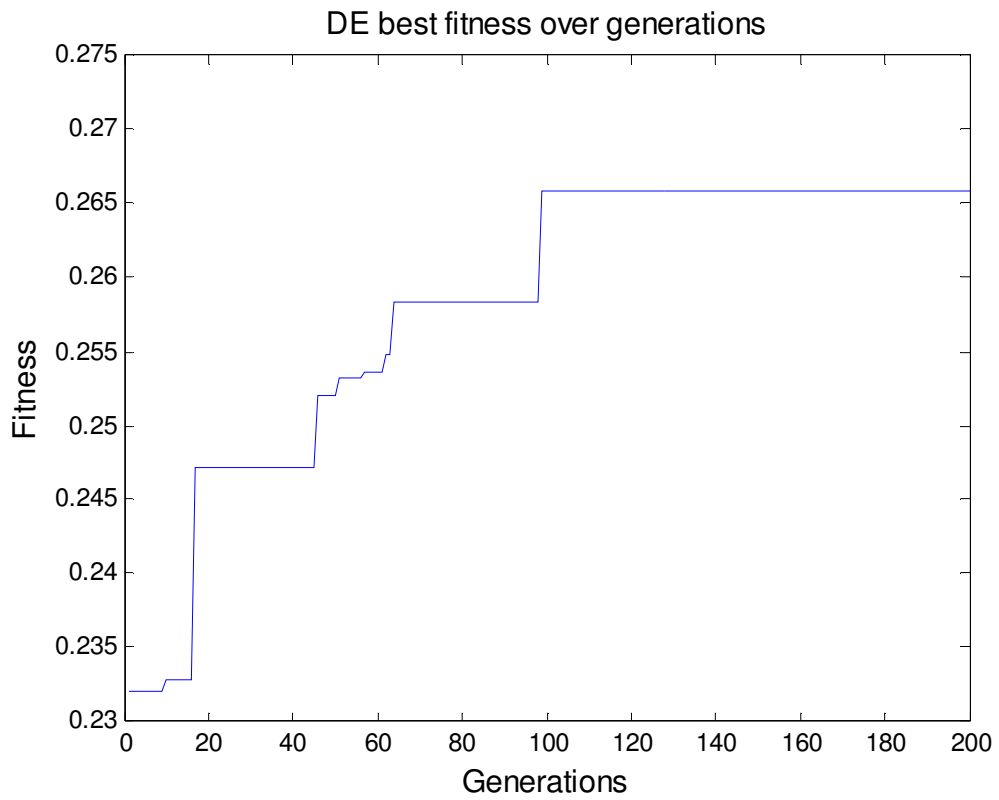


Figure 6.1: DE fitness curve

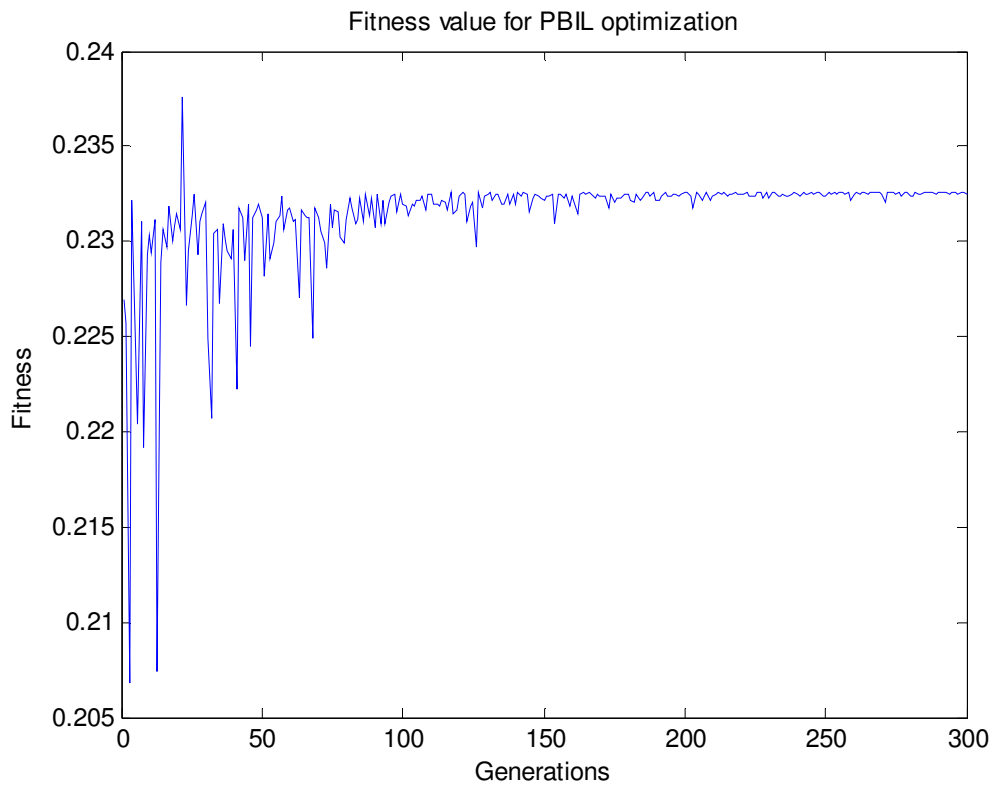


Figure 6.2: PBIL fitness curve

Table 6.1 shows the DE, PBIL and CPSS parameters obtained after optimization.

Table 6.1: SMIB PSS parameters

	K	T1	T2	T3	T4
CPSS	13.868	3.7440	0.8778	3.7440	0.8778
DE-PSS	19.985	4.9376	0.4339	0.0103	0.1290
PBIL	16.361	1.7217	1.8110	4.9435	0.5568

6.3 Modal Analysis

Table 6.2 shows the eigenvalues corresponding to the electromechanical modes of the closed-loop and the open loop system and the respective damping ratio in brackets. As expected, the implementation of all PSS has significantly improved the system. In Case 1, the nominal condition, DE has a slightly higher damping ratio of 52.1% whereas PBIL has 45.15% and CPSS, 34.73%. It can be deduced that, DE based PSS provides 6.92% more damping than PBIL and 17.37% more damping than CPSS. PBIL on the other hand provides 10.42% more damping than CPSS. From the analysis of the eigenvalues, it can be deduced that DE-PSS has transients that decay to within the $\pm 2\%$ band in less than 2.5 seconds whilst PBIL-PSS and CPSS have transients that decay in 3 and 3.5 seconds, respectively. In Case 2, as the transmission lines are increased by 50% and the loading decreased, the damping ratio provided by all PSSs slightly drops. DE-PSS damping ratio is reduced by 6% to a value of 45.37% whilst that of PBIL is reduced to 38.96% and that of CPSS to 30.49%. With this, it is deduced that the PSSs will damp the oscillations to $\pm 2\%$ of their final values in 3, 4 and 4.5 seconds for DE, PBIL and CPSS, respectively. As the power transmitted is increased to 1.0 p.u for case 3, the damping is further reduced for all PSSs to 39.11%, 30.14% and 23.31% for DE-PSS, PBIL-PSS and CPSS, respectively. DE-PSS has a damping ratio of 37.98%, PBIL and CPSS damping ratios of 33.63% and 26.02%, respectively. However, when the system is heavily loaded combined with weak transmission lines (case 5), DE-PSS has a damping of 26.59 % whilst PBIL has a damping ratio 23.25% and the CPSS has the lowest damping ratio of 18.55%. However it is noticeable that the addition of PSSs has slightly reduced the frequency of oscillation of the electromechanical mode for all cases.

Table 6.2: SMIB electromechanical modes of the system with different PSSs Designs

Cases	DE-PSS	PBIL-PSS	CPSS	No PSS
1	$-1.6950 \pm j2.7780$ (0.521)	$-1.3296 \pm j2.6279i$ 0.4515	$-1.1033 \pm j 2.9792$ 0.3473	$-0.177 \pm j3.764$ (0.047)
2	$-1.2323 \pm j2.4201$ (0.4537)	$-0.9570 \pm j2.2621$ (0.3896)	$-0.8089 \pm j2.5271$ (0.3049)	$-0.1512 \pm j3.3950$ (0.0445)
3	$-1.1576 \pm j2.7233$ (0.3917)	$-0.8936 \pm j2.8265$ (0.3014)	$-0.7281 \pm j3.0379$ (0.2331)	$-0.119 \pm j3.245$ (0.037)
4	$-0.9102 \pm j2.2167$ (0.3798)	$-0.7160 \pm j2.0048$ (0.3363)	$-0.5957 \pm j2.2107$ (0.2602)	$-0.108 \pm j2.94$ (0.036)
5	$-0.7065 \pm j2.5401$ (0.2659)	$-0.6520 \pm j2.6040$ (0.2325)	$-0.5114 \pm j2.7087$ (0.1855)	$-0.076 \pm j2.781$ (0.027)

6.4 Time Domain Simulation

The time domain simulations were performed to validate the results of modal analysis. The effectiveness of the PSSs is assessed by their ability to damp low frequency oscillations under various operating conditions. Furthermore, the PSS must be able to stabilize the system under transient conditions. Therefore, two types of time domain simulations are performed; small signal and transient simulations.

6.4.1 Small Disturbance Simulation

The small disturbance simulations have been performed for all five cases previously mentioned by applying 10% step change in the reference voltage. The speed deviation responses of the generator under the step change in the reference voltage are presented in Figure 6.3 to Figure 6.7.

Figure 6.3 illustrates the step response of all PSSs for the nominal operating condition (Case 1). The results show that all PSSs are able to adequately damp the oscillations; hence improving the system dynamic stability. Both DE-PSS and PBIL-PSS settled within 3 seconds whilst CPSS settled in about 3.5 seconds. However DE-PSS has the least overshoot and CPSS the highest overshoot whereas PBIL has the least undershoot.

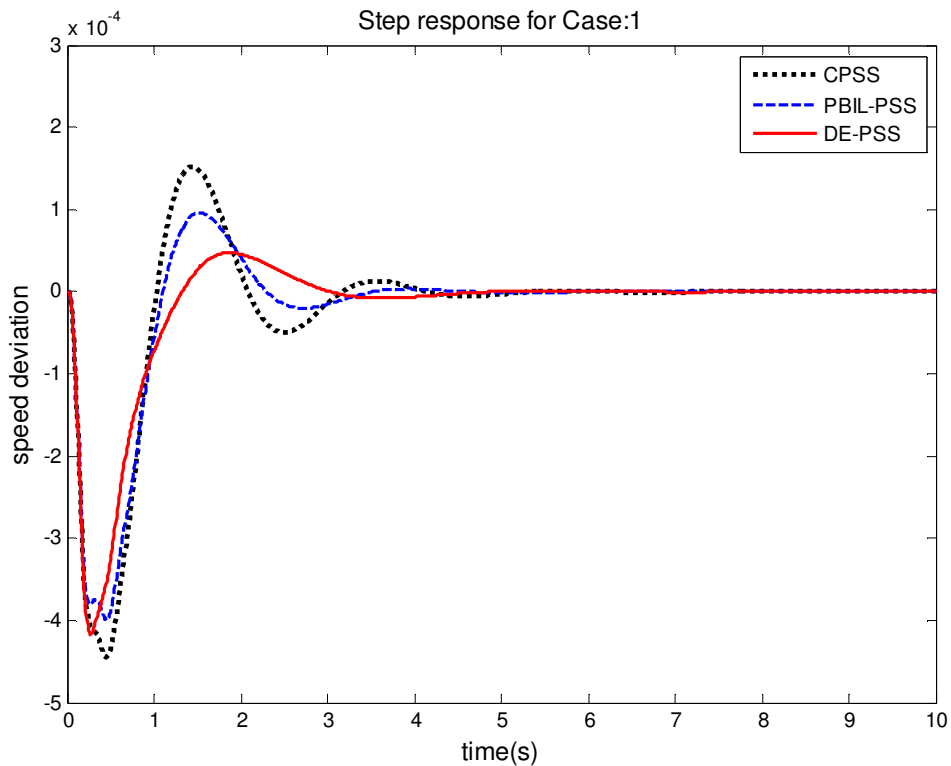


Figure 6.3: Rotor speed responses for Case 1

Figure 6.4 shows the responses of the rotor speed deviations for case 2. It is observed that DE-PSS displays better performance with a settling time of about 3 seconds as compared to a settling time of 4 and 5 seconds, respectively for both PBIL-PSS and CPSS. Moreover, DE-PSS has the lowest overshoot and similar undershoot with that of PBIL-PSS. CPSS on the other hand, has more oscillations and higher overshoots and undershoots as compared to the other PSSs. The performance of the CPSS is degrading with changes of operating conditions as expected.

Figure 6.5 shows the generator speed responses for case 3. Although the system remains stable under all the PSSs, PBIL-PSS and DE-PSS improve the system dynamics significantly. However, DE-PSS displays remarkable performance with settling time of about 4 seconds. Furthermore, DE-PSS is better in terms of overshoots and undershoots. PBIL-PSS settles within 5 seconds and the CPSS within 6 seconds.

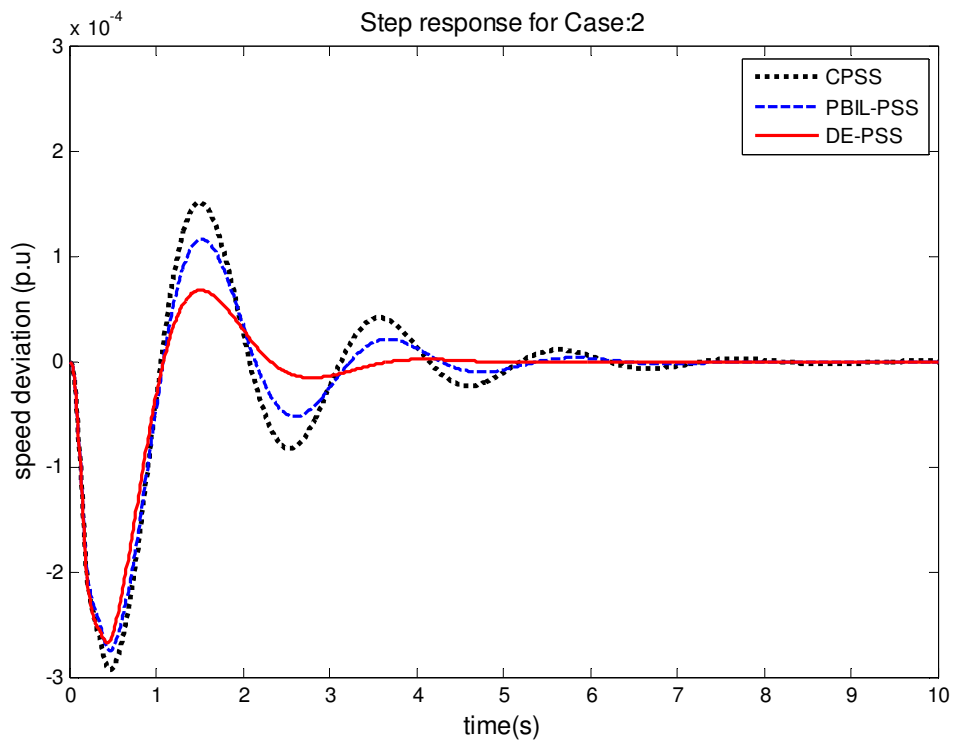


Figure 6.4: Rotor speed responses for Case 2

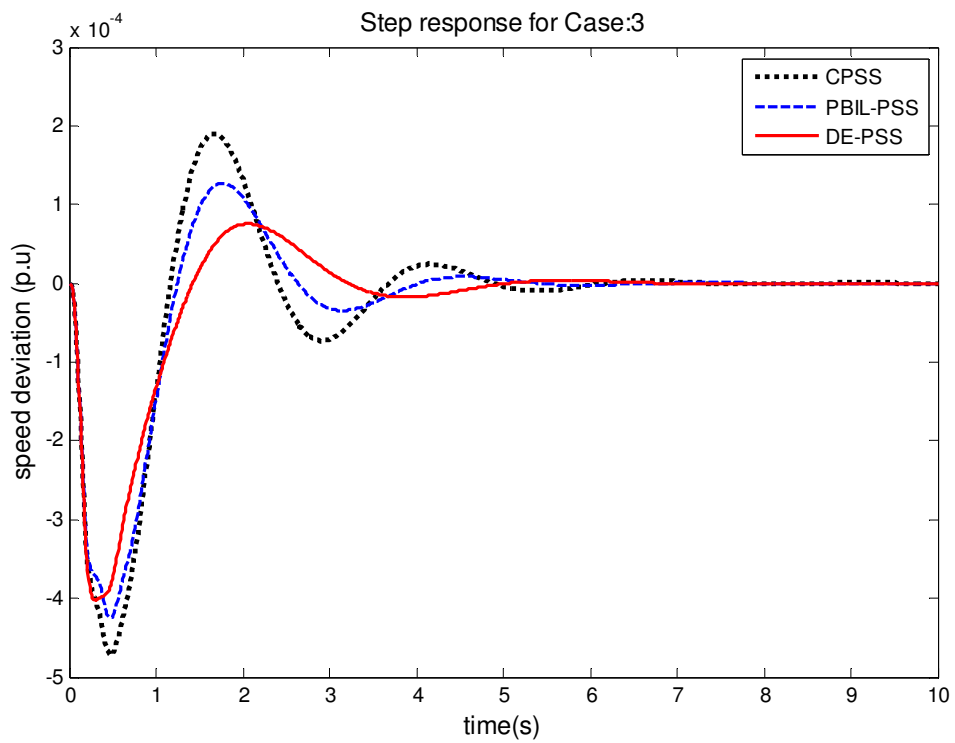


Figure 6.5: Rotor speed responses for Case 3

Figure 6.6 shows the step responses for case 4. Both DE-PSS and PBIL-PSS exhibit better performances than the CPSS. However, DE-PSS has once more outperformed both PBIL-PSS and CPSS with a settling time of 4.5 seconds. Whereas CPSS only settles after 6 seconds with higher overshoots and undershoots of 2.0×10^{-4} pu and -4.9×10^{-4} pu, respectively. PBIL-PSS settles after 5 seconds.

Figure 6.7 shows the speed responses of the system for cases 5. In this particular case, the system is heavily loaded with weak tie-line. In this scenario, the CPSS has a faster response but oscillatory, hence a longer settling time of about 8 seconds. Whereas DE-PSS has a slower response but exhibits the lowest overshoots and undershoots with a settling time of about 5 seconds. Similarly, PBIL-PSS responds reasonably fast with more oscillation than the DE and higher overshoots. PBIL-PSS settles in about 5.5 seconds.

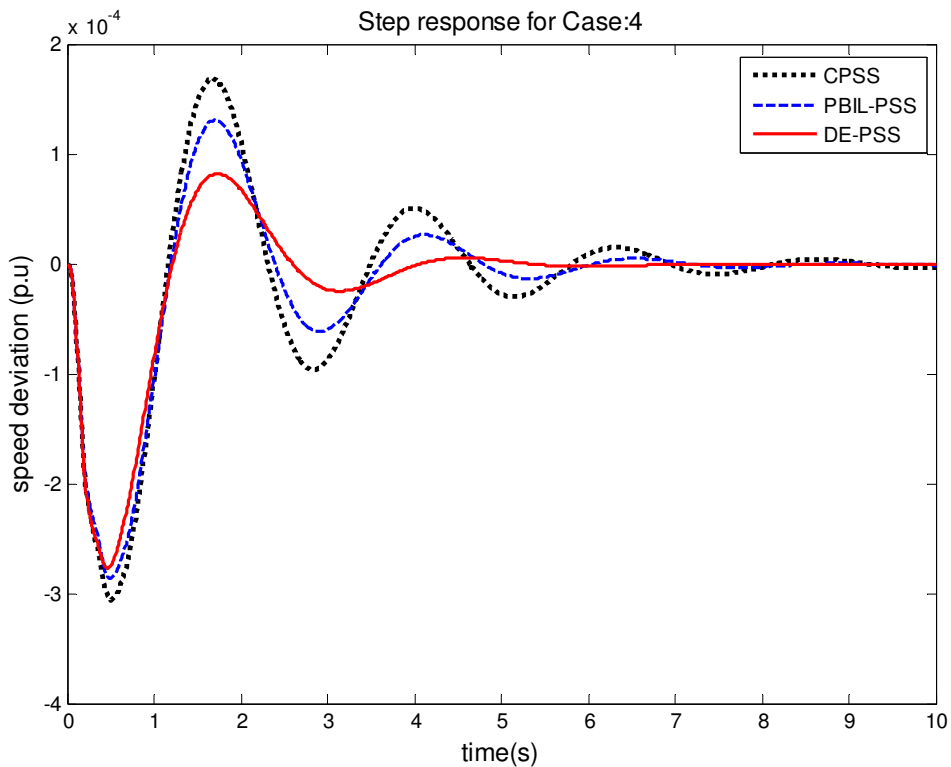


Figure 6.6: Speed response for Case 4

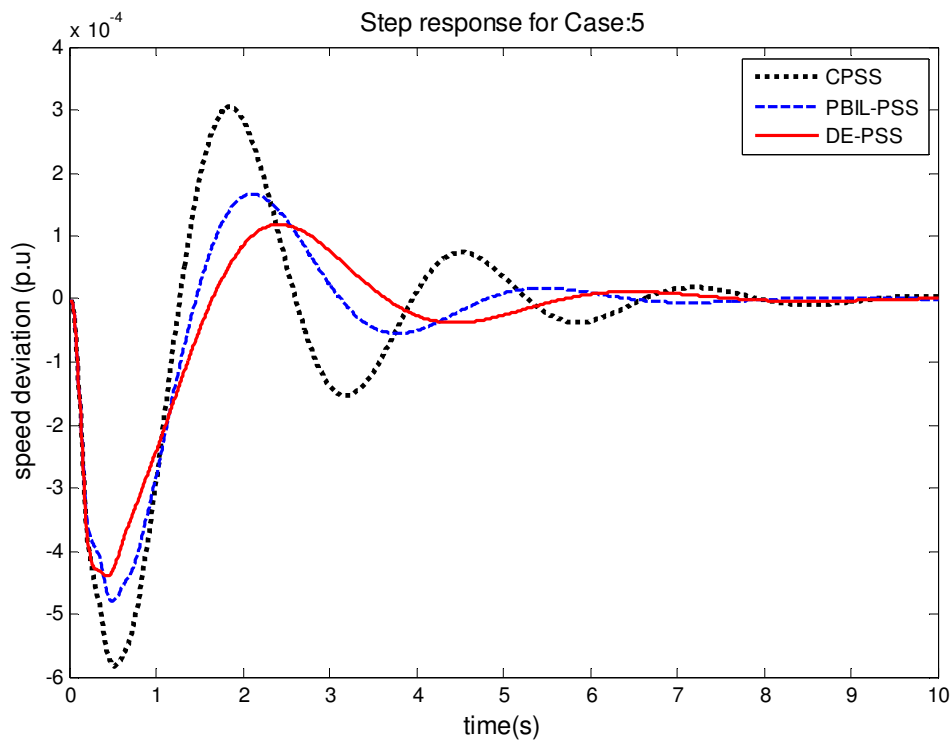


Figure 6.7: Speed responses for Case 5

6.4.2 Large disturbance simulation

For the large disturbance simulation, also known as transient stability, a more severe fault is considered to further evaluate the PSSs robustness. The fault consists of subjecting the system to a 5-cycles 3-phase fault. The fault is applied after 1 second on line 1 (Figure 5.6). The fault was cleared by disconnecting the line.

Transient Stability was performed for Case 1 to 4. Case 5 was not considered for these simulations because of the inability of the system to regain stability after the fault was cleared. This is mainly due to high value of the rotor pre-fault angle.

6.4.2.1 Case 1: nominal operating condition ($P=1.0\text{pu}$; $Q= 0.32\text{pu}$; $X_e=0.5\text{pu}$)

Figure 6.8 to Figure 6.12 show the responses of the rotor speed, rotor angle, voltage terminal, field voltage and active power following a three – phase short circuit fault on line 1. Observations of the system responses show that all PSSs perform adequately. From Fig, 6.8, it can be seen that the rotor angle increases when the fault is applied and settles to a new value when the fault was cleared. The system equipped with DE-PSS

displays a slightly better performance in terms of rotor angle and PBIL-PSS shows a slight overshoot whereas the CPSS response is slightly slower to settle down. The pre-fault angle was about 22.5° and increased to about 30° at post-fault. Similar trends are observed for the rotor speed responses (see Fig. 6.9) whereby the speed increases during the fault and returns to initial speed when the fault is cleared. DE-PSS settles slightly faster whilst PBIL-PSS has a slightly higher undershoot. However, the CPSS has the least undershoot. Figure 6.10 shows for the responses of the terminal voltage. The voltage dips to 0.8pu during the fault and returns to the initial value of 1.05p.u after the fault was cleared. All controllers perform similarly with slight undershoot and overshoot. The electric field responses are shown in Figure 6.11. It can be seen that it first increases to the upper limit of voltage regulator, then decreases to lower limit before returning to steady state when the fault is cleared. All the PSSs have similar responses; however, DE-PSS has a higher overshoot. This may be attributed to the high gain of DEPSS in comparison to PBIL-PSS and CPSS. Figure 6.12 shows the output power response of all PSSs. It can be observed that all controllers perform similarly.

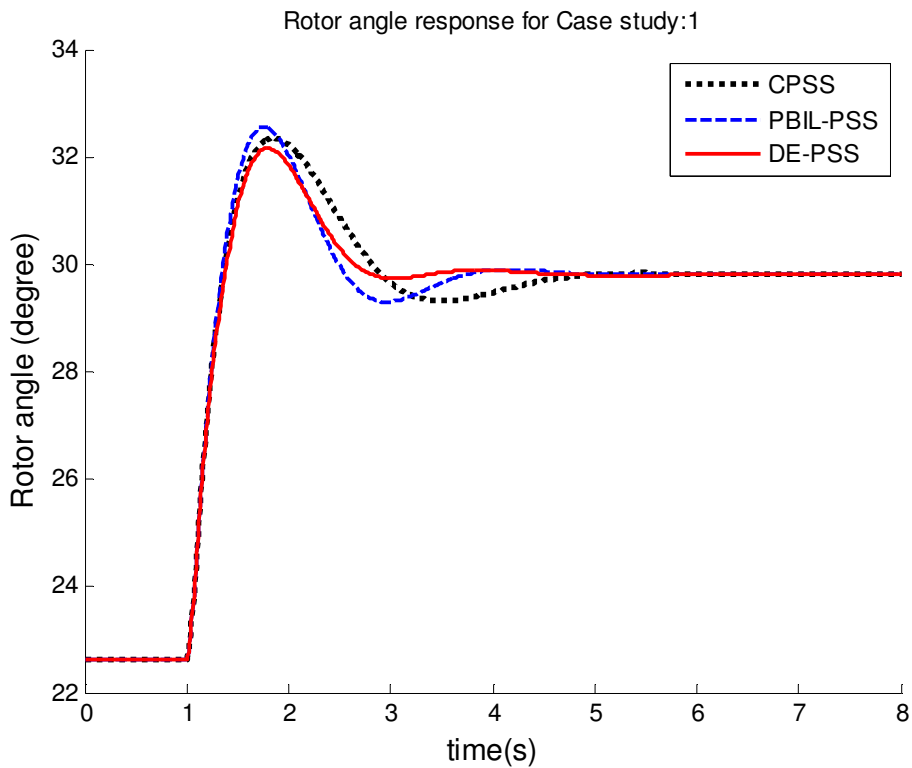


Figure 6.8: Rotor angle responses following 3-phase fault for case 1

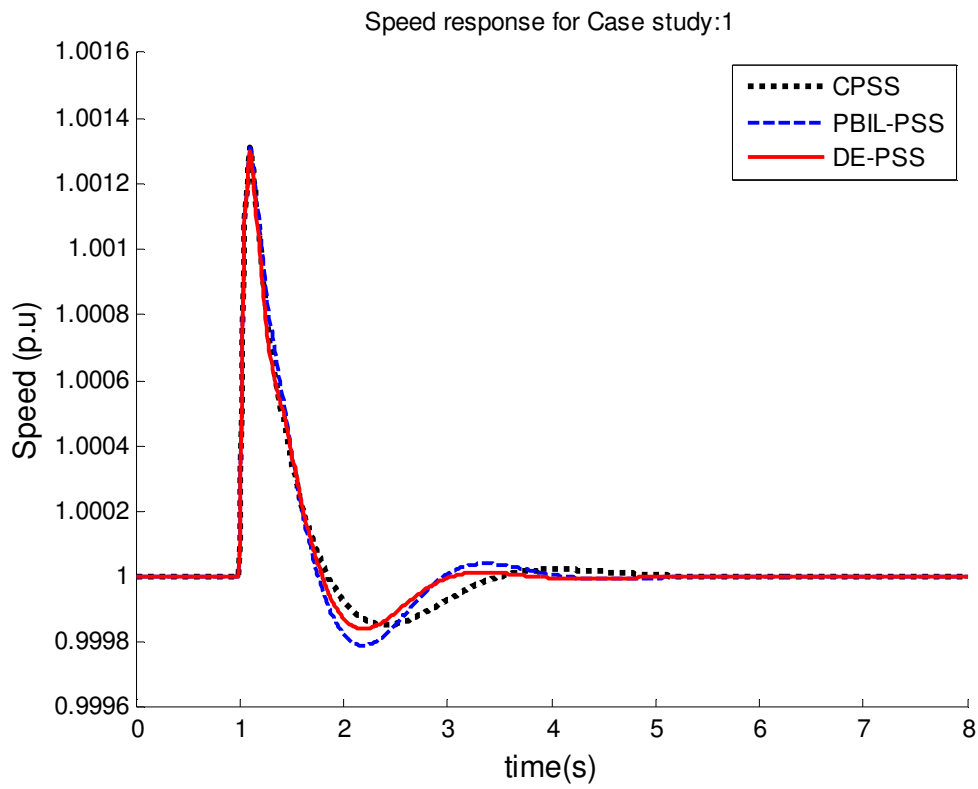


Figure 6.9: Rotor speed responses following 3-phase fault for case 1

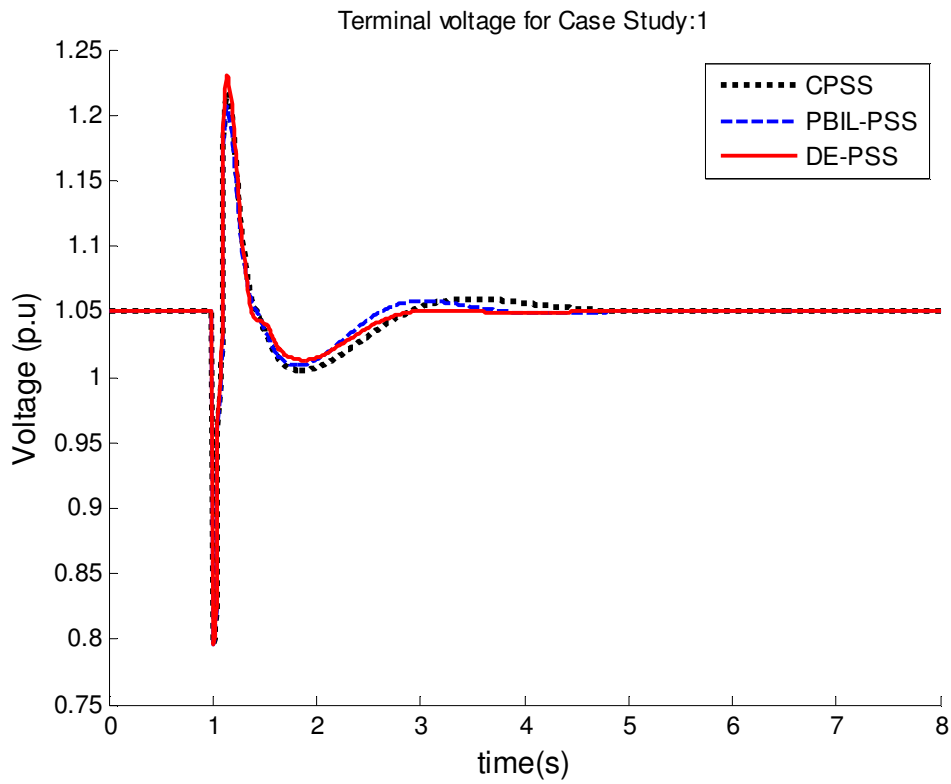


Figure 6.10: Terminal Voltage following 3-phase fault for case 1

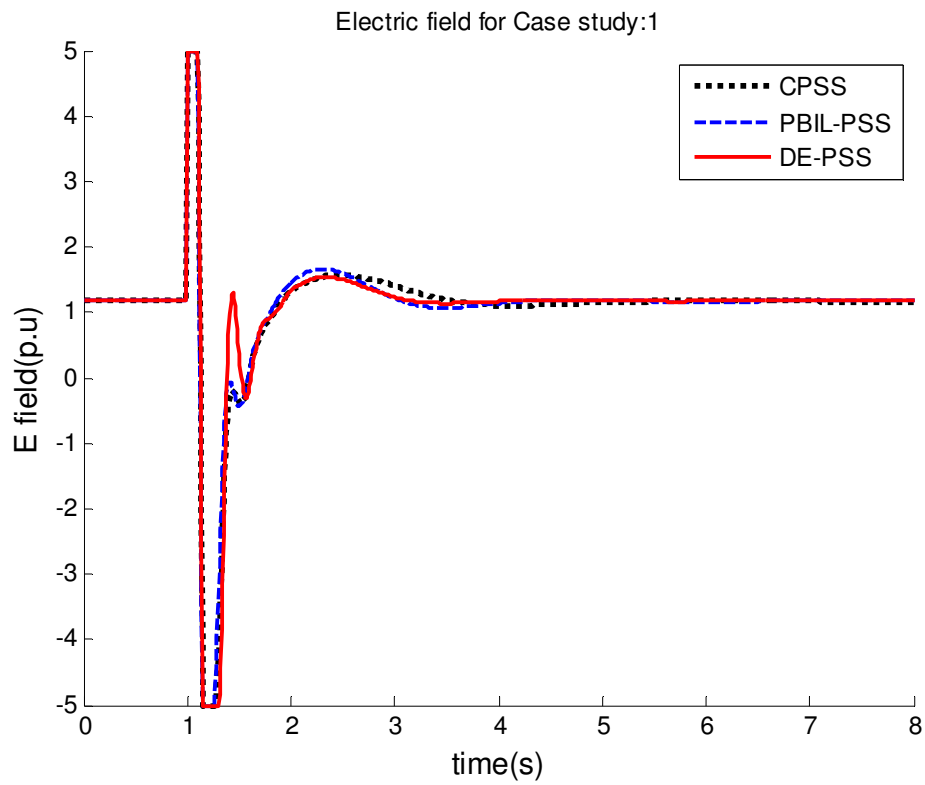


Figure 6.11: Electric field voltage following 3-phase fault for case 1

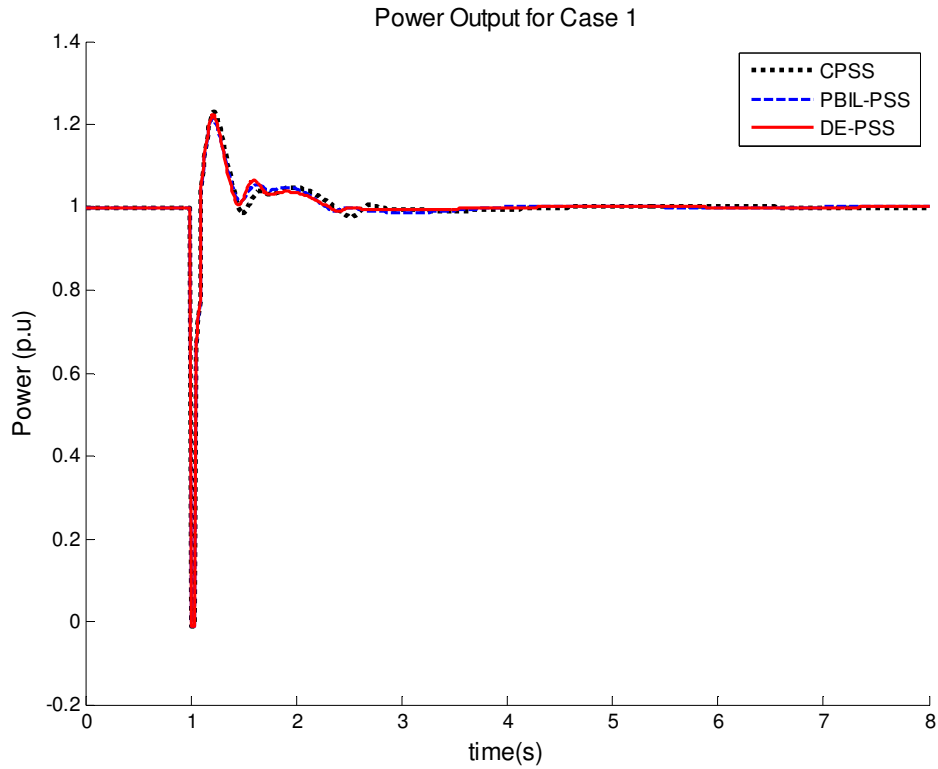


Figure 6.12: field voltage following 3-phase fault for case 1

6.4.2.2 Case 2: System operating at P=0.5pu; Q= 0.17pu; Xe=0.7pu

Figure 6.13 to Figure 6.17 illustrate the rotor angle, the rotor speed, the terminal voltage at bus 1, the electric field voltage and the power output response, respectively, following a 3-phase fault on line 1. The different PSS responses followed a similar trend to Case 1. Figure 6.13 shows the initial angle was 25.5° before fault and 36° at post-fault. Moreover, DE-PSS settles within 3.5 seconds with the least overshoot and undershoot, faster than its counterparts, PBIL-PSS and CPSS which settle in 4 and 4.5 seconds, respectively. In Figure 6.14, the speed response yields similar settling times to those of the rotor angle. The terminal voltage is shown in Figure 6.15. It can be observed that DE-PSS exhibits slightly higher overshoot whilst PBIL-PSS has the lowest overshoot. However DE-PSS settles slightly faster (in 3.5 seconds) than PBIL-PSS and CPSS, which settled in 4 and 5 seconds, respectively. Figure 6.16 shows the systems electric field responses during and after the fault. The difference in performance between all the different PSSs is minimal. For the power response shown in Figure 6.17, all PSSs perform similarly.

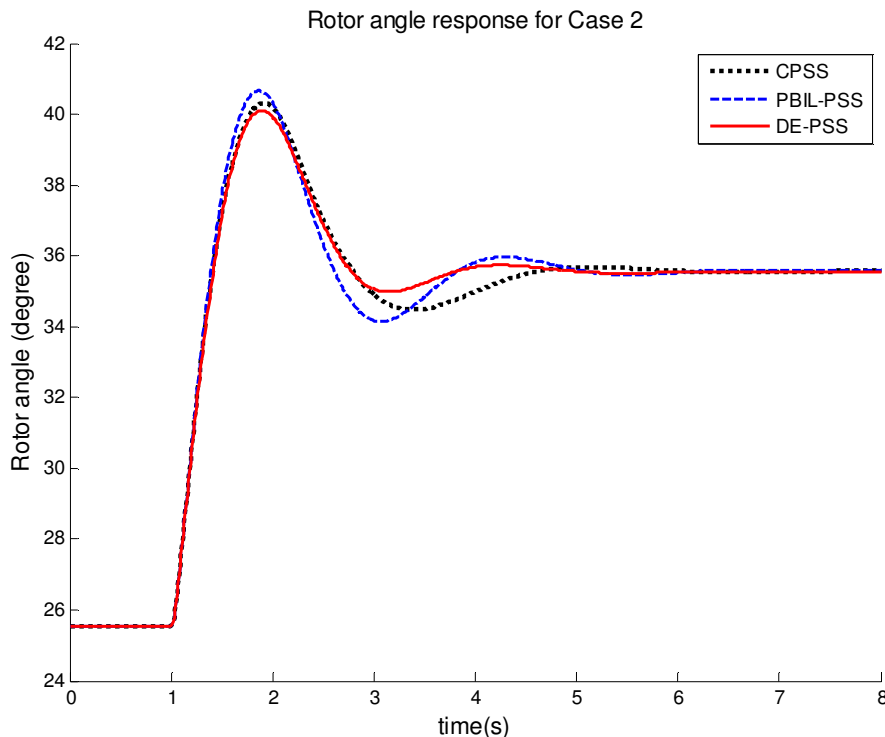


Figure 6.13: Rotor angle response following 3-phase fault for Case

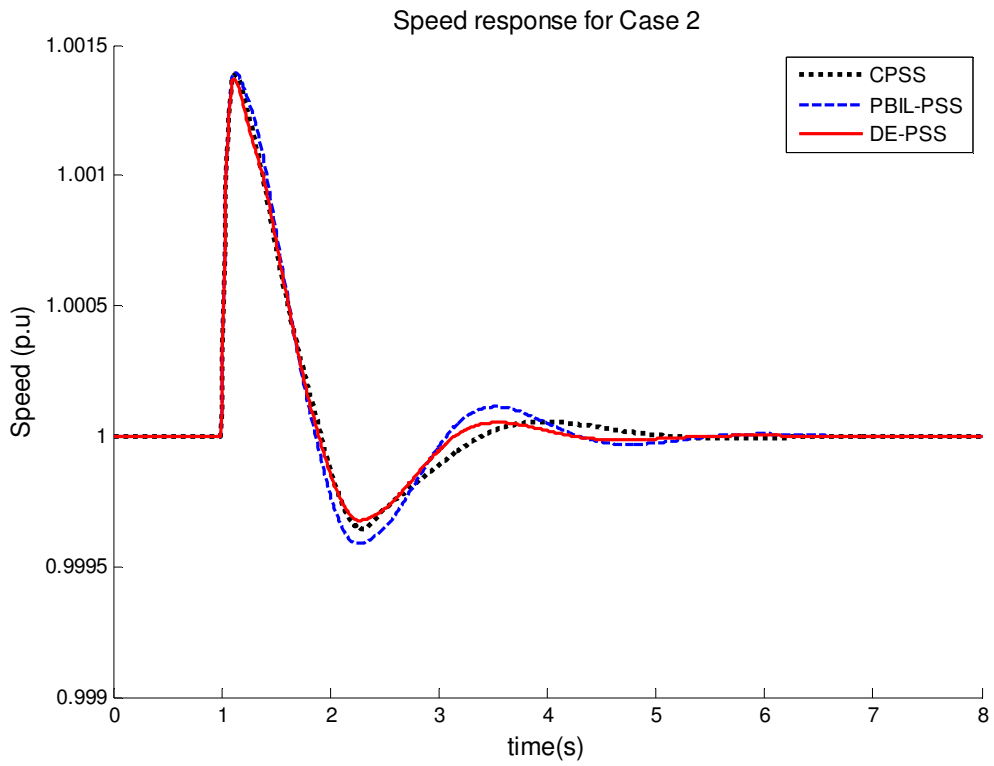


Figure 6.14: Speed response following a 3-phase fault for Case 2

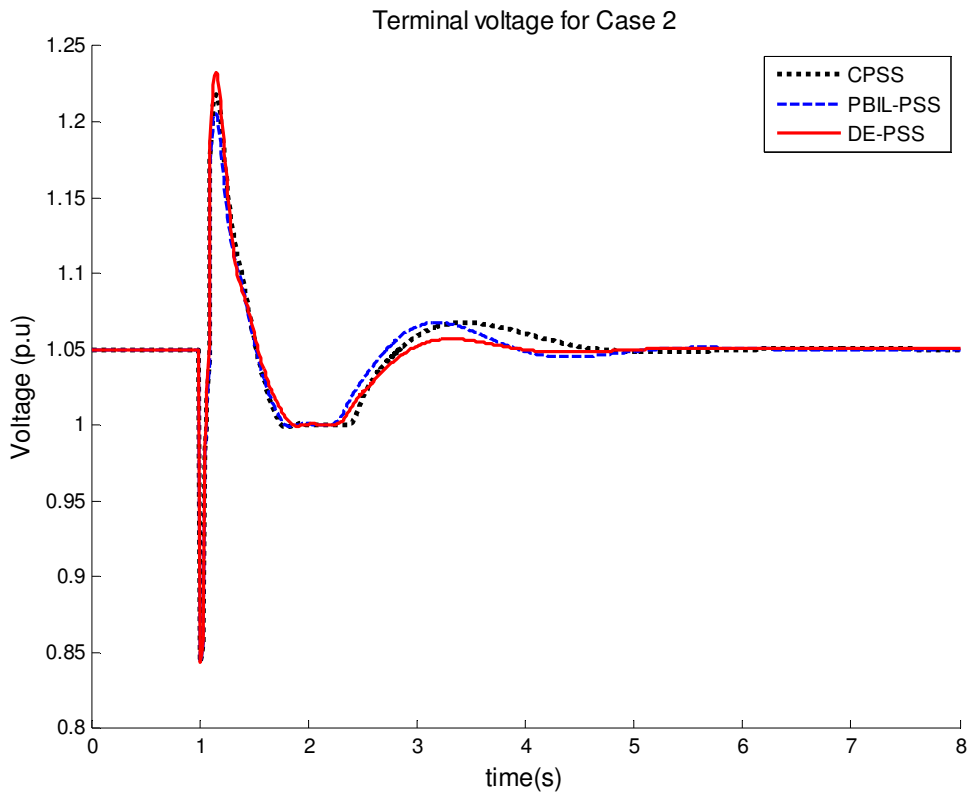


Figure 6.15: Terminal voltage response following 3-phase fault for Case 2

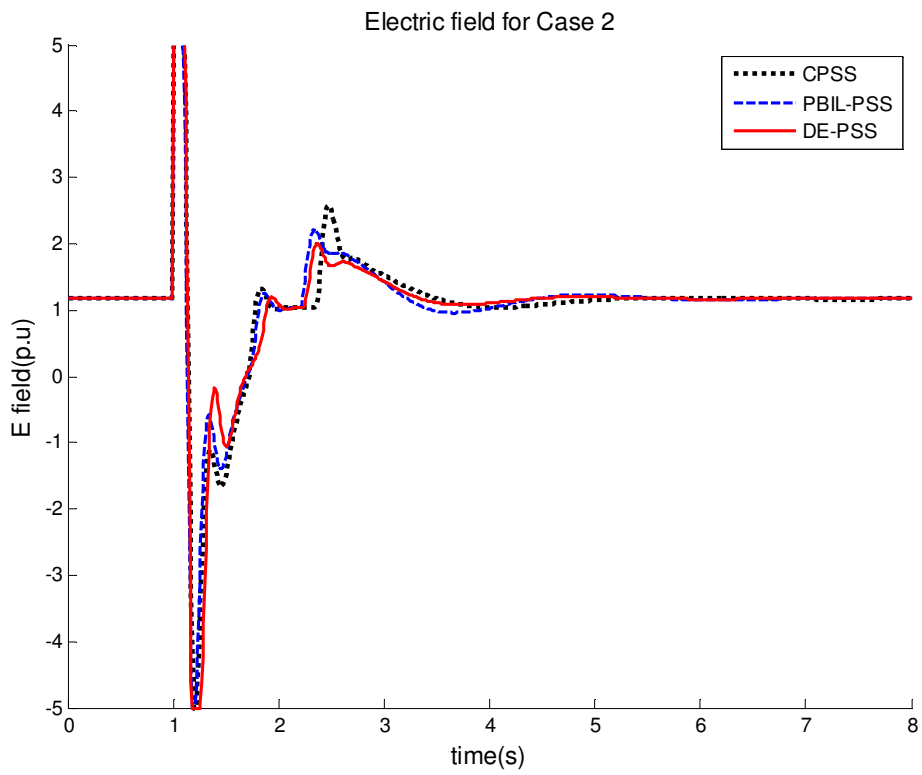


Figure 6.16: Electric field response for Case 2

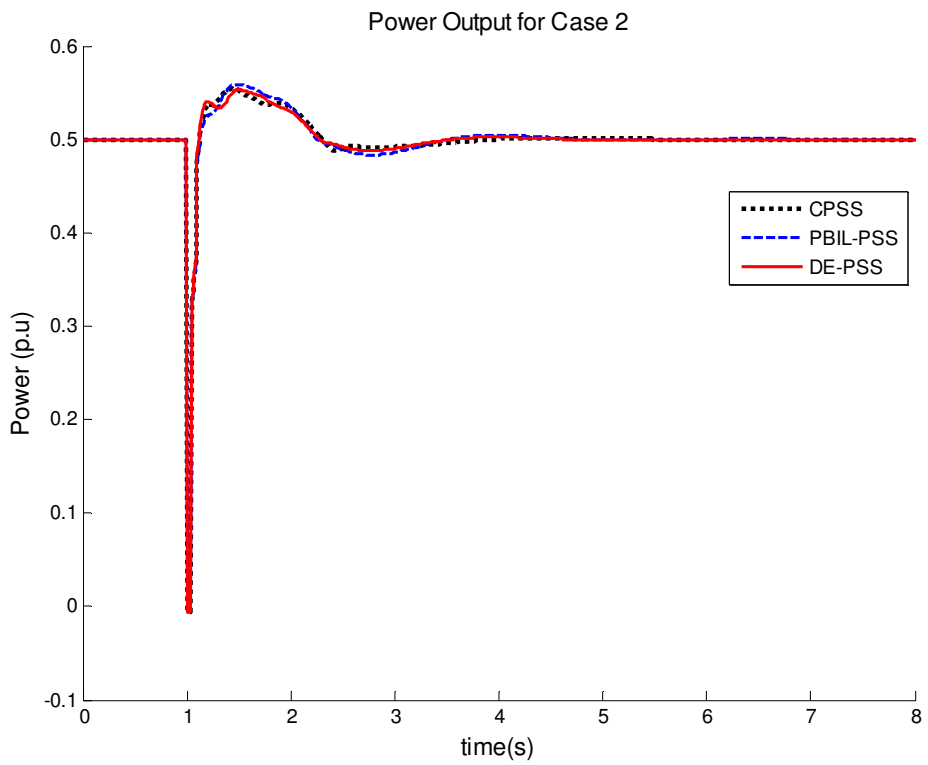


Figure 6.17: Power output for Case 2

6.4.2.3 Case 3: System condition, $P=1.0\text{pu}$; $Q=0.34\text{pu}$; $X_e=0.7\text{pu}$

In this case, the fault was applied when the system was heavily loaded with a relatively weak tie-line. Figure 6.18 to Figure 6.20 show that the performance of the CPSS has deteriorated. It has more oscillations and a settling time of over 8 seconds for the rotor angle, the rotor speed and the terminal voltage whereas DE and PBIL subside in about 5 and 6 seconds, respectively. Furthermore, DE displays the least overshoots and undershoots. For the electric field, both DE and PBIL have similar performance as opposed to the CPSS which is slightly slower in settling time. As for the active power output in Figure 6.22, all PSSs show similar performances.

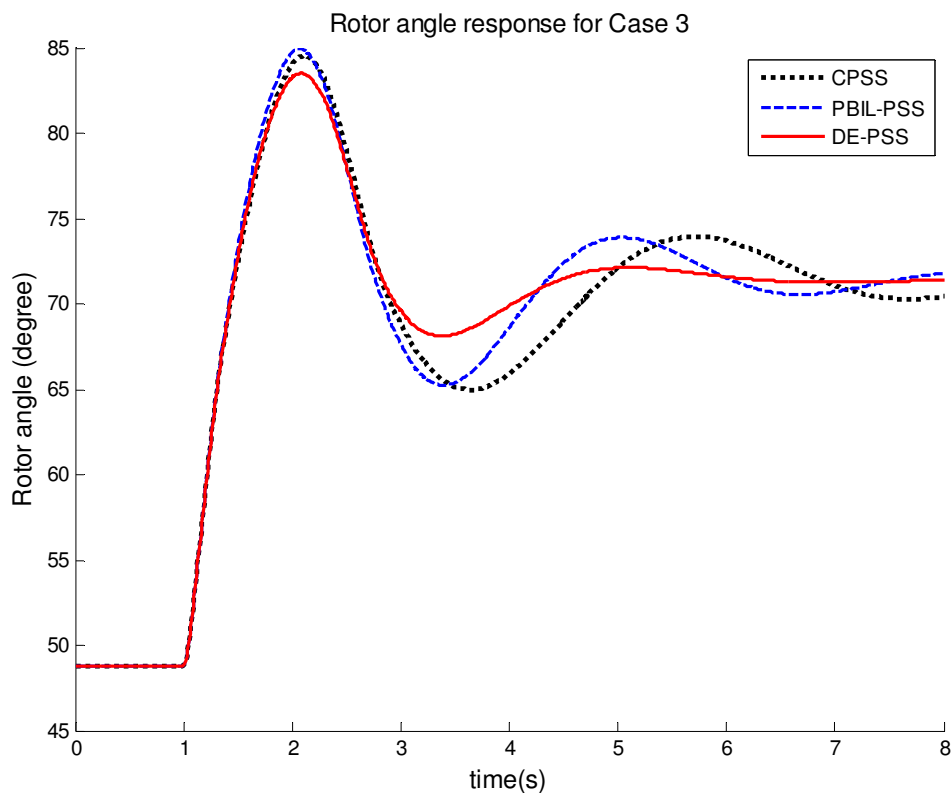


Figure 6.18: Rotor angle responses for Case 3

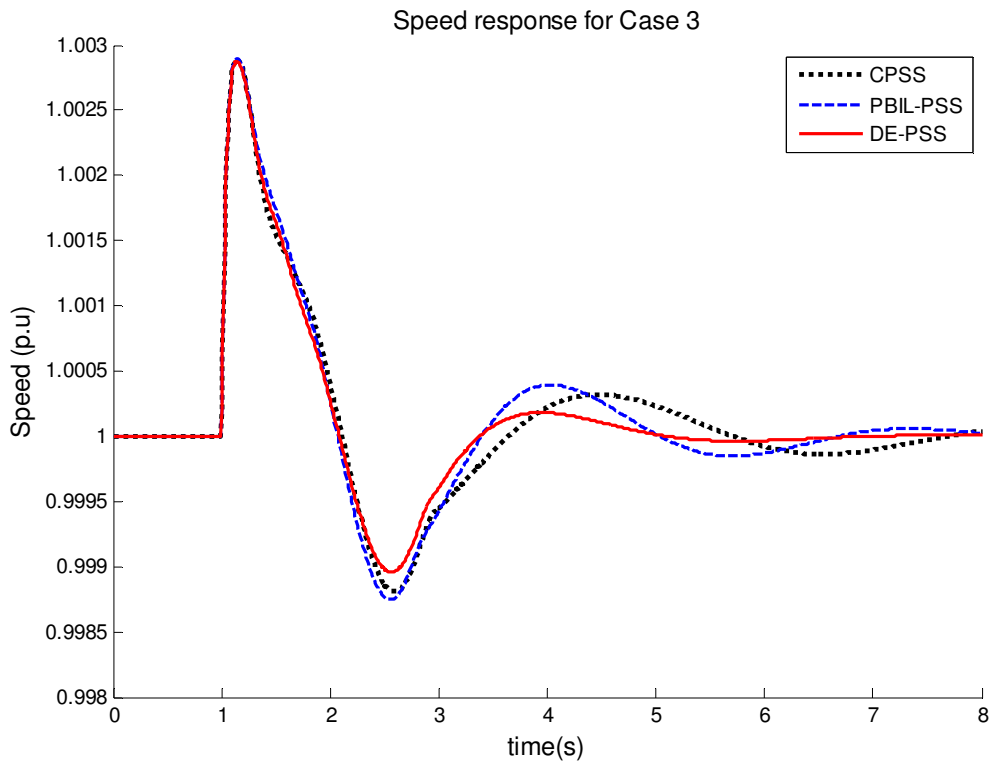


Figure 6.19: Rotor speed responses for Case 3

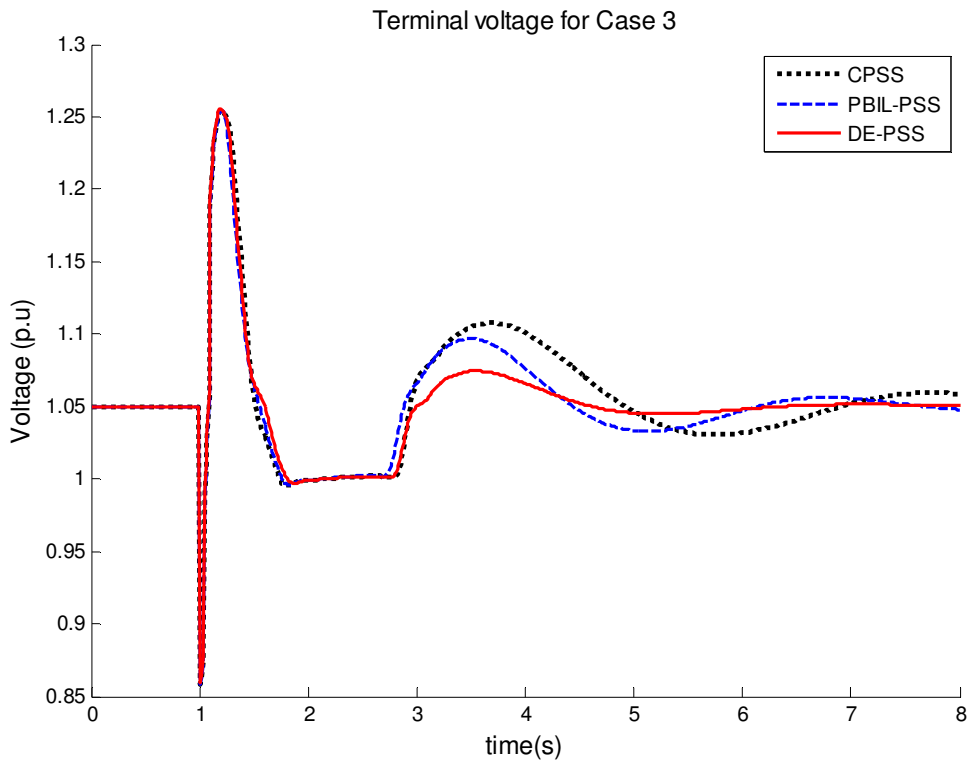


Figure 6.20: Terminal voltage responses at bus 1 for Case 3

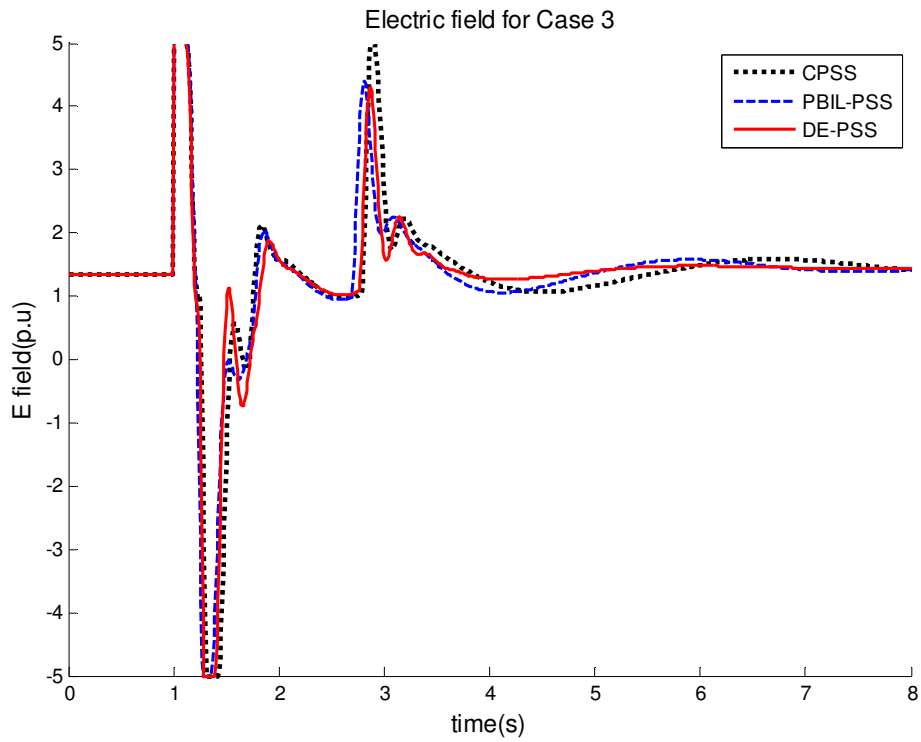


Figure 6.21: Electric field voltage for Case 3

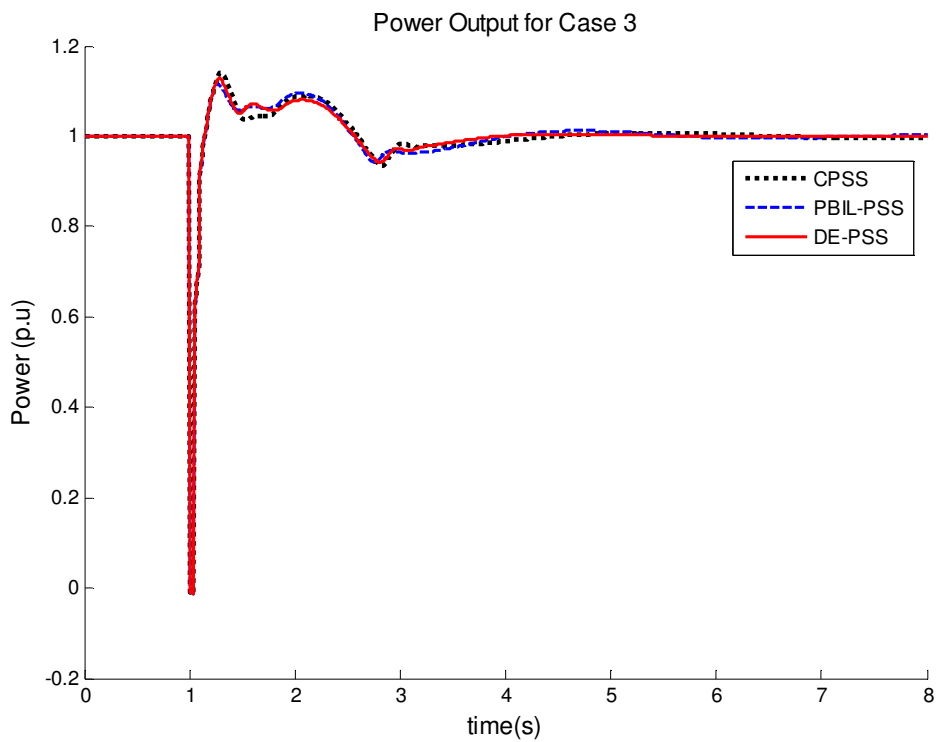


Figure 6.22: Active Power output responses for Case 3

6.4.2.4 Case 4: System condition, $P=0.5pu$; $Q=0.16pu$; $X_e=0.9pu$

Fig 6.23 to Figure 6.27 show the system responses of the rotor angle, the rotor speed, the terminal voltage, the field voltage and the active power, respectively. Under this large disturbance, DE-PSS slightly outperforms both PBIL-PSS and CPSS with less overshoots and undershoots with a settling time of about 4.5 seconds for the rotor angle and rotor speed whereas PBIL-PSS shows slightly high overshoots and undershoots for the rotor angle and speed with a faster response and settling time than that of CPSS (see Figures 6.23-24). However, for both the terminal voltage responses (Figure 6. 25) and electric field voltage (Figure 6.26), DE has higher overshoots and undershoots following the fault clearance but still regain the stability faster than both CPSS and PBIL.

Under the aforementioned operating condition, the CPSS showcase adequate performance with a settling time of around 4 seconds for the active power responses shown in Figure 6.27. All PSSs perform similarly with slight differences for the active power.

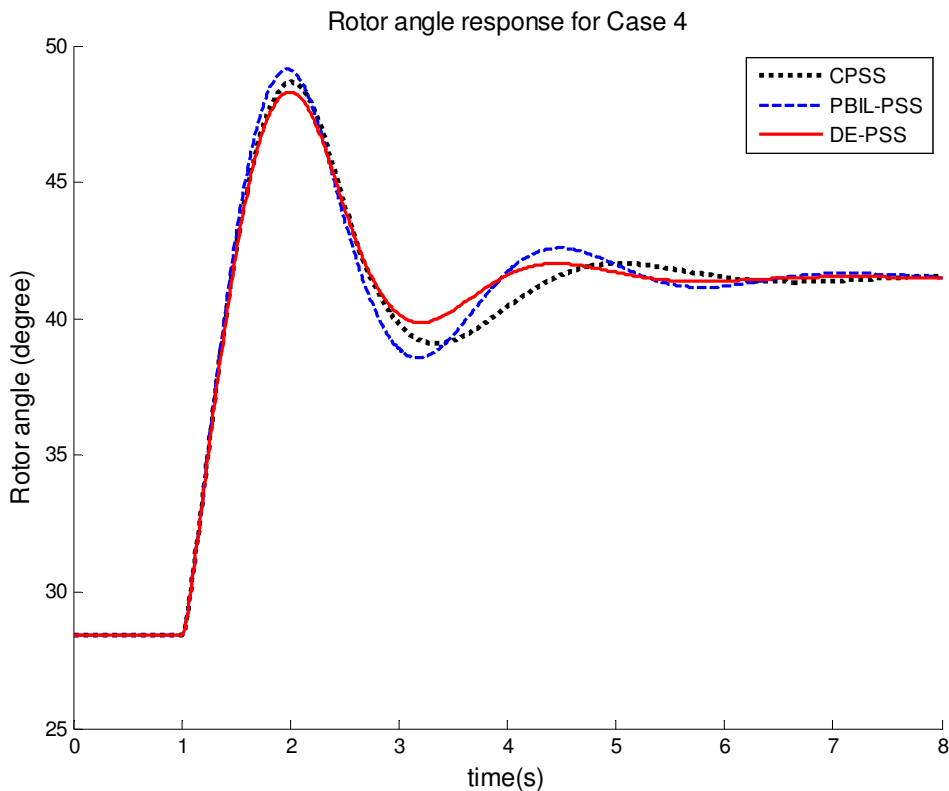


Figure 6.23: Rotor angle response following 3-phase fault for Case 4

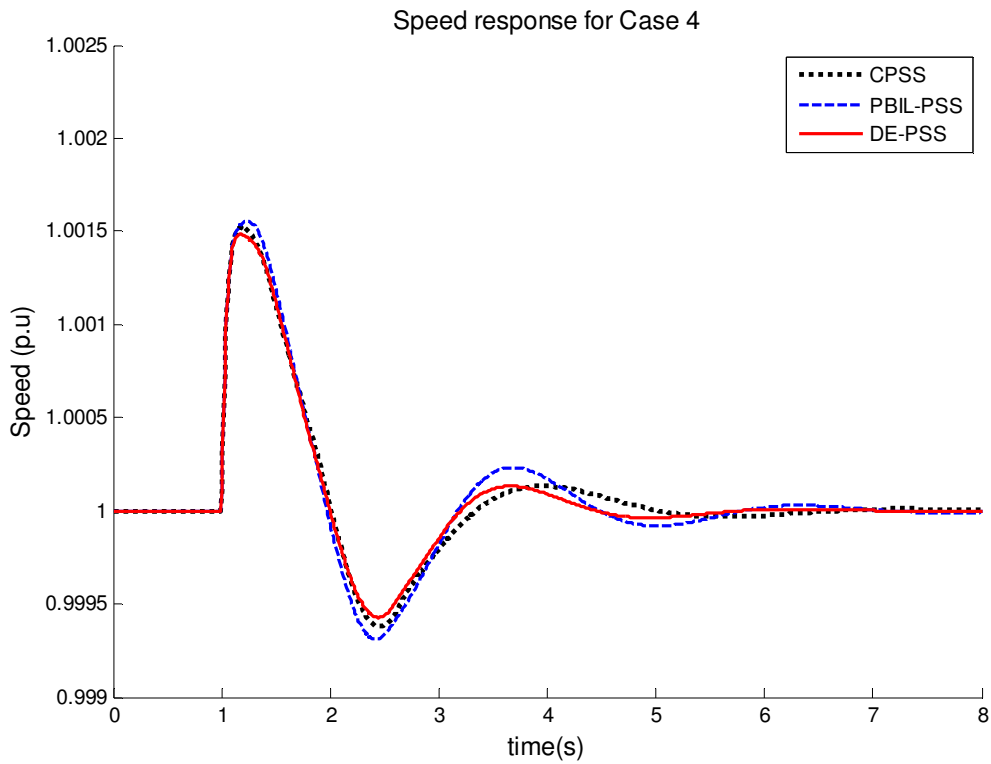


Figure 6.24: Rotor speed response following 3-phase fault for Case 4

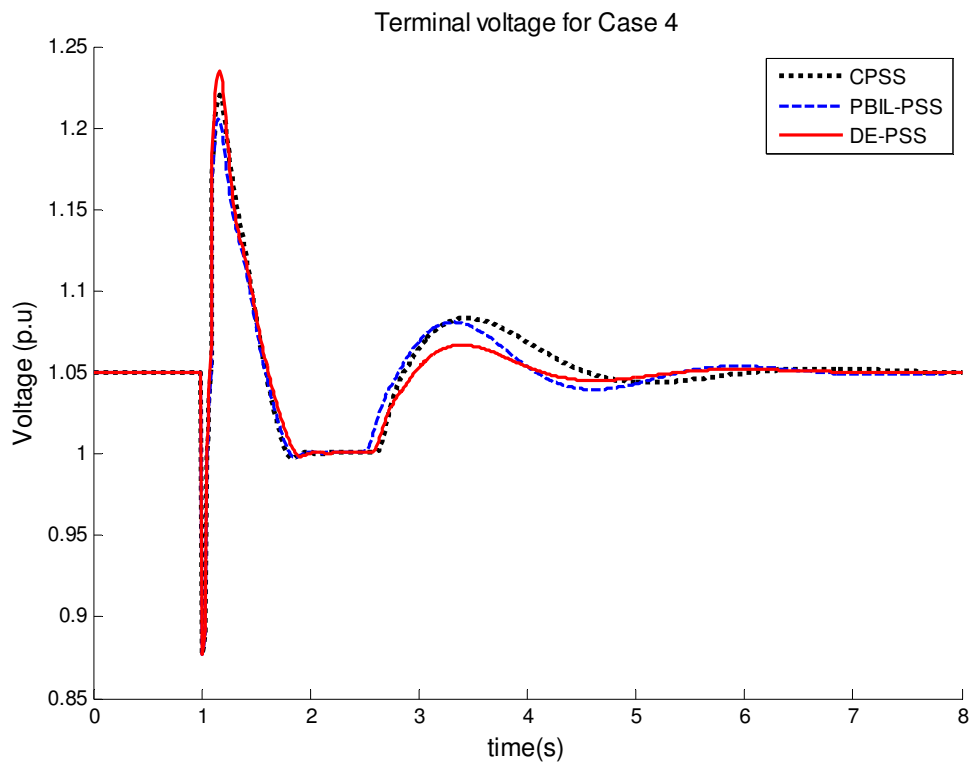


Figure 6.25: Terminal voltage response following 3-phase for Case 4

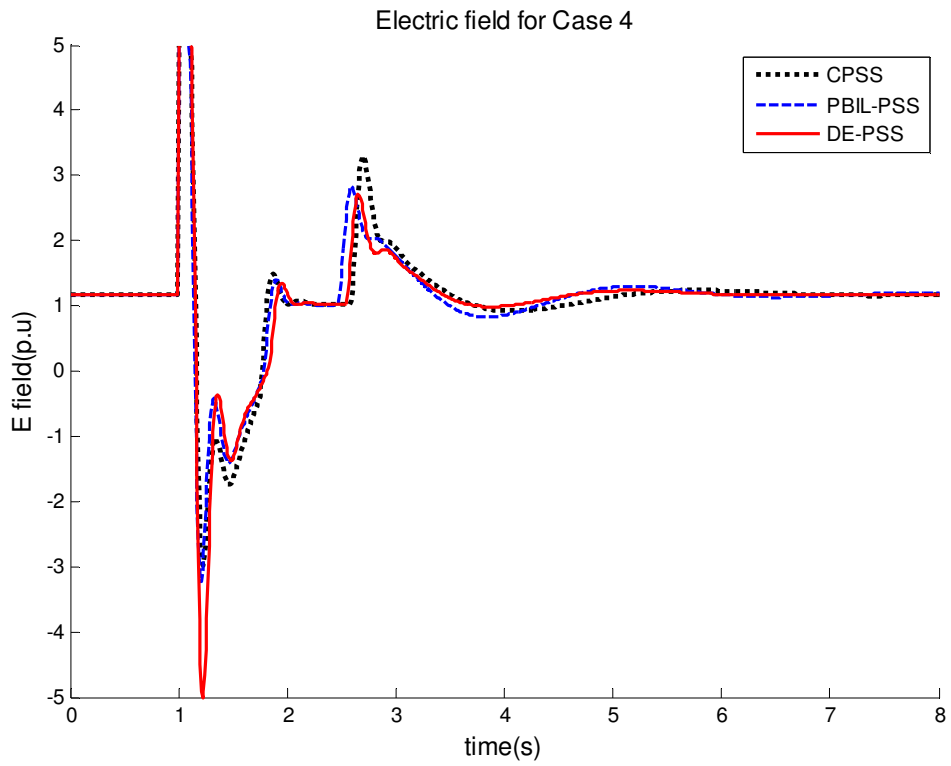


Figure 6.26: Electric field following a 3-phase fault for Case 4

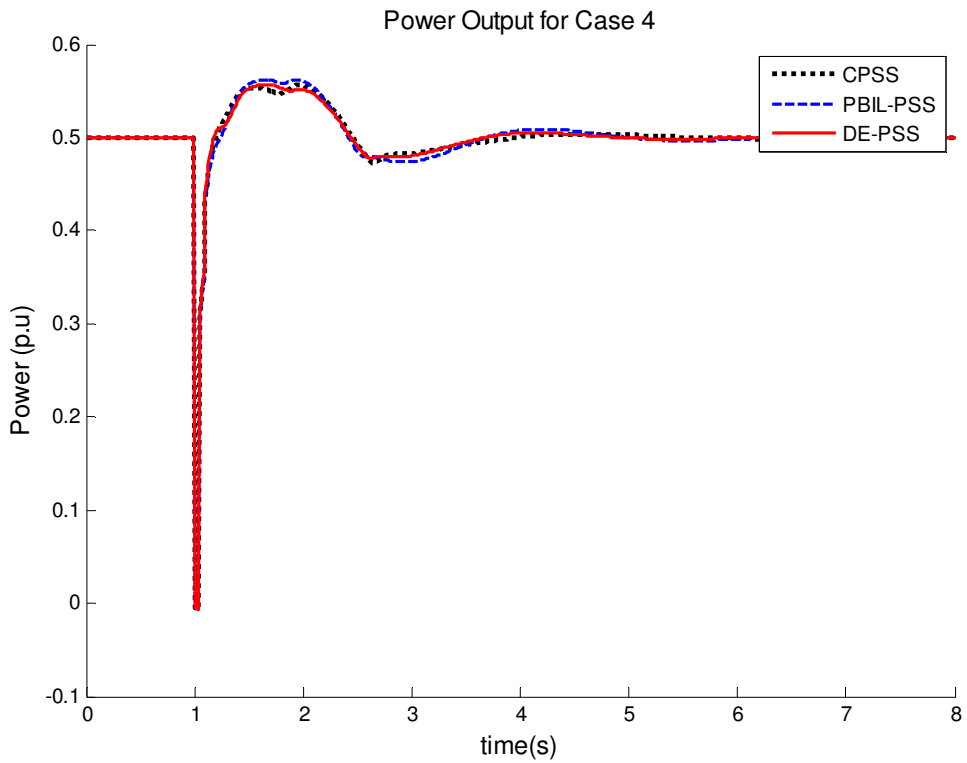


Figure 6.27: Active power output following 3-phase fault for Case 4

6.5 Summary

In this chapter, the application of two proposed Evolution Algorithms (DE and PBIL) for tuning the parameters of PSS has been tested for an SMIB system. The resulting PSSs were compared with those of the CPSS. The modal analysis shows that all the PSSs are capable of improving the dynamic stability of the system. In particular, DE-PSS performs better than PBIL-PSS and CPSS for all cases that have been discussed. These results have been validated in time domain simulations where DE-PSS outperformed both PBIL-PSS and CPSS with faster settling times and less overshoots and undershoots in some cases. As expected, the CPSS performs well for the nominal condition however degrades when the operating conditions change for both small and large disturbances.

Chapter 7

Simulation Results for the Two-Area Multi-machine Systems

7.1 Introduction

This chapter presents the simulation results for the Two-Area Multi Machine system. The performances of the PSSs are assessed via the modal analysis and validated using time domain simulations. The effectiveness of the resulting PSSs to damp the low frequency oscillations is tested under various operating conditions (case 1 to case 6) using small and large disturbances. The comparison is carried out between the system equipped with CPSS, DE-PSS and PBIL-PSS. First, the final value obtained from the optimization process of both DE and PBIL are presented. Next, the performances of PSSs are evaluated and compared with modal analysis.

7.2 PSS Parameters optimization

Following the application of DE and PBIL to tune the PSS, Figure 7.1 and 7.2 show the optimization fitness curve of both algorithms. The optimization consisted of finding a set of parameters that would give the best damping for the dominant poles of the system. Figure 7.1 shows DE converging to the optimum damping ratio value of 0.2263 corresponding to the electromechanical damping ratio of Case 6. Whereas PBIL, as illustrated in Figure 7.2, reached the optimum value of damping ratio 0.2095. These results indicate that DE performs slightly better than PBIL in finding the optimum value.

It is deduced from Figure 7.1 that DE is able to explore the search space to only converge after 100 generations for a maximum of 180 generations. PBIL, on the other hand, explores the search space in a parallelized manner and starts settling after 200 for a maximum of 500 generations. The difference in the maximum number of generations is to allow PBIL to settle since this approach is based on learning which takes some time.

However, the speed of convergence can be improved by changing the learning rate which may affect the ability of the algorithm to find the optimum value.

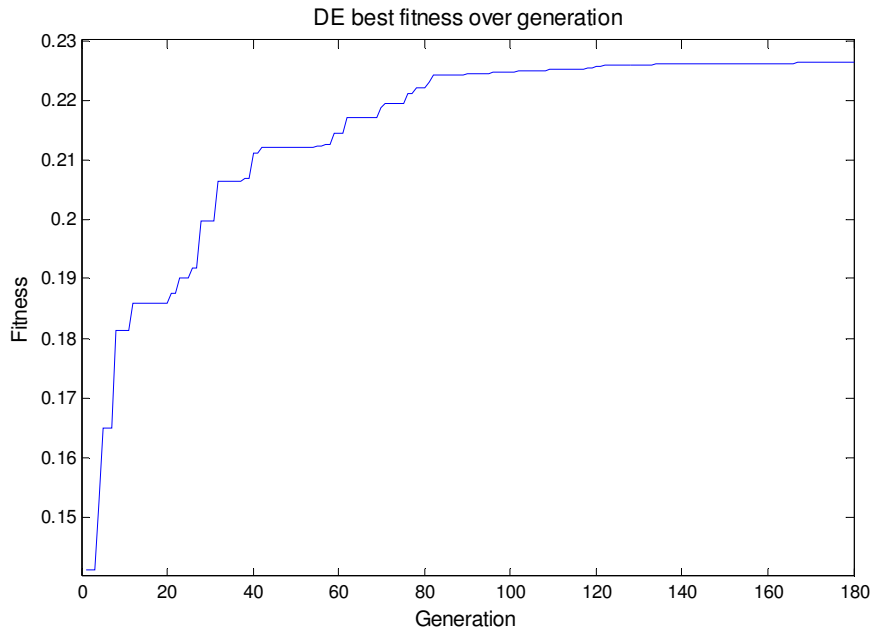


Figure 7.1: DE fitness curve

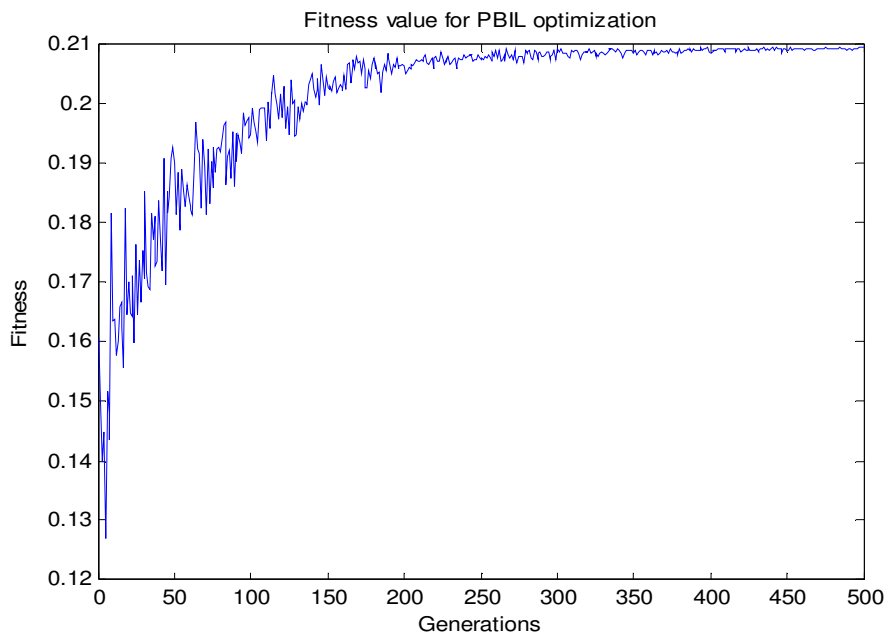


Figure 7.2: PBIL fitness curve

Table 7.1 shows the PSSs' parameters obtained after optimization along with the CPSS' parameters. The system is constituted of four identical generators in two-different areas as discussed in section 5 [3]. Hence, the PSSs in each area are identical such that generator 1 and 2 have similar PSSs. The washout time constant used in the Multi-machine system is 10seconds.

Table 7.1: Two-area PSS parameters

		K_p	T_1	T_2	T_3	T_4
PBIL-PSS	<i>Gen 1&2</i>	19.191	0.238	0.012	0.049	0.014
	<i>Gen 3&4</i>	16.633	0.119	0.083	0.055	0.010
DE-PSS	<i>Gen 1&2</i>	19.992	0.051	0.019	0.053	0.015
	<i>Gen 3&4</i>	19.997	0.116	0.015	0.111	0.016
CPSS	<i>Gen 1&2</i>	9.877	0.307	0.015	0.051	0.013
	<i>Gen 3&4</i>	13.685	0.126	0.085	0.062	0.010

7.3 Modal Analysis

Table 7.2 shows the eigenvalues corresponding to the inter-area mode of the closed loop and the open loop system and the respective damping ratios in brackets. As expected, the implementation of all PSS has significantly improved the system's stability. In Case 1, the nominal condition, DE has a slightly higher damping ratio of 27.07% whereas PBIL has a damping ratio of 23.62% and CPSS a damping ratio of 17.61%. Hence, the damping provided by DE-PSS is 3.45% higher than that of PBIL and 9.46% better than that CPSS. PBIL on the other hand is 6.42% higher than CPSS. From the analysis of the eigenvalues, it can be deduced that DE-PSS has transient that decay to within the ± 2 % band in less than 3.2 seconds whilst PBIL-PSS and CPSS have transient that decay in 3.5 and 5 seconds, respectively. In Case 2, as the system loading is increased to 200MW, the damping ratio provided by all PSSs slightly drops. DE-PSS' damping ratio is reduced to a value of 26.12% whilst PBIL and CPSS are reduced to 22.65% and 16.56%, respectively. Hence, from the real part of the eigenvalue corresponding to the electromechanical mode, it is deduced that there are transient that decay to within the ± 2 % band in less than 3.3, 4 and 5.3 seconds for DE, PBIL and CPSS, respectively. As the power transmitted is

gradually increased to 300MW, 400MW and 500MW for Case 3, 4 and 5, the damping is further reduced for all PSSs. DE-PSS provides damping ratio of 24.92%, 23.84% and 22.63%, with transient that decay in 3.5 to 4 seconds. Whereas PBIL-PSS provides damping ratios of 22.45%, 22.18% and 20.95%, respectively with a settling time of 4 to 6 seconds. CPSS, as expected, provides the lowest damping of 16.42%, 16.23% and 15.06% for case 3, 4 and 5, respectively. Similar trends are observed in Case 6 whereby the PSSs' damping ratios are slightly reduced to 22.59% for DE-PSS, 20.94% for PBIL-PSS and 15.02% for the CPSS. The modal analysis also reveals that CPSS takes longer to subside low frequency oscillations. This is characterized by eigenvalues corresponding to the inter-area mode being closest to the imaginary axis on the s-plan. DE-PSS has the fastest settling time, subsequently the furthest eigenvalues in the left-hand side of the s-plan. PBIL-PSS displays better performance than that of the CPSS and slightly less than that of DE-PSS.

Table 7.2: Inter-area modes for Two-Area Multi-machine system

Case	DE-PSS	PBIL-PSS	CPSS	No-PSS
1	-1.273±j4.528 (0.2707)	-1.115±j4.589 (0.2362)	-0.807±j4.509 (0.1761)	-0.006 ±j4.962 (0.0012)
2	-1.219±j4.507 (0.2612)	-1.062±j4.566 (0.2265)	-0.753±j4.484 (0.1656)	0.0461 ±j4.937 (-0.0093)
3	-1.154±j4.477 (0.2492)	-1.044±j4.534 (0.2245)	-0.741±j4.453 (0.1642)	0.0428 ±j4.894 (-0.0087)
4	-1.089±j4.436 (0.2384)	-1.022±j4.491 (0.2218)	-0.725±j4.412 (0.1623)	0.0806 ±j4.792 (-0.0165)
5	-1.015±j4.404 (0.2263)	-0.950±j4.42 (0.2098)	-0.648±j4.384 (0.1506)	0.0771±j4.697 (-0.0168)
6	-0.806±j3.477 (0.2259)	-0.748±j3.496 (0.2094)	-0.518±j4.411 (0.1502)	0.0589 ±j3.419 (-0.0170)

Table 7.3 shows the local modes for Area 1. The addition of PSSs improved the mode's damping ratio from poorly damped to adequately damped. The CPSS provides 37.23%, 35.13%, 34.72%, 34.52%, 33.28% and 30.95%, for Case 1 to 6, respectively,. The optimized PSS using PBIL provides 38.96% for Case 1, 37.46% for Case 2, 37.22% for Case 3, 36.85% for Case 4, 35.17% for Case 5 and 33.44% for Case 6 whereas DE provides 40.10% for Case 1, 38.62% for Case 2, 38.37% for Case 3, 38.01% for Case 4, 36.67% for Case 5 and 34.62% for Case 6. It can be observed that DE has a slightly higher damping than those of PBIL-PSS and CPSS.

Table 7.3: Local Area mode 1

Case	DE-PSS	PBIL-PSS	CPSS	No-PSS
1	-2.384±j5.446 (0.4010)	-2.372±j5.607 (0.3896)	-2.098±j5.229 (0.3723)	-0.606 ±j7.28 (0.0829)
2	-2.304±j5.503 (0.3862)	-2.288±j5.663 (0.3746)	-2.010±j5.308 (0.3513)	-0.4916 ±j7.290 (0.0735)
3	-2.291±j5.513 (0.3837)	-2.274±j5.673 (0.3721)	-1.997±j5.322 (0.3472)	-0.4955 ±j7.280 (0.0707)
4	-2.271±j5.525 (0.3801)	-2.253±j5.685 (0.3685)	-1.978±j5.340 (0.3452)	-0.5178 ±j7.300 (0.0678)
5	-2.195±j5.570 (0.3667)	-2.160±j5.709 (0.3517)	-1.907±j5.404 (0.3328)	-0.5357 ±j7.269 (0.0673)
6	-2.075±j5.623 (0.3462)	-2.049±j5.774 (0.3344)	-1.783±j5.477 (0.3095)	-0.4550±j7.33 (0.0617)

Table 7.4 shows the local modes for Area 2. Again, the addition of PSSs significantly improved the mode's damping ratio. The CPSS provides the lowest damping ratio of 67.23%, 65.41%, 63.95%, 62.09%, 58.82% and 53.96% for Case 1 to 6, respectively. PBIL-PSS, on the other hand, provides the highest damping of 77.25% for Case 1, 75.67% for Case 2, 74.34% for Case 3, 72.64% for Case 4, 68.21% for Case 5 and 65.53% for Case 6. DE- PSS provides slightly lower damping than those of PBIL with 74.76% for Case 1, 73.23% for Case 2, 71.94% for Case 3, 70.31% for Case 4, 67.66% for Case 5 and 63.39% for Case 6.

Table 7.4: Local Area mode 2

Case	DE-PSS	PBIL-PSS	CPSS	No-PSS
1	-5.966±j5.299 (0.7476)	-6.460±j5.310 (0.7725)	-5.191±j5.726 (0.6723)	-0.488 ±j7.42 (0.0656)
2	-5.812±j5.405 (0.7323)	-6.302±j5.444 (0.7567)	-5.047±j5.835 (0.6541)	-0.4990 ±j7.420 (0.0670)
3	-5.685±j5.488 (0.7194)	-6.170±j5.550 (0.7434)	-4.925±j5.920 (0.6395)	-0.5069 ±j7.405 (0.0683)
4	-5.526±j5.589 (0.7031)	-6.004±j5.679 (0.7264)	-4.771±j6.022 (0.6209)	-0.3721 ±j7.344 (0.0506)
5	-5.298±j5.765 (0.6766)	-5.601±j6.007 (0.6821)	-4.538±j6.239 (0.5882)	-0.3591 ±j7.335 (0.0488)
6	-4.901±j5.978 (0.6339)	-5.353±j6.16 (0.6553)	-4.130±j6.444 (0.5396)	-0.3875±j7.224 (0.0526)

7.4 Time domain Simulations

The time domain simulation is divided into two sections namely small and large disturbances.

7.4.1 Small disturbance

In this section, the time domain simulations were performed to validate the results of modal analysis. The PSSs are assessed by their ability to damp low frequency oscillations following a 10% step change in the reference voltage of the generator that has the largest participation on the inter-area mode, which is generator 2. The disturbance is performed for all 6 operating conditions. The step responses of active power deviation on all generators are presented in Figure 7.3 to Figure 7.26.

Figure 7.3 to Figure 7.8 illustrate the active power deviation response of generator 1 to the a 10% step change in reference voltage applied on generator 2 for Cases 1 to Case 6, respectively. The results show that DE-PSS has the least undershoot in all cases and a settling time varying between 3 and 4 seconds. Similarly to DE, PBIL-PSS settles within that same range of time with the least overshoot whilst CPSS is slower, settling between 4.5 to 6 seconds. In Case 6, DE settles slightly faster than PBIL whereas the system equipped with CPSS has slightly longer settling time.

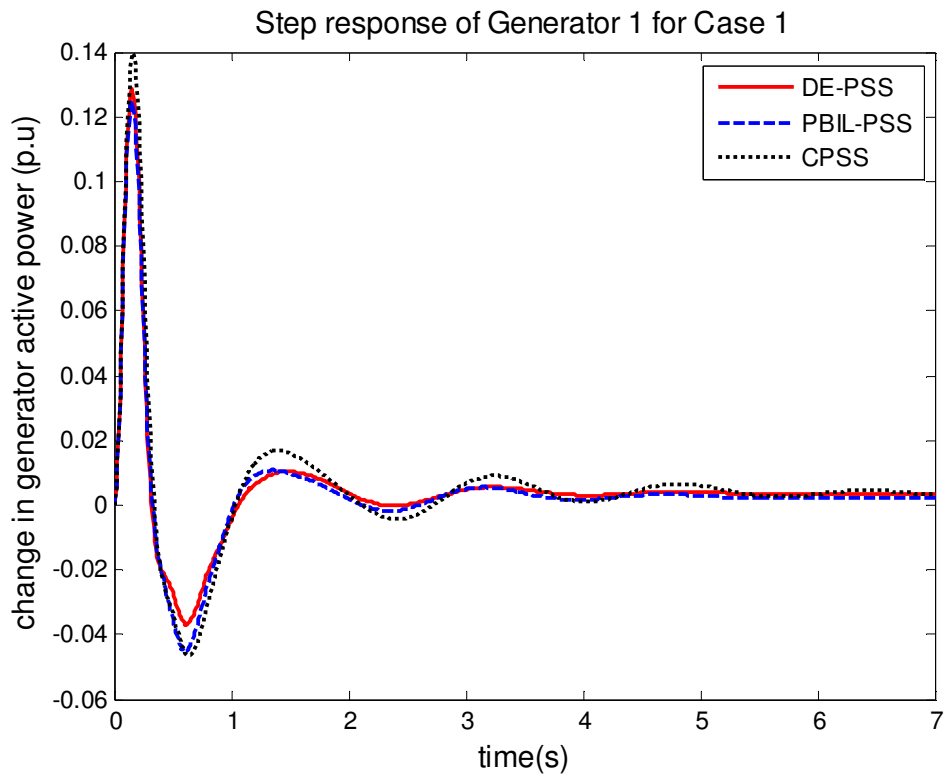


Figure 7.3: Active power response of G1 for Case 1

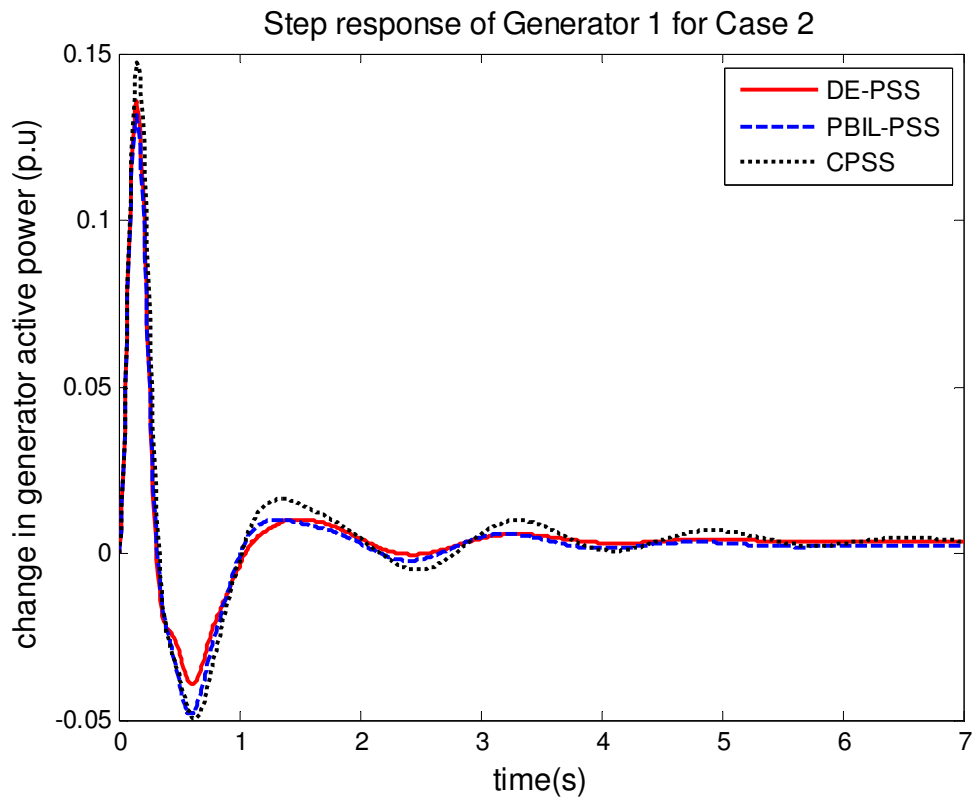


Figure 7.4: Active power response of G1 for Case 2

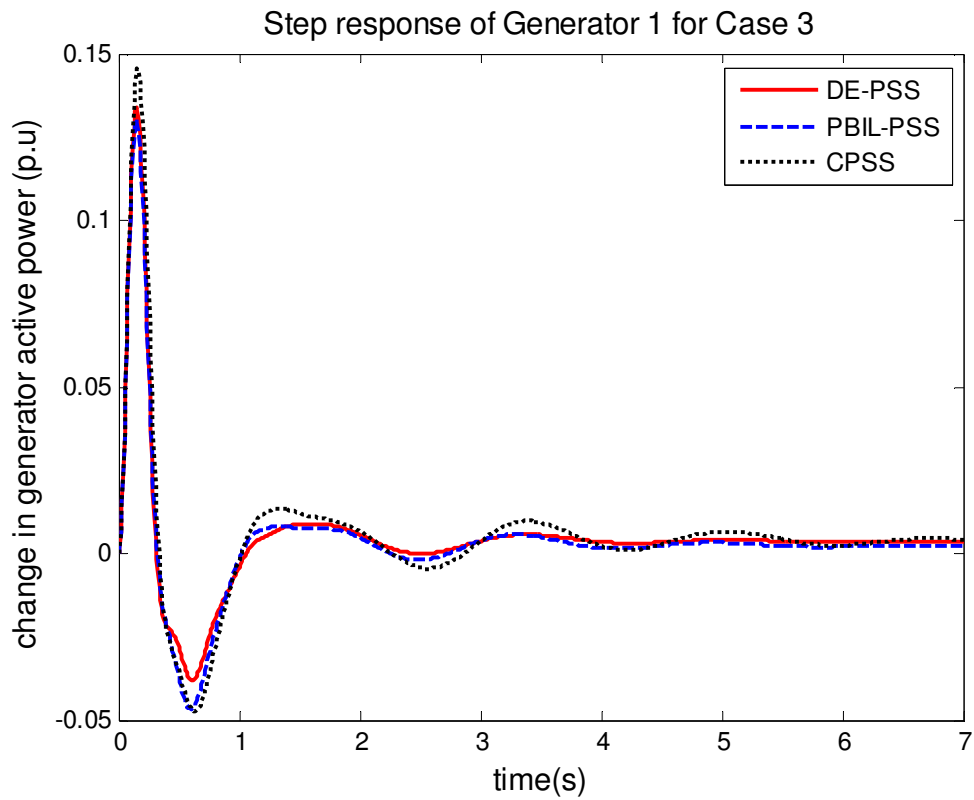


Figure 7.5: Active power response of G1 for Case 3

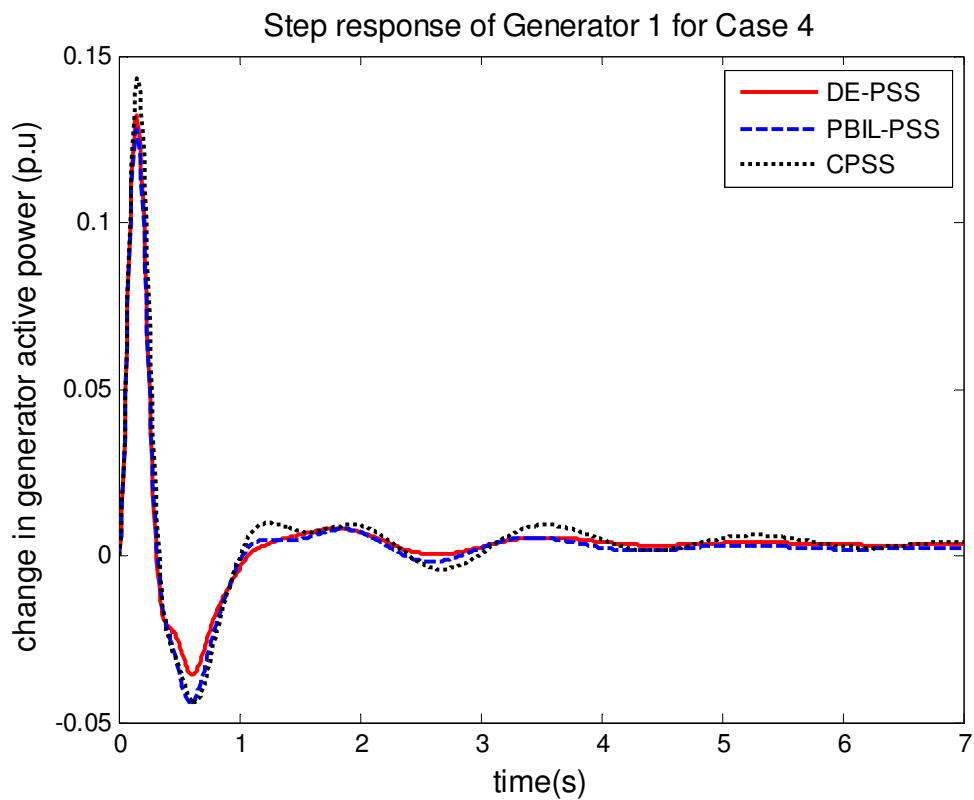


Figure 7.6: Active power response of G1 for Case 4

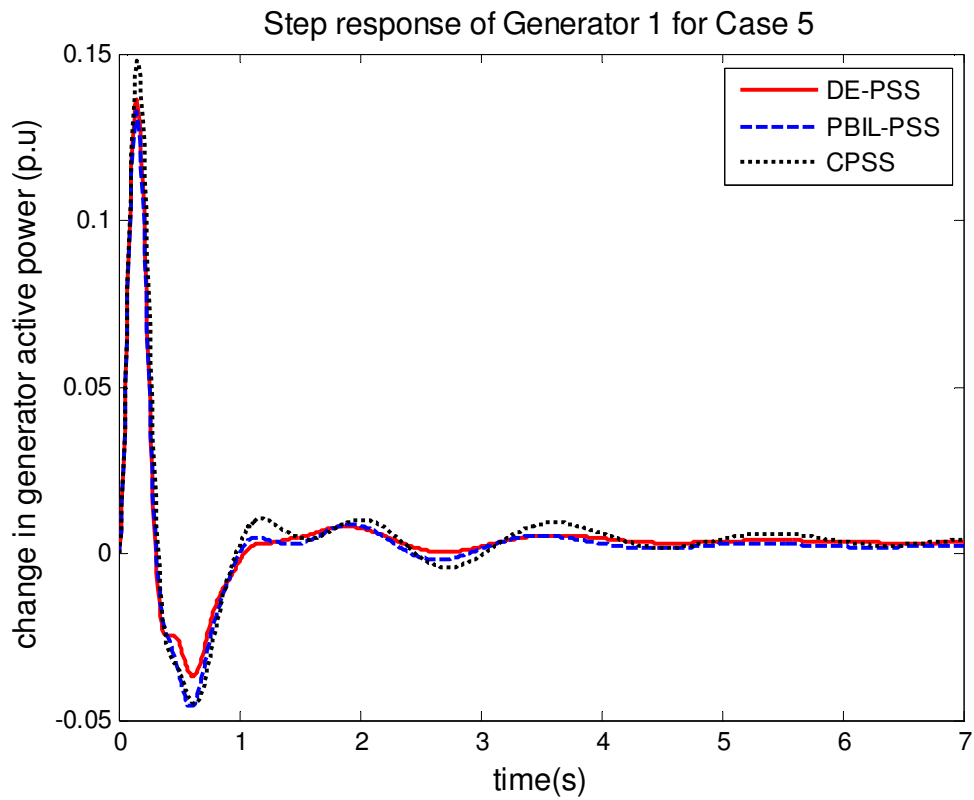


Figure 7.7: Active power response of G1 for Case 5

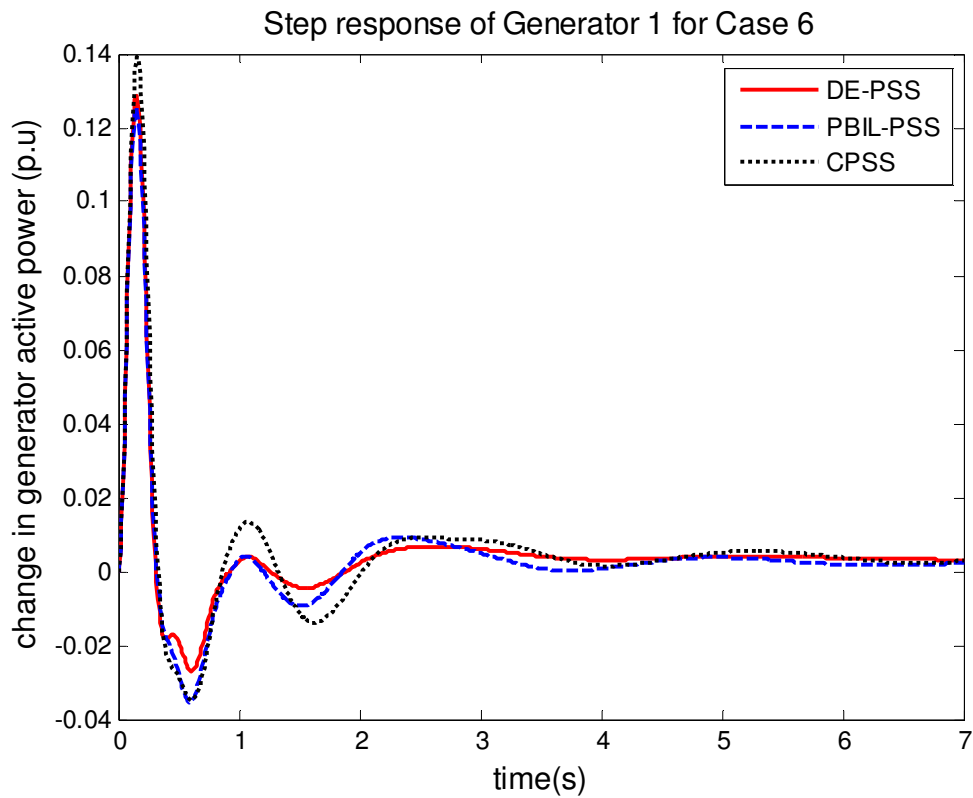


Figure 7.8: Active power response of G1 for Case 5

Figure 7.9 to Figure 7.14 illustrate the active power response of generator 2 when 10% step change in reference voltage is applied on the same generator. Despite the fact that both PSSs installed on generator 1 and 2 are identical, the results are different due to the fact that generator 2 has the highest participation on the inter-area mode. Hence, unlike generator 1, DE-PSS performs better than both PBIL-PSS and CPSS for all cases. It can be observed that DE-PSS displays the least overshoots and undershoots. Moreover, its transient decay within the $\pm 2\%$ band between 3 and 4 seconds deduced from the real part of the eigenvalue analysed. PBIL-PSS, however, is slightly more oscillatory and only settles after 4 seconds. Similarly to generator 1, the oscillations on the system equipped with CPSS are sustained longer with a settling time of 5 to 6 seconds.

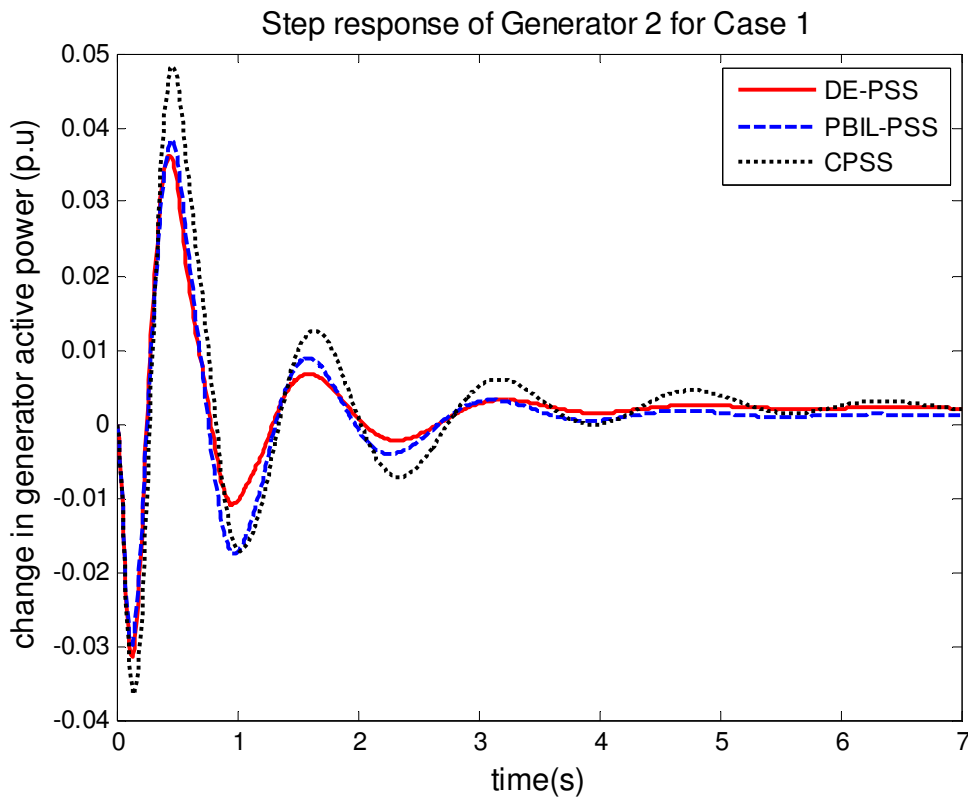


Figure 7.9: Active power response of G2 for Case 1

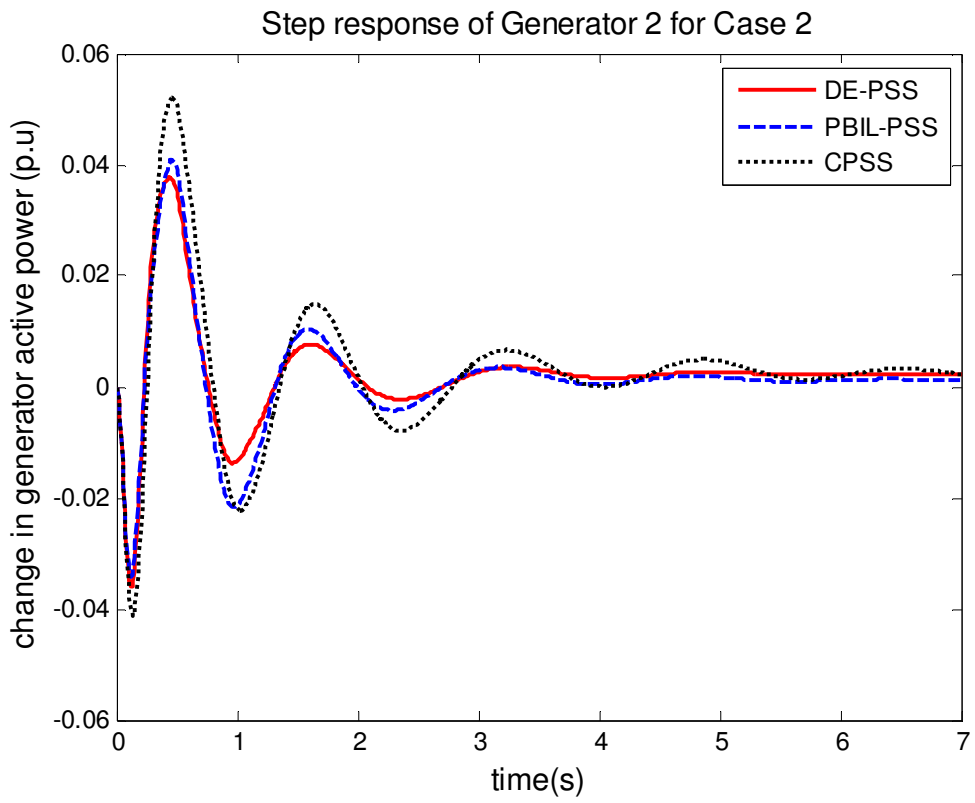


Figure 7.10: Active power response of G2 for Case 2

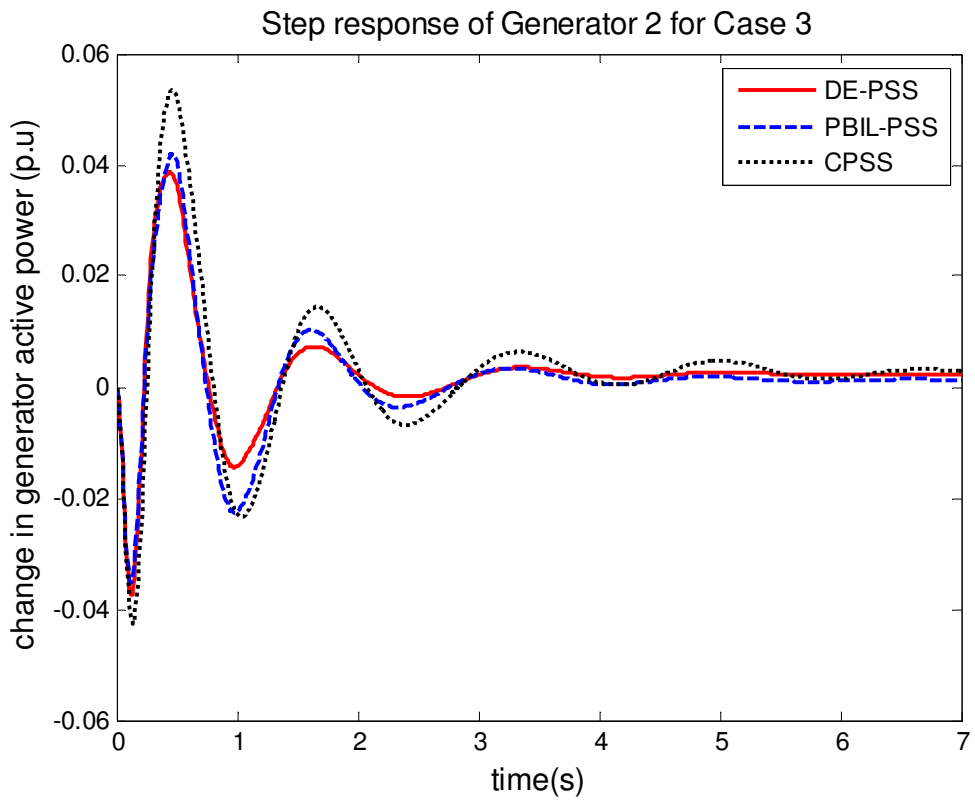


Figure 7.11: Active power response of G2 for Case 3

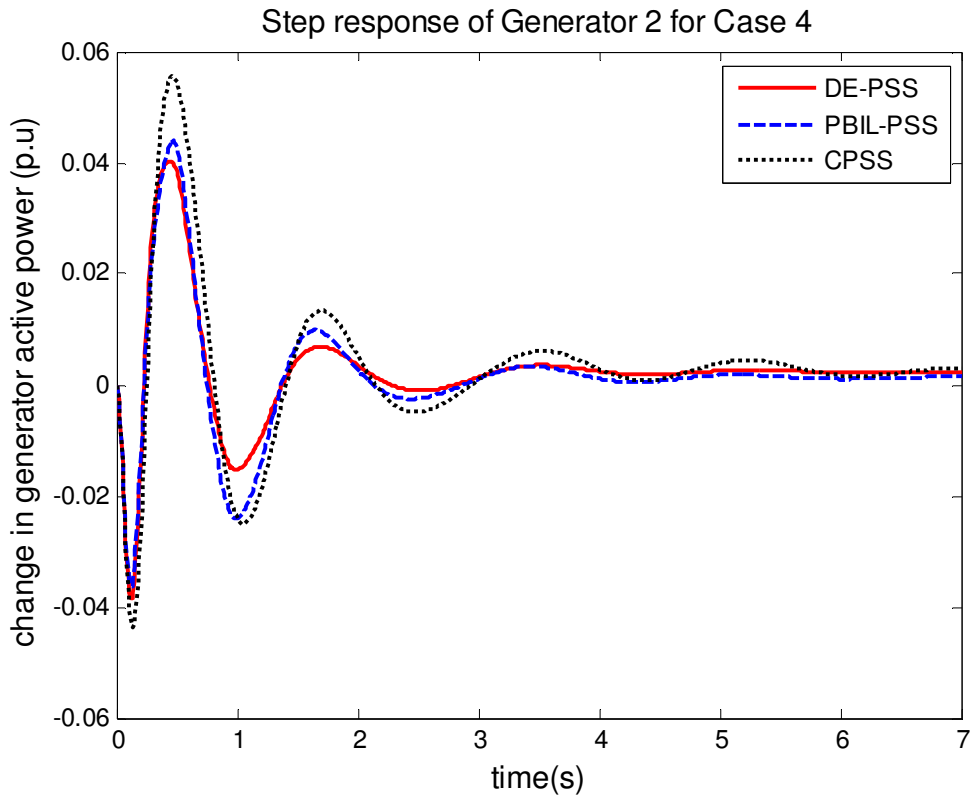


Figure 7.12: Active power response of G2 for Case 4

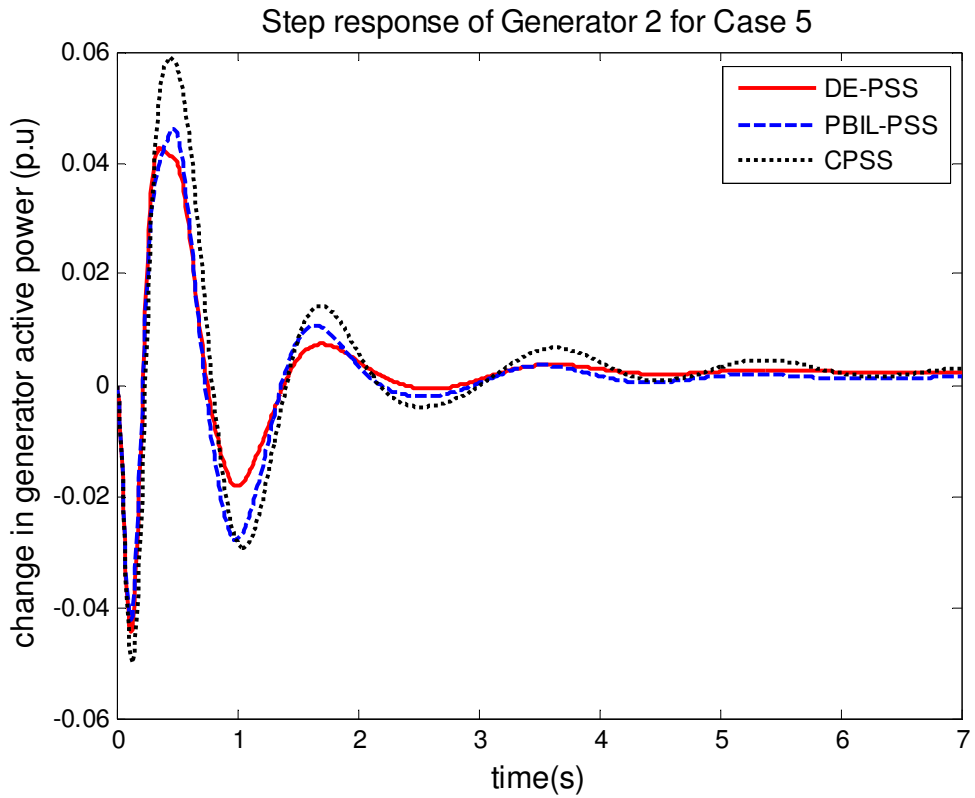


Figure 7.13: Active power response of G2 for Case 5

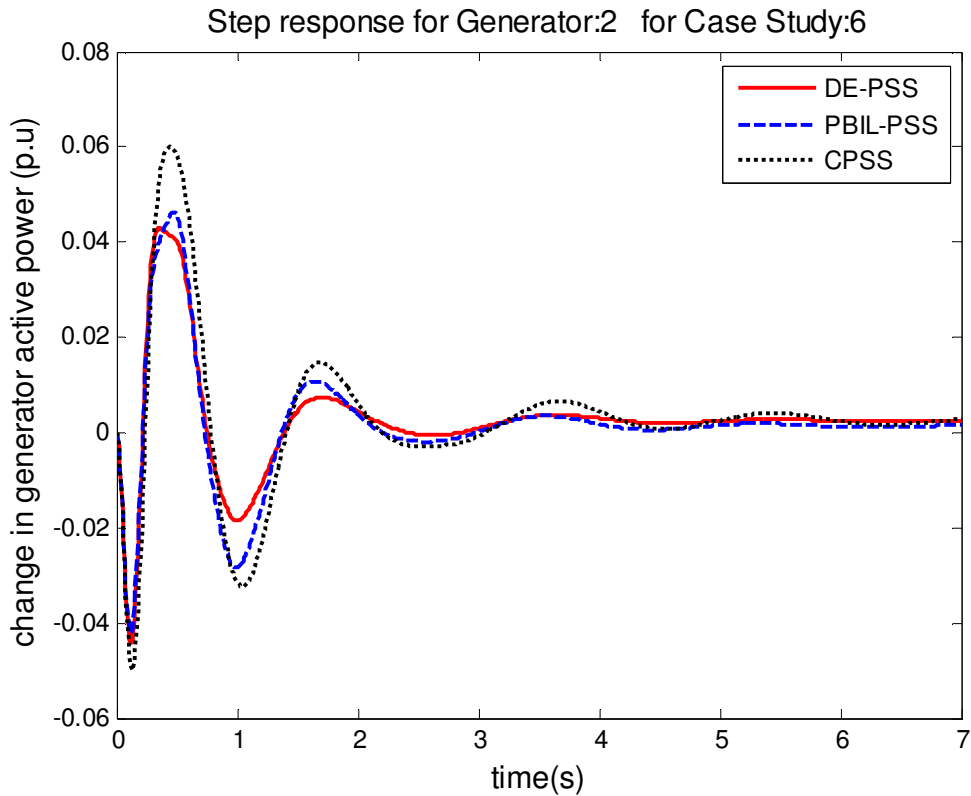


Figure 7.14: Active power response of G2 for Case 6

The step responses of generator 3 are illustrated in Figure 7.15 to Figure 7.20. Similar trends to generator 2 are observed with generator 3. The CPSS displays the least performance characterized by the largest overshoots and undershoots and slower settling times of 7 seconds across all operating conditions. As expected, both EA-PSSs perform better than the CPSS. However PBIL-PSS has larger undershoots than DE-PSS, reaching -0.013pu and a settling time of just over 4 seconds for the first three cases and over 5 seconds for the remaining cases (4, 5 and 6). Furthermore, the system equipped with PBIL-PSS stabilises slightly below those of DE-PSS and CPSS. It is observed that DE-PSS has the best performance with the least overshoots and undershoots. This system settles within 3.5 seconds for Case 1, 2 and 3 and a little over 4 seconds for Case 4, 5 and 6.

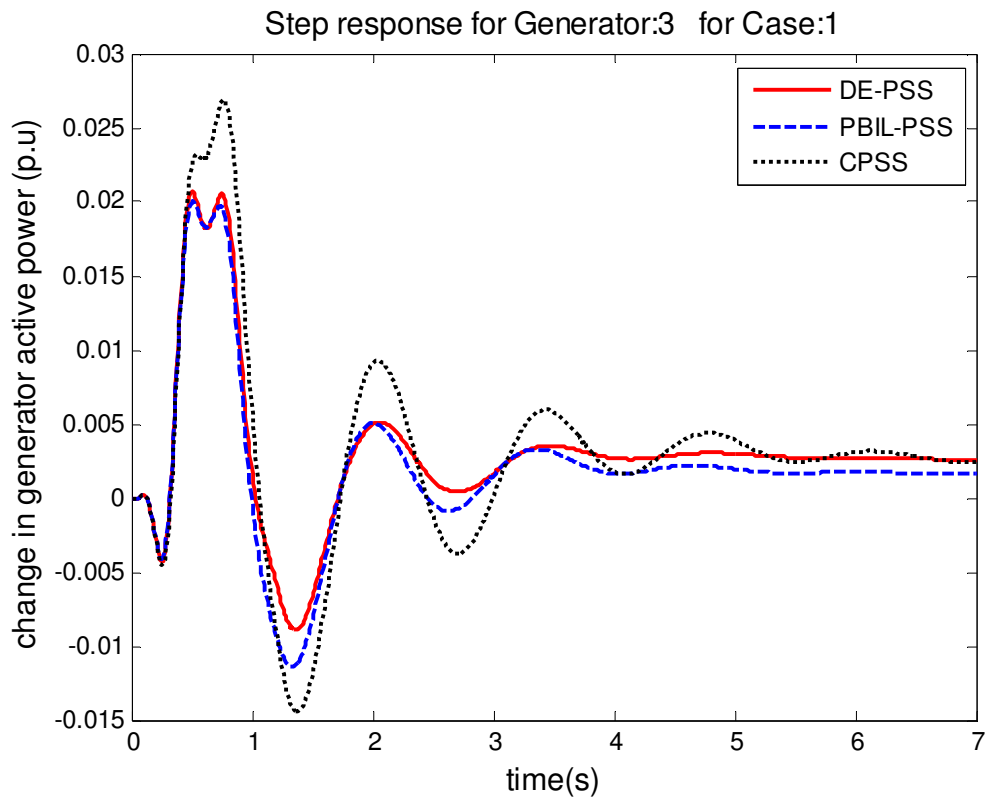


Figure 7.15: Active power response of G3 for Case 1

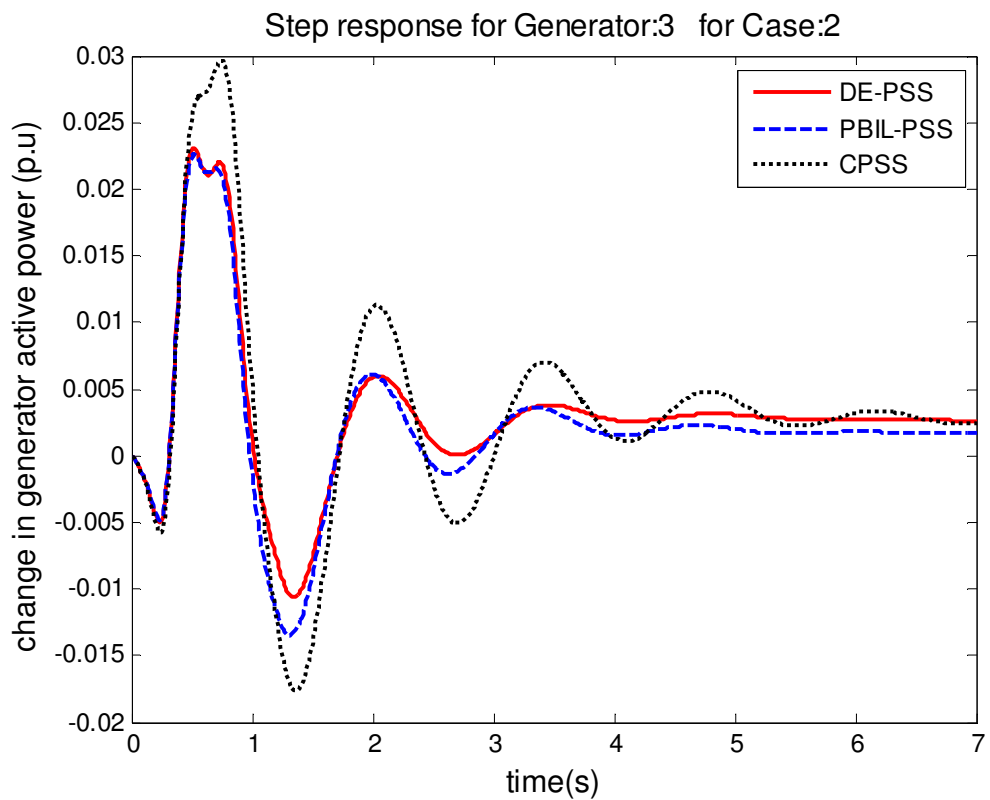


Figure 7.16: Active power response of G3 for Case 2

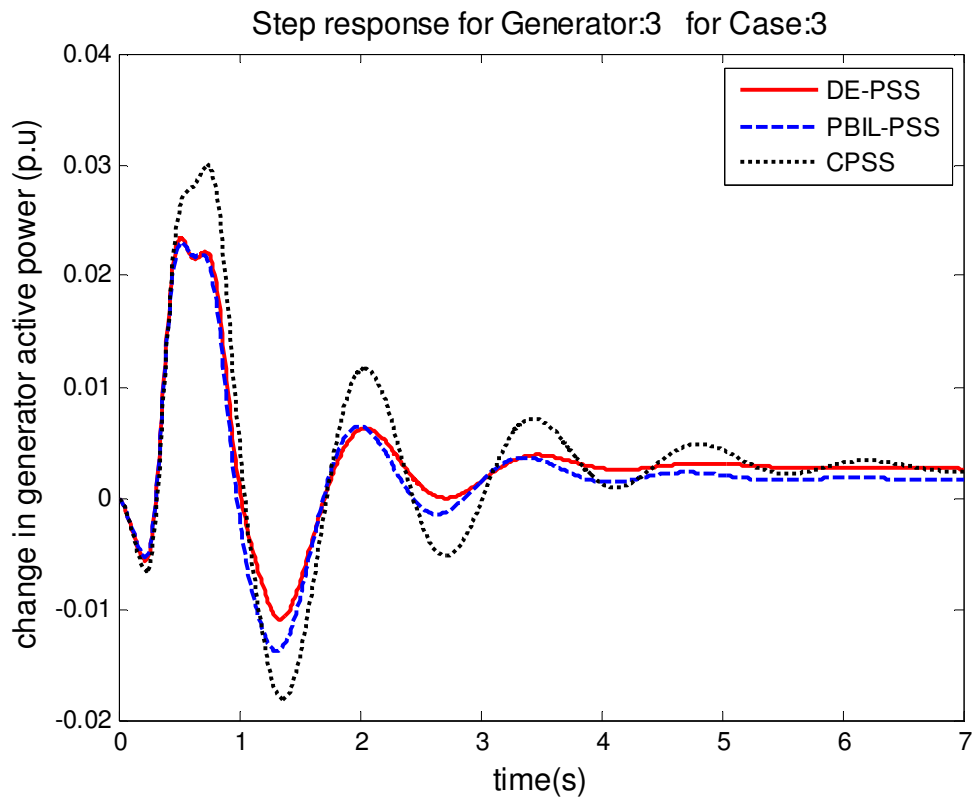


Figure 7.17: Active power response of G3 for Case 3

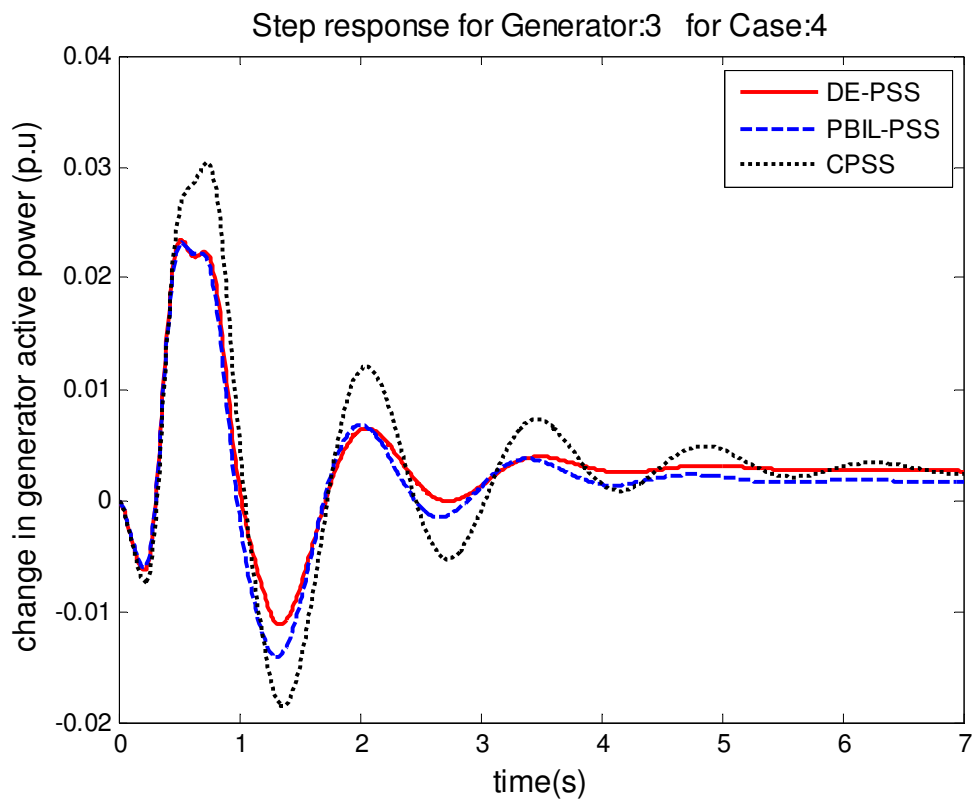


Figure 7.18: Active power response of G3 for Case 4

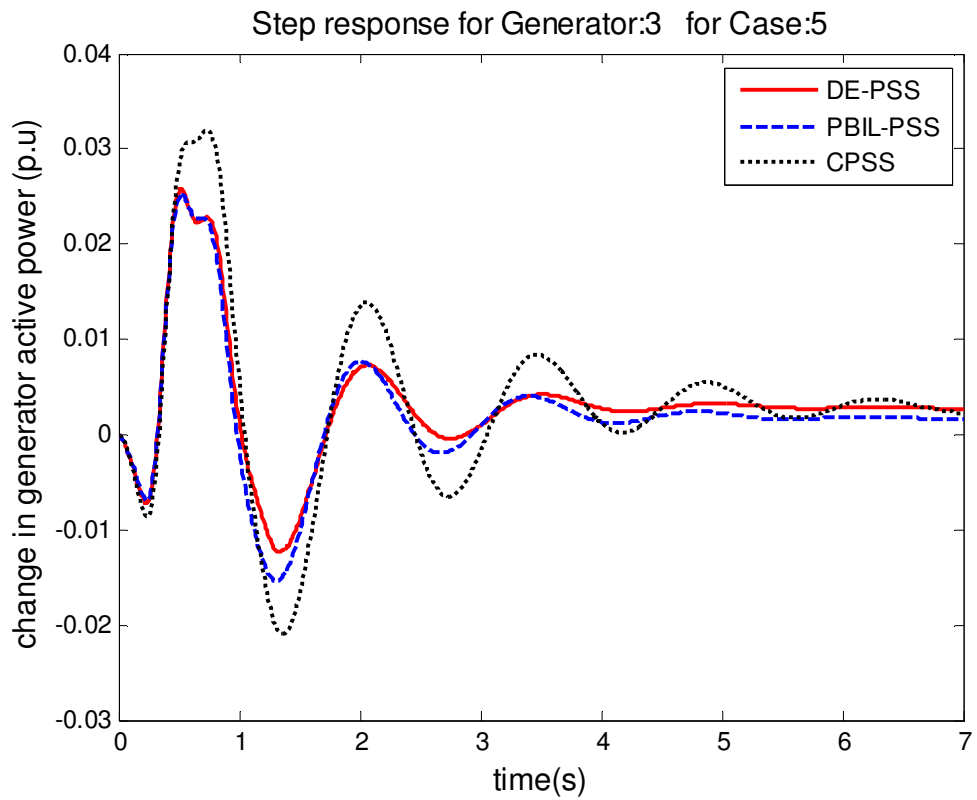


Figure 7.19: Active power response of G3 for Case 5

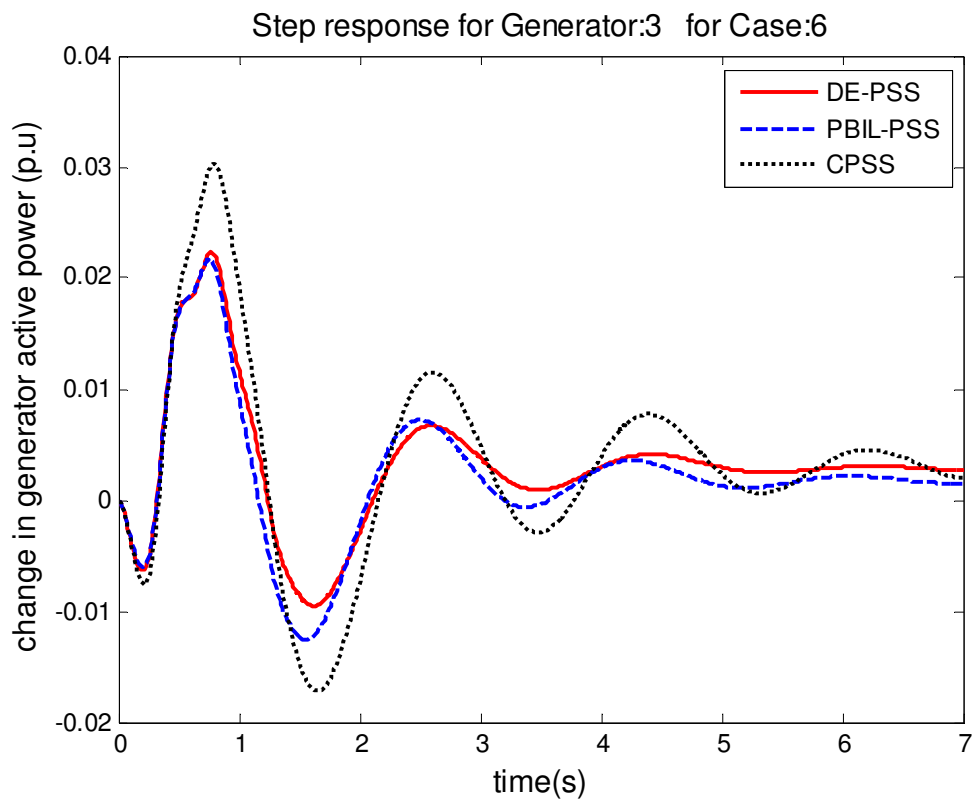


Figure 7.20: Active power response of G3 for Case 6

Figure 7.21 to Figure 7.26 show the step responses of generator 4. Similar to generator 3, the system equipped with CPSS performs the least. In Case 1, the system settles after 5 seconds. However as the system loading increases so does the settling time. Thus, CPSS settles in 6 and after 7 seconds for Case 2 to 6. PBIL-PSS settles slightly faster, in 4 to 4.5 seconds for Case 1 to 5 whilst in Case 6, the system settles in 5.5 seconds. DE-PSS settles around 3 seconds for Case 1, 2, 3 and 4 whilst for Case 5 and 6, the system settles around 4 seconds. Moreover, it can be observed that DE-PSS exhibits smaller overshoots and undershoots. This indicates that the DE-PSS performs better than PBIL-PSS and of course CPSS.

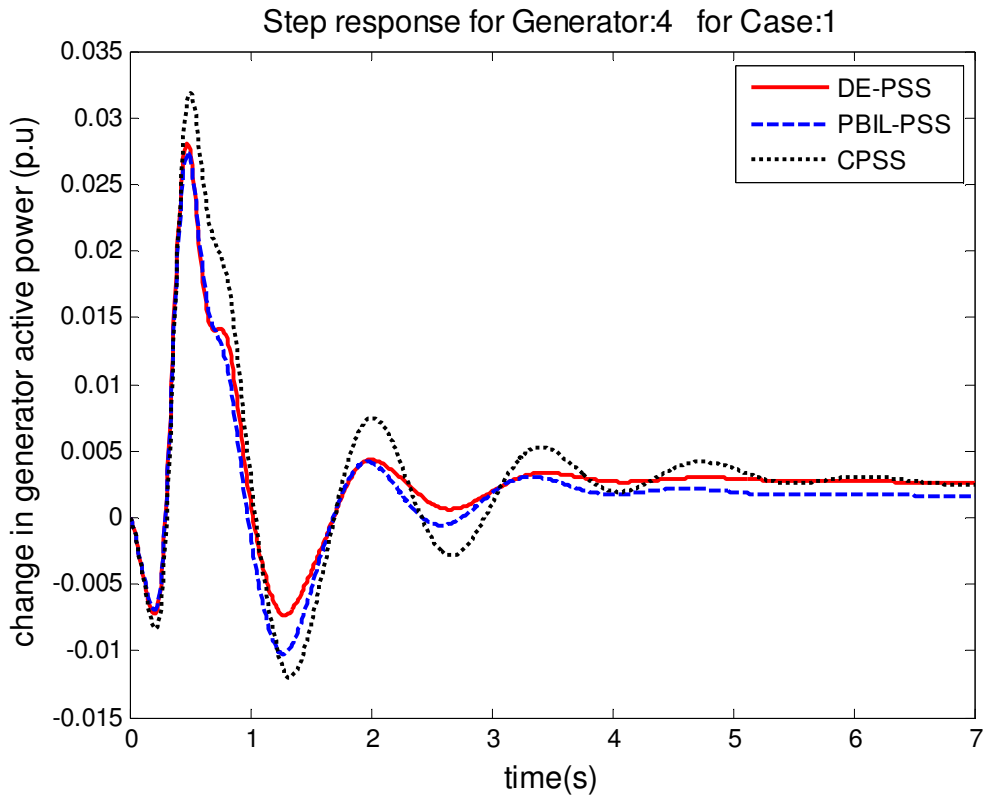


Figure 7.21: Active power response of G4 for Case 1

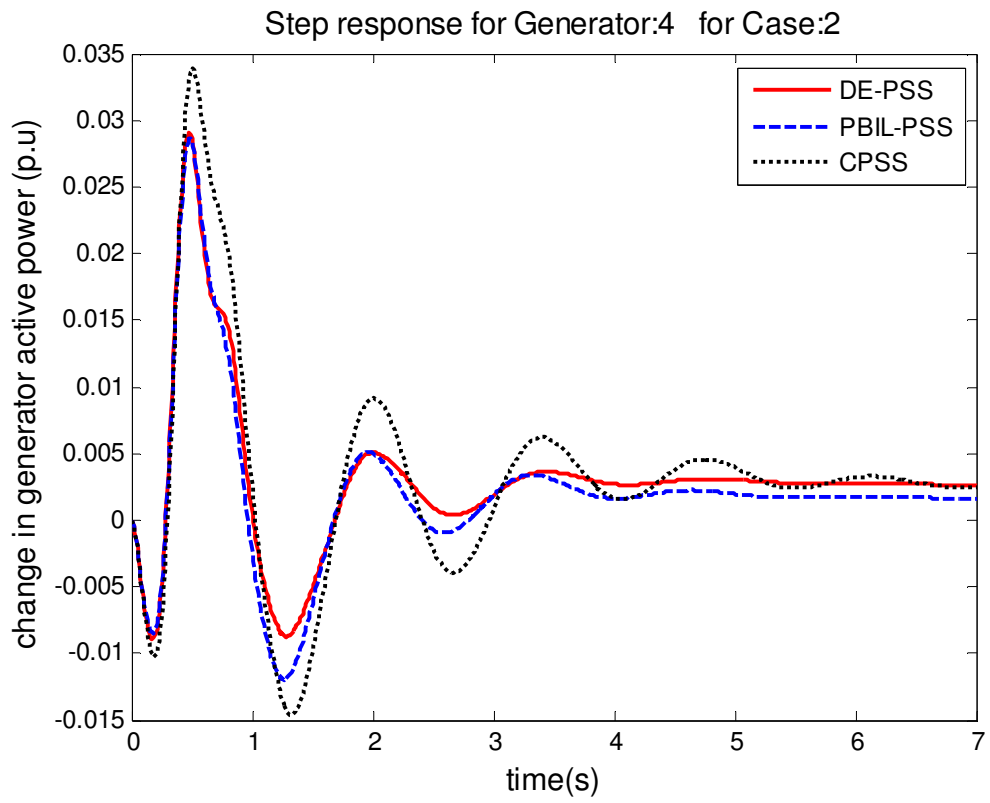


Figure 7.22: Active power response of G4 for Case 2

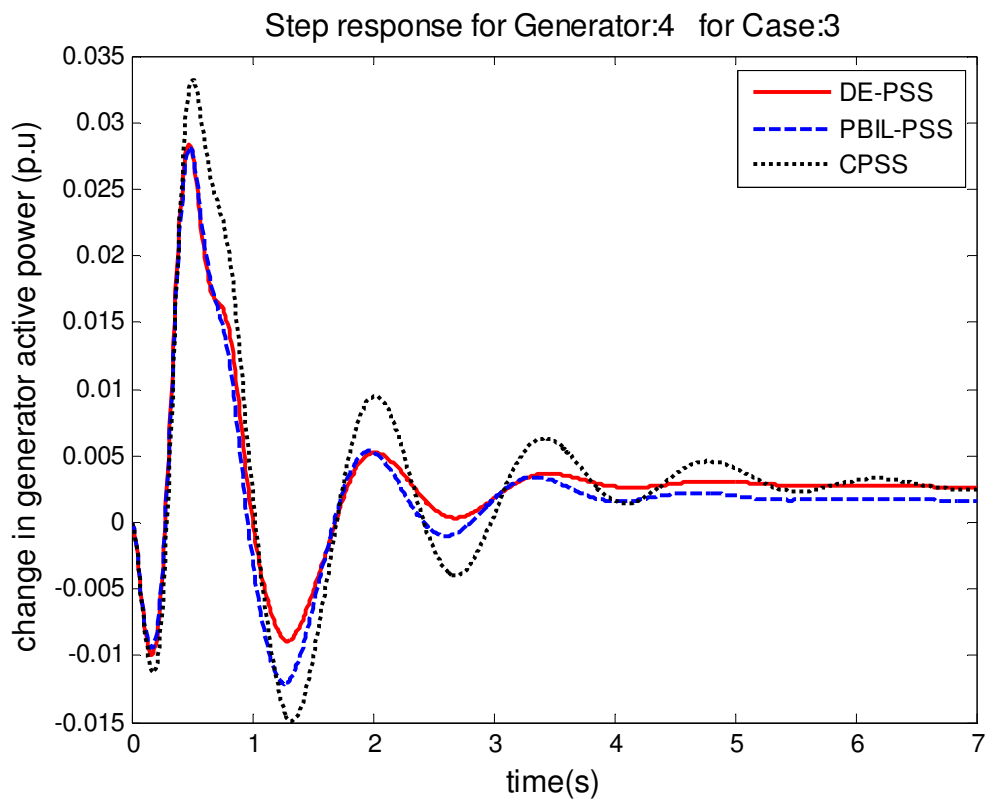


Figure 7.23: Active power response of G4 for Case 3

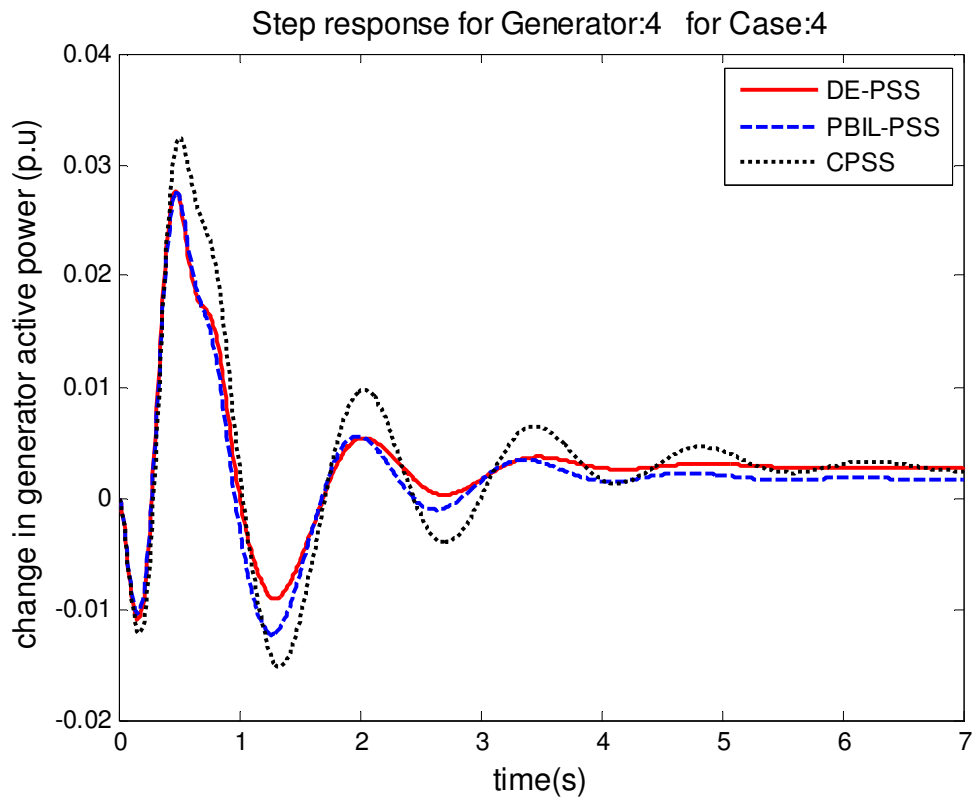


Figure 7.24: Active power response of G4 for Case 4

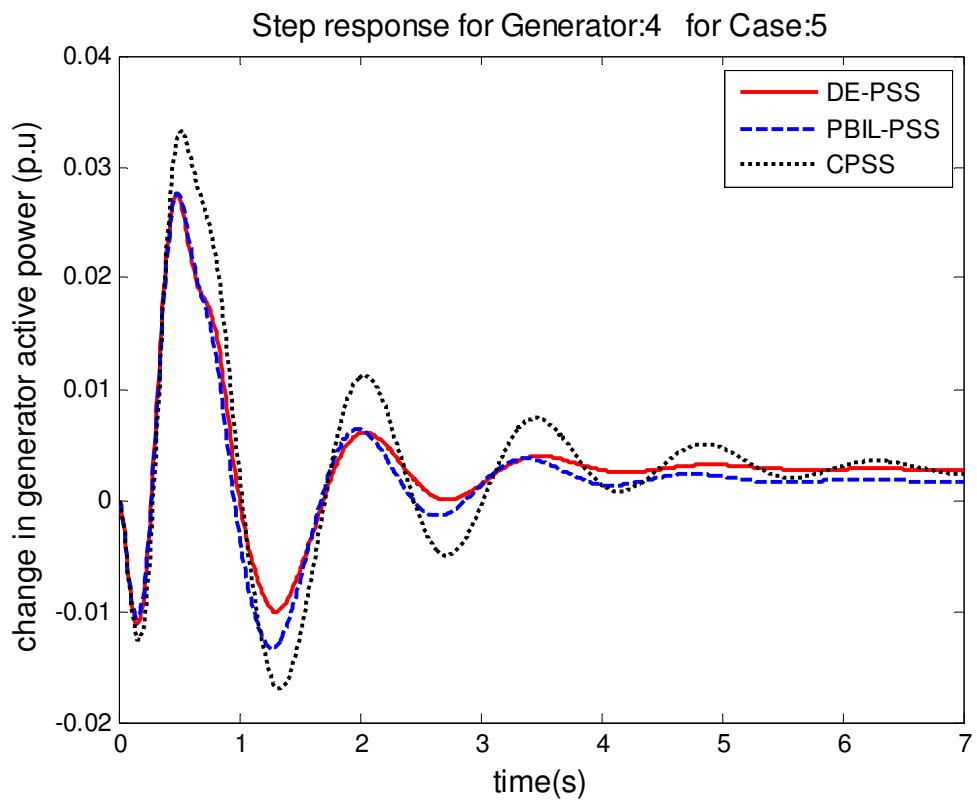


Figure 7.25: Active power response of G4 for Case 5

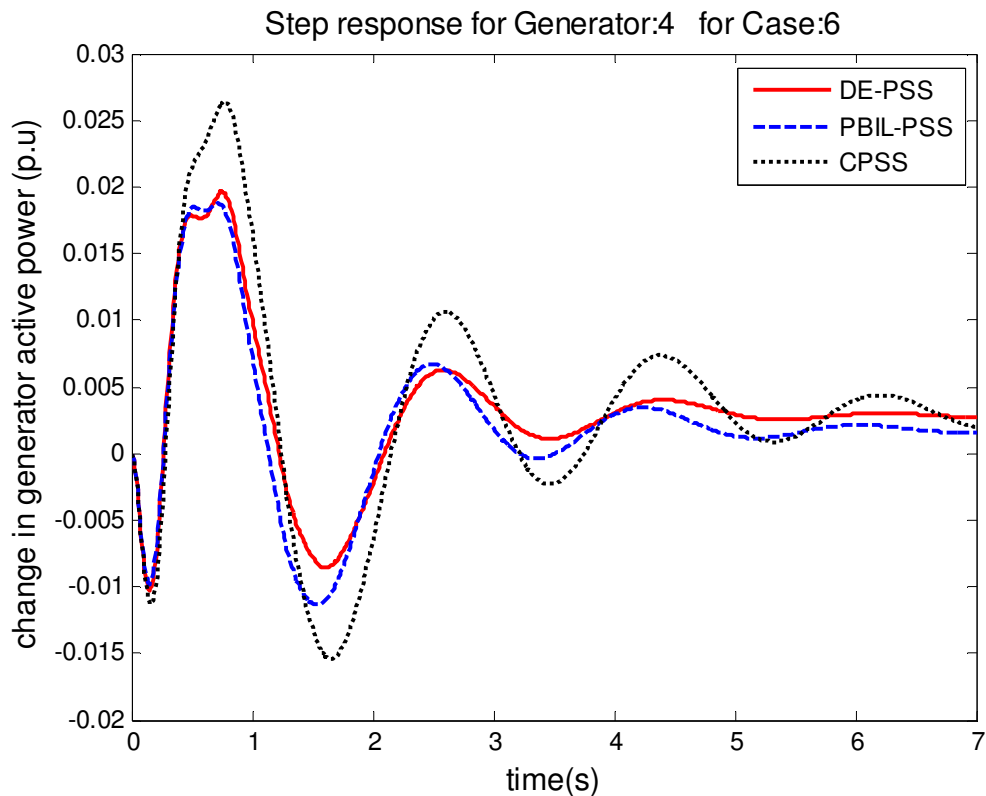


Figure 7.26: Active power response of G4 for Case 6

7.4.2 Large disturbance simulation

For the transient stability, a more severe fault was considered to further evaluate the PSSs robustness. The fault consists of submitting the system to a 5-cycles 3-phase fault applied after 1 second on line 1 (Figure 5.8). The fault is cleared by disconnecting the line (between bus 5 and 9).

Transient stability was performed for Case 1 to 4. Case 5 and 6 were not considered for the transient stability analysis because in Case 5, the system is unable to sustain the severity of the fault due to the high power angle. As for Case 6, the system is operating with one tie-line.

Since generator 2 has the highest participation on the inter-area mode, the results presented below are those observed from G2.

7.4.2.1 Case 1: nominal condition where 100MW are transmitted over the tie-lines;

Figure 7.27 to Figure 7.30 illustrate the responses of the terminal voltage at bus 3, the rotor speed, the field voltage and active power following a three phase short circuit fault on line 1 (bus 5 to bus 6) at nominal operating condition. All PSSs perform adequately and rapidly return to steady state operation after the fault is cleared. The three PSSs have similar settling times for the terminal voltage, electric field and active power. Whilst for the speed response, the CPSS is slightly slower, settling after 5 seconds. However, in some instances the CPSS displays larger overshoots and undershoots. Both DE and PBIL PSSs perform similarly despite the slight overshoots observed on the system equipped with DE-PSS that are due to its high PSS gain.

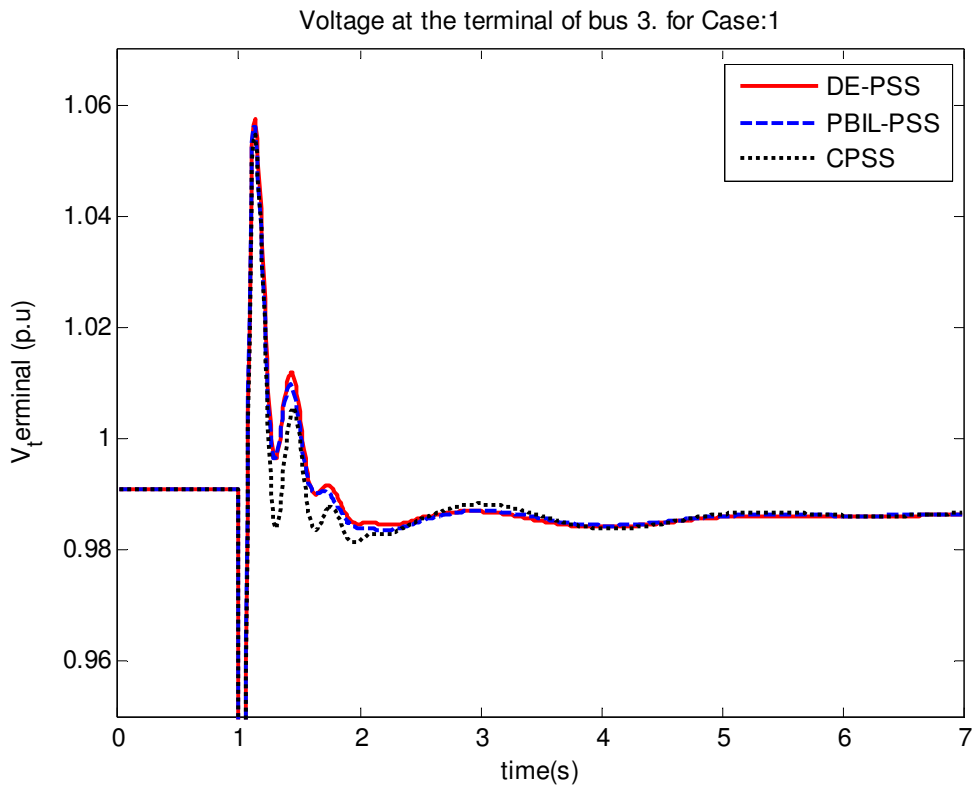


Figure 7.27: Terminal voltage response following a 3-phase fault for Case 1

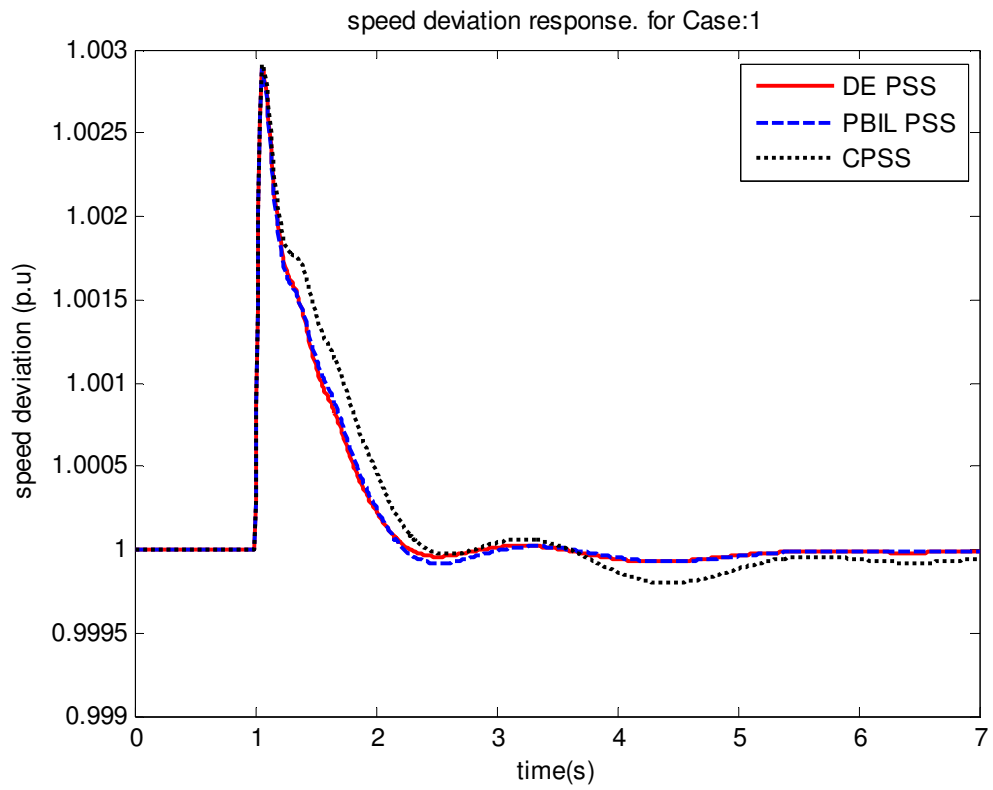


Figure 7.28: Rotor speed response following a 3-phase fault for Case 1

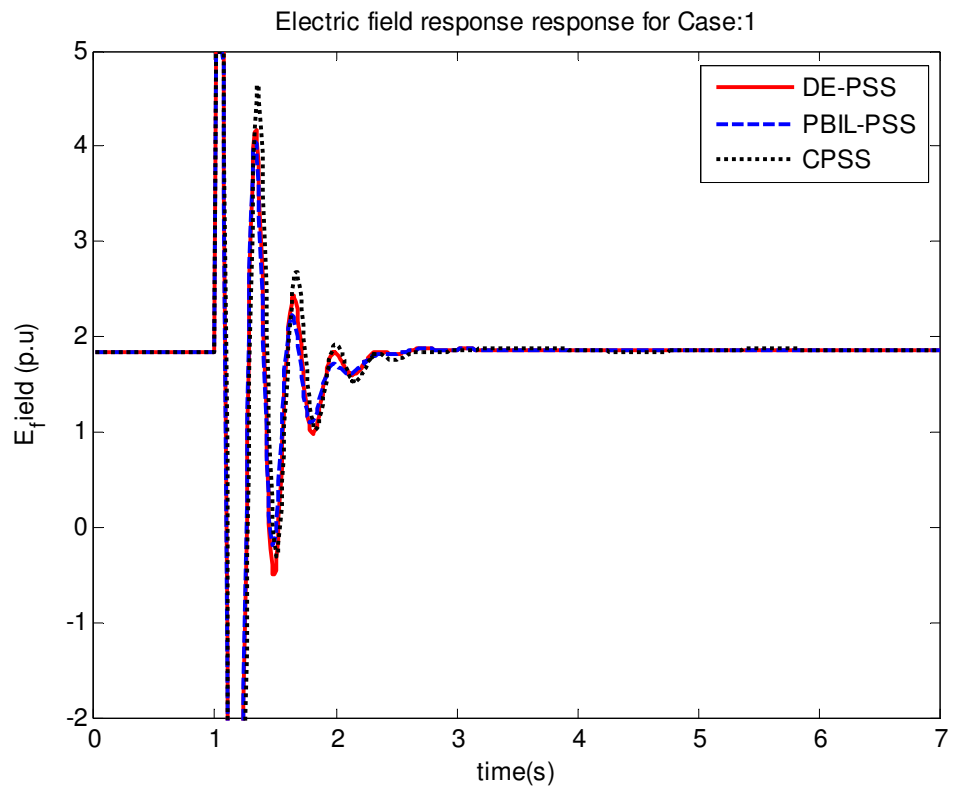


Figure 7.29: Electric field following a 3-phase fault for Case 1

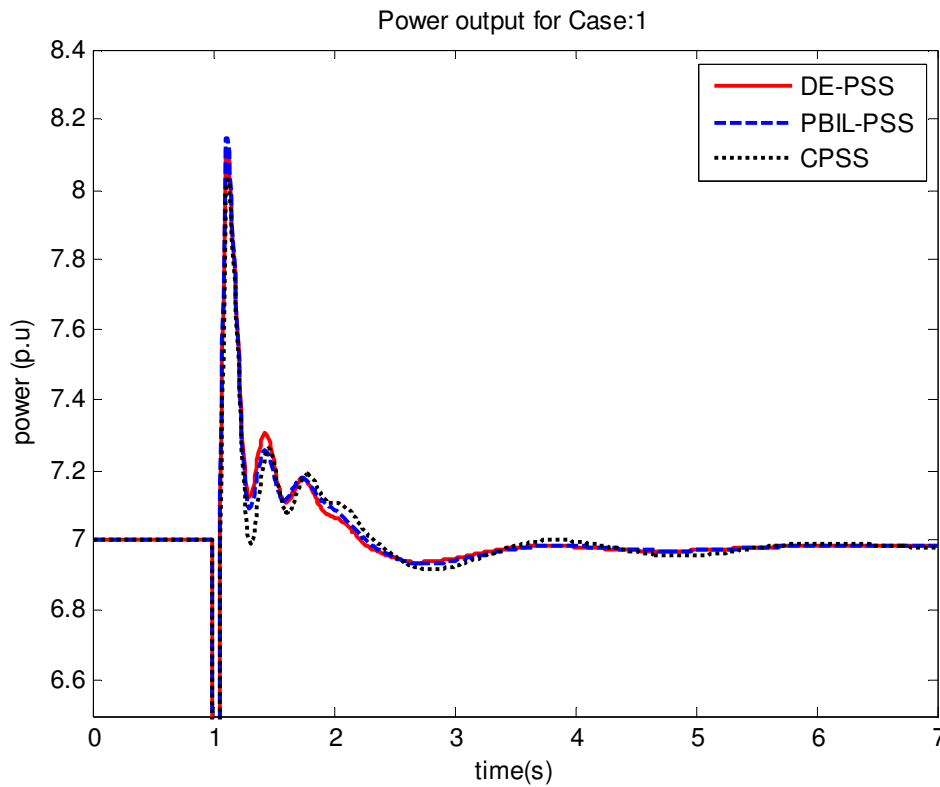


Figure 7.30: Active power output following 3-phase fault for Case 1

7.4.2.2 Case 2: 3-phase fault applied when 200MW are transmitted from area 1 to 2

Figure 7.31 to Figure 7.34 illustrate the responses of the terminal voltage at bus 3, the rotor speed, the field voltage and active power, respectively. It can be observed that the CPSS is slightly more oscillatory than the previous case. In Figure 7.31 the CPSS settles after 7 seconds whereas in Figure 7.32 to 7.34 it settles in 6 seconds. Despite the slightly higher overshoots observed in Figure 7.31, 7.33 and 7.34, DE-PSS performs similarly to PBIL-PSS with a settling time of about 4 seconds for the terminal voltage and active power, whilst settling in 3 seconds for rotor speed and electric field. Hence the PBIL and DE based PSSs perform better than the CPSS.

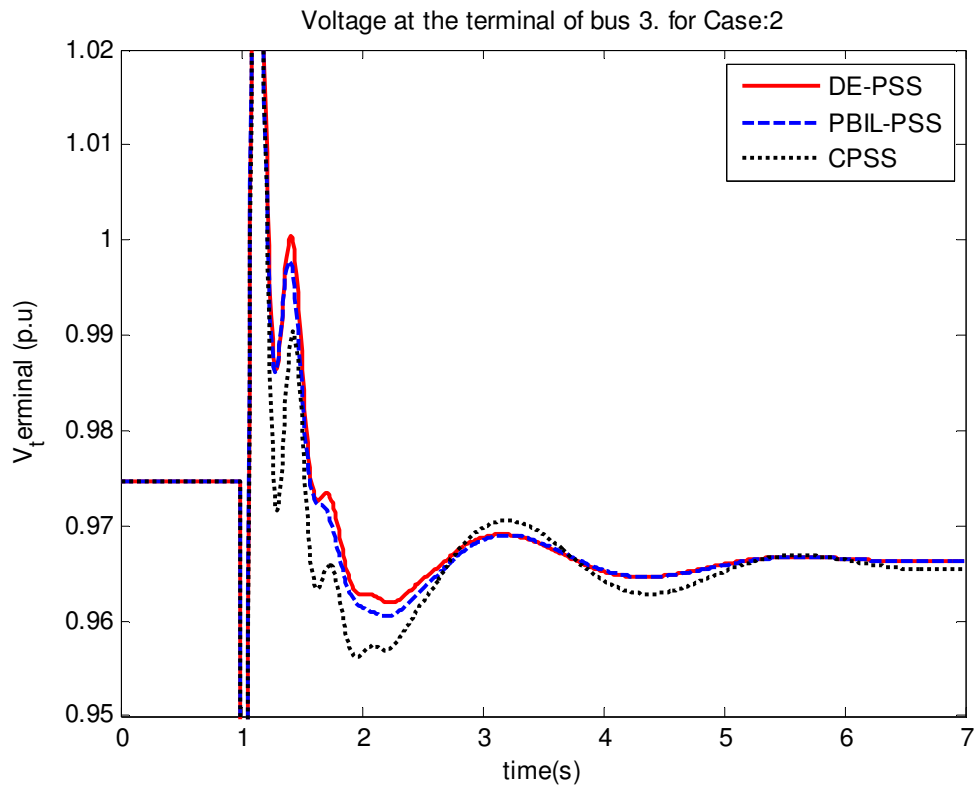


Figure 7.31: Terminal voltage response following a 3-phase fault for Case 2

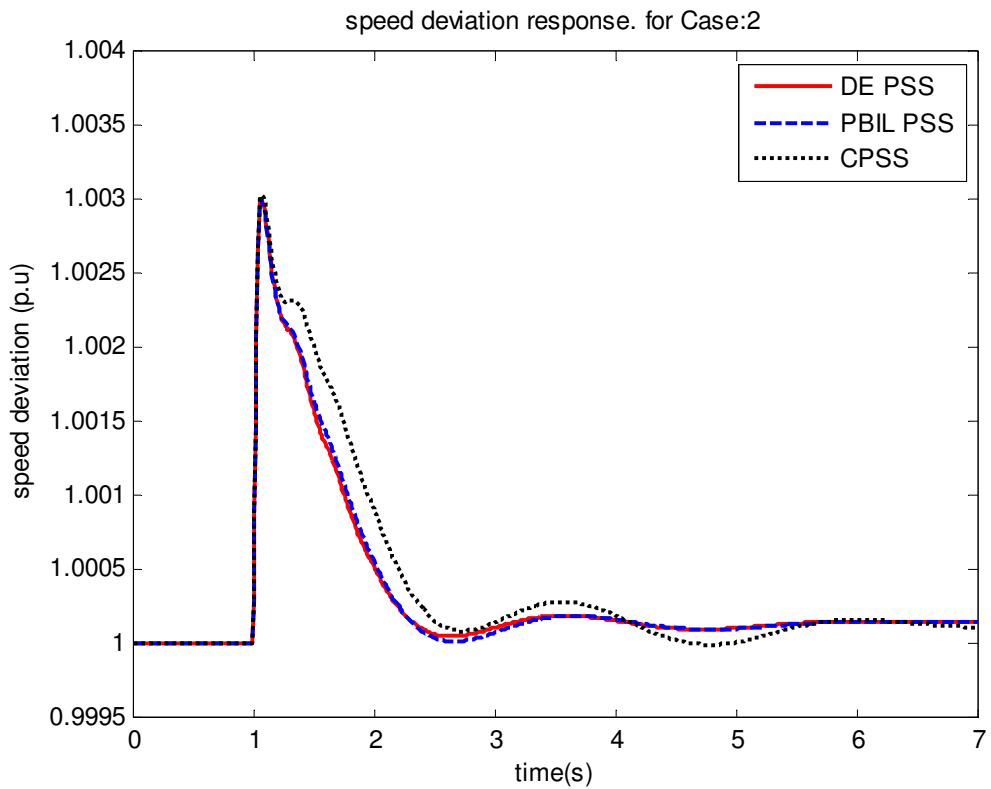


Figure 7.32: Rotor speed response following a 3-phase fault for Case 2

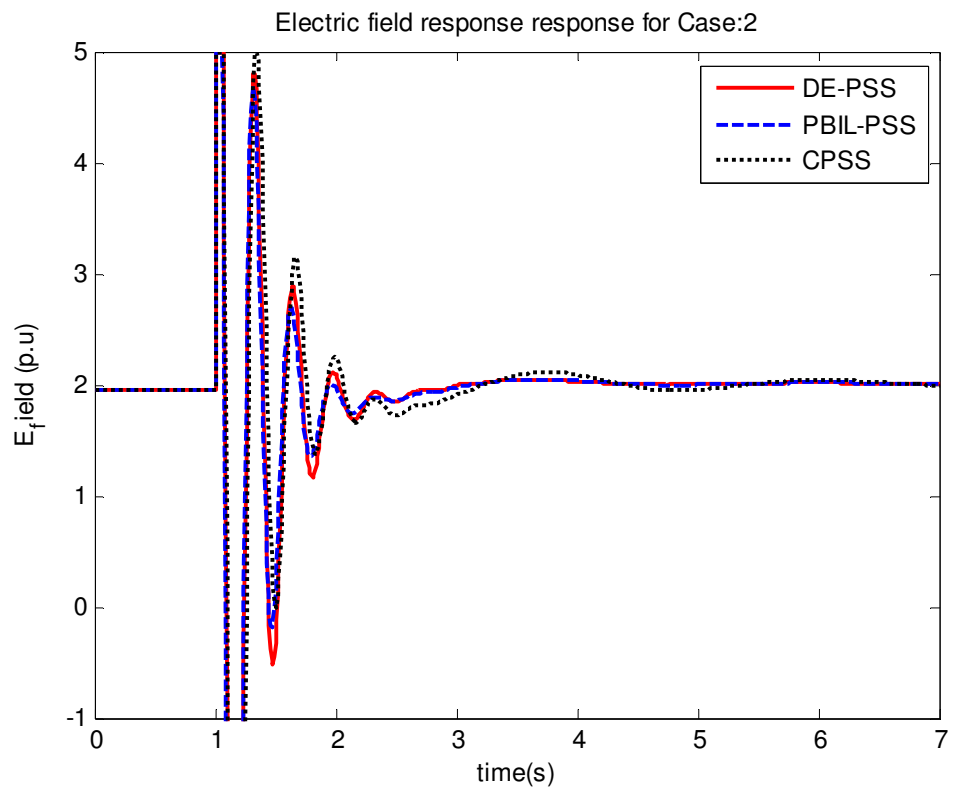


Figure 7.33: Electric field following a 3-phase fault for Case 2

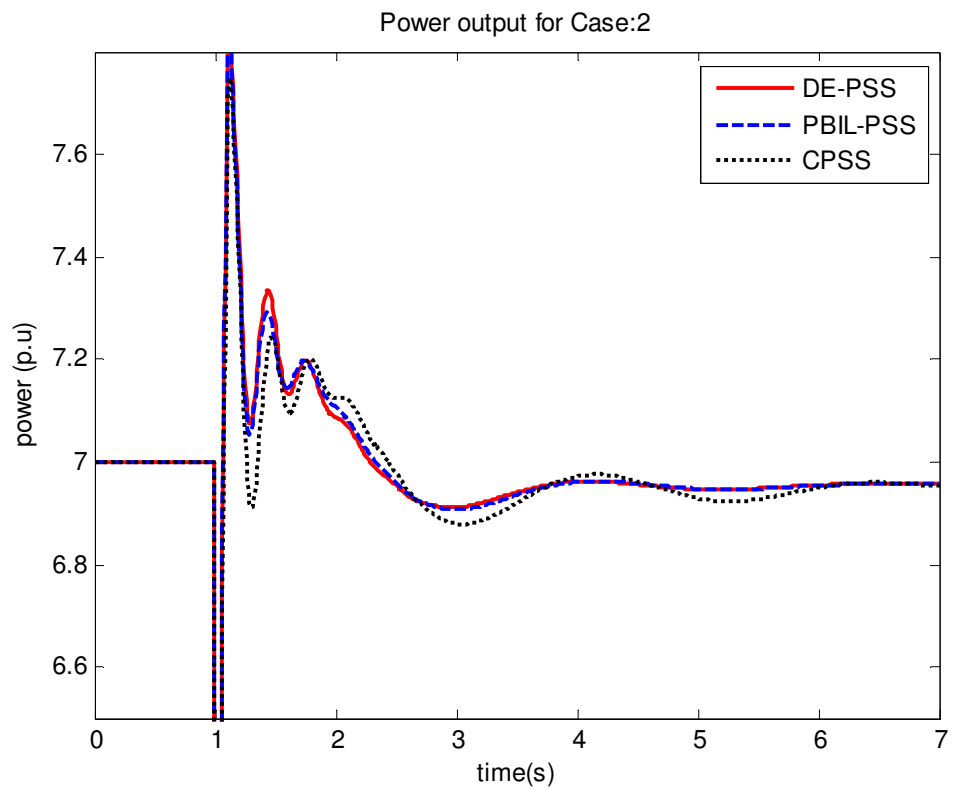


Figure 7.34: Active power output following 3-phase fault for Case 2

7.4.2.3 Case 3: 3-phase fault applied when 300MW are transmitted from area 1 to 2

The responses of the terminal voltage at bus 3, the rotor speed, the field voltage and active power are shown in Figure 7.35 to Figure 7.38, respectively. In a similar trend to the previous cases, both DE-PSS and PBIL-PSS perform similarly with settling times of 4.5 seconds for the terminal voltage and 4 seconds for the speed, electric field and power. As expected, the CPSS is slower, settling after 7 seconds with slight overshoots.

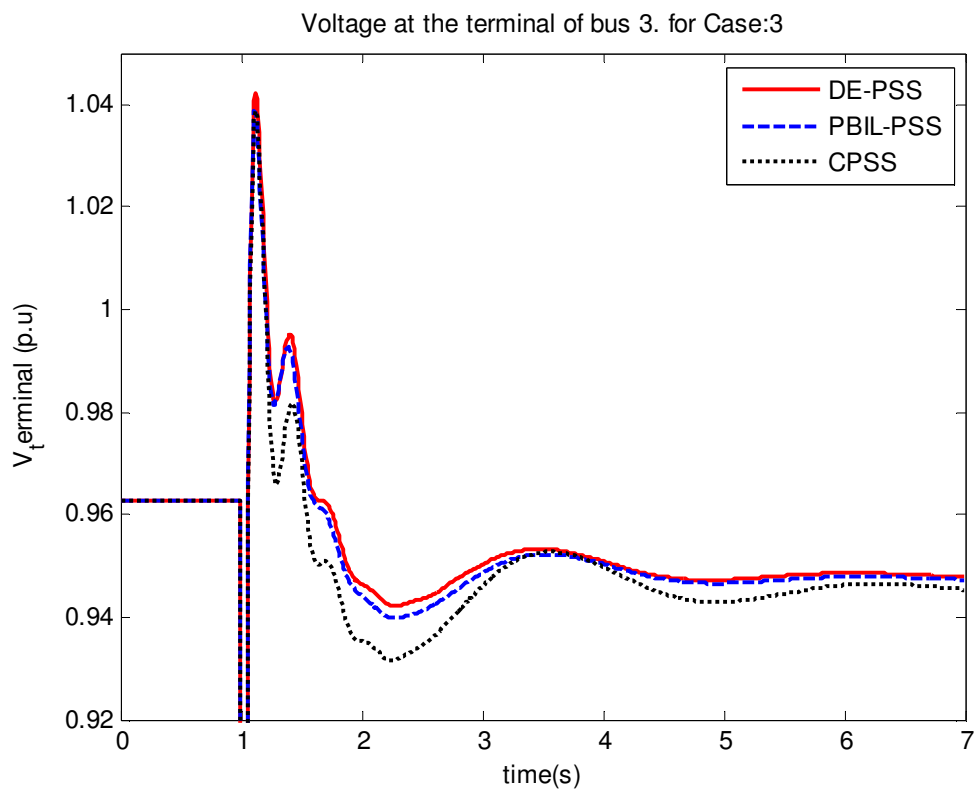


Figure 7.35: Terminal voltage response following a 3-phase fault for Case 3

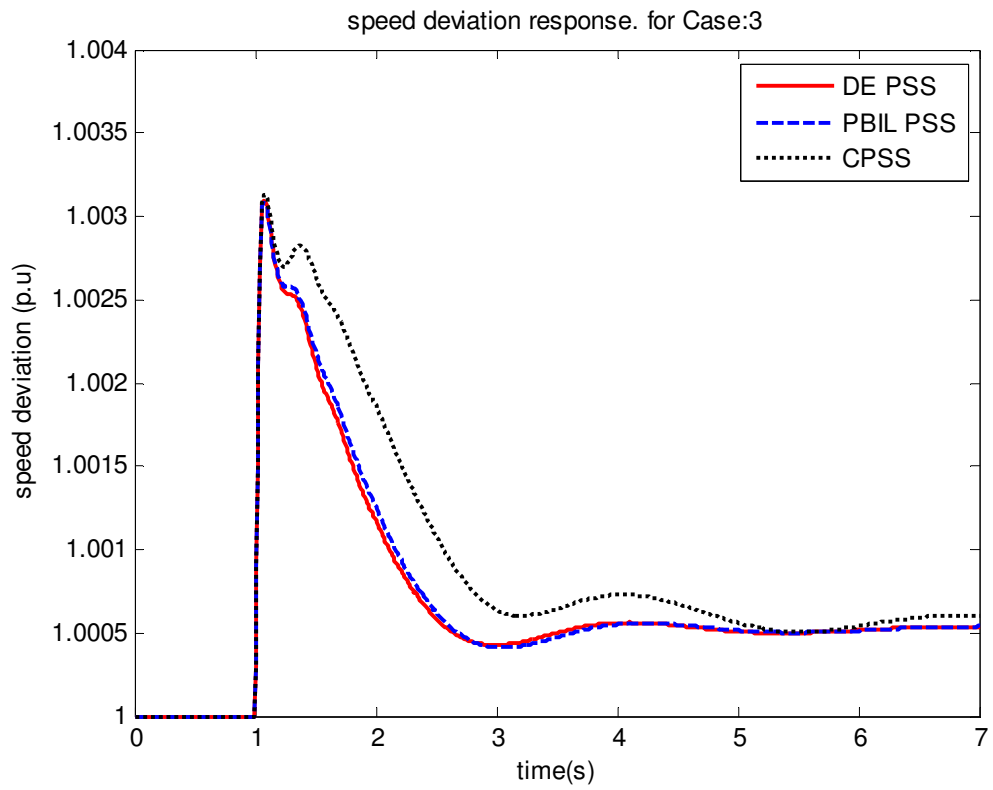


Figure 7.36: Rotor speed response following a 3-phase fault for Case 3

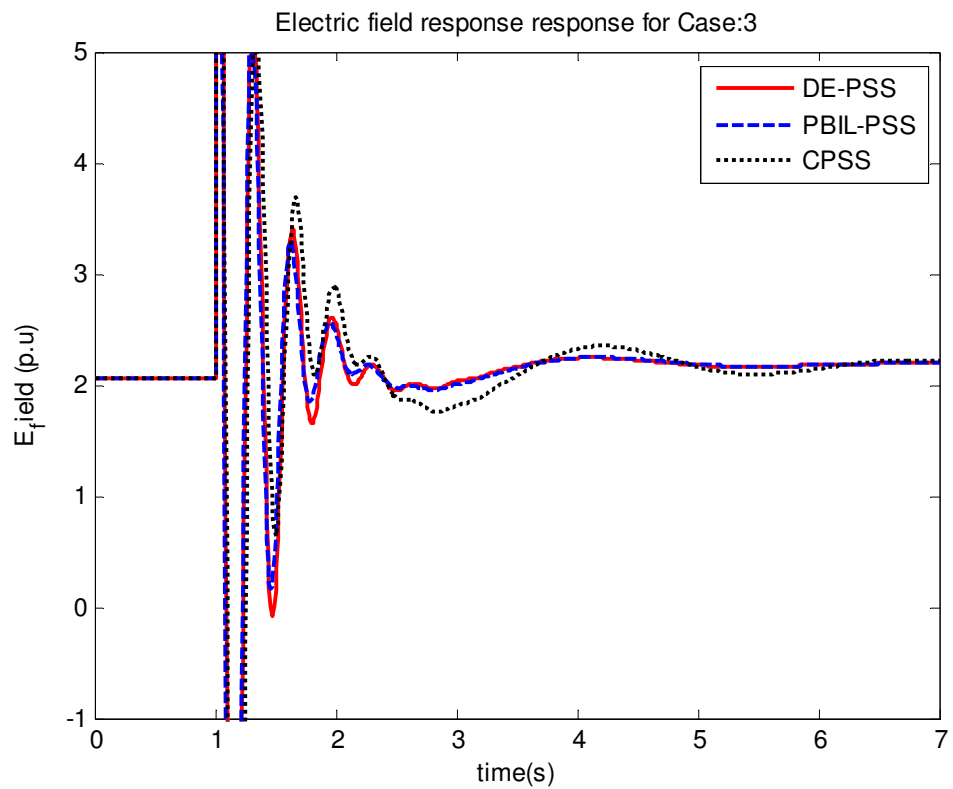


Figure 7.37: Electric field following a 3-phase fault for Case 3

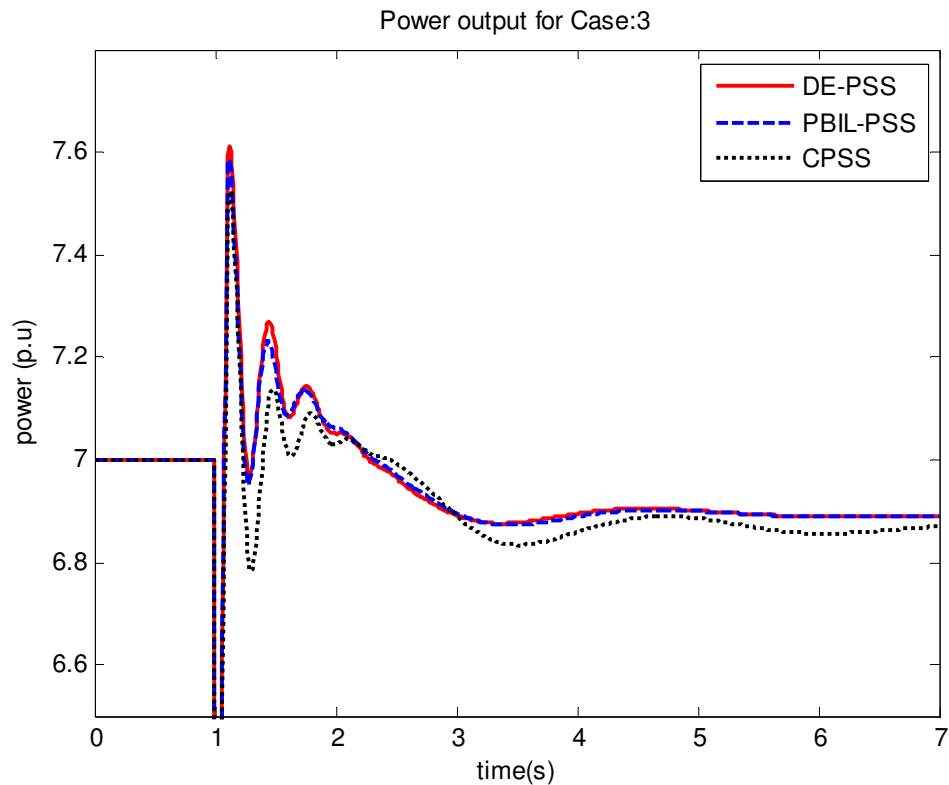


Figure 7.38: Active power output following 3-phase fault for Case 3

7.4.2.4 Case 4: 3-phase fault applied when 400MW are transmitted

As the system loading increases, the performance of the CPSS deteriorates as it can be observed in Figure 7.39 to Figure 7.42. The CPSS is increasingly slower to recover from the fault taking over 7 seconds to damp the oscillations. Both DE and PBIL based PSSs, on the other hand, display adequate and similar performance. Both systems settled just over 4 seconds for the rotor speed responses in Figure 7.40, the electric field in Figure 7.41 and the active power in Figure 7.42, whilst settling in just over 5 seconds for the terminal voltage in Figure 7.39.

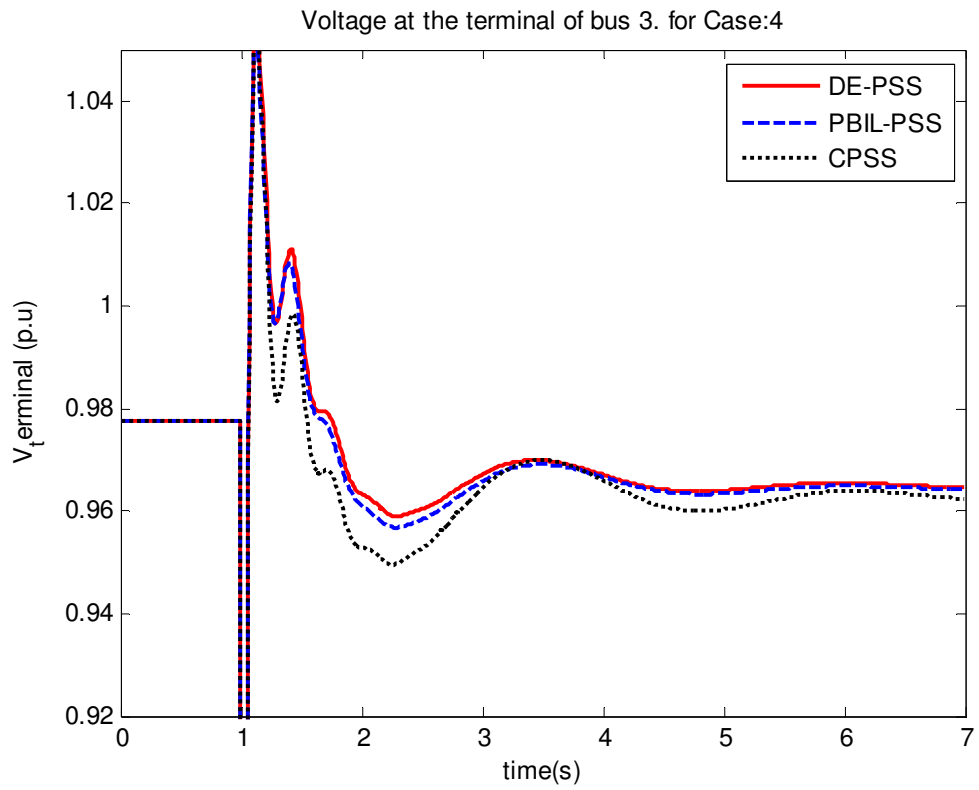


Figure 7.39: Terminal voltage response following a 3-phase fault for Case 4

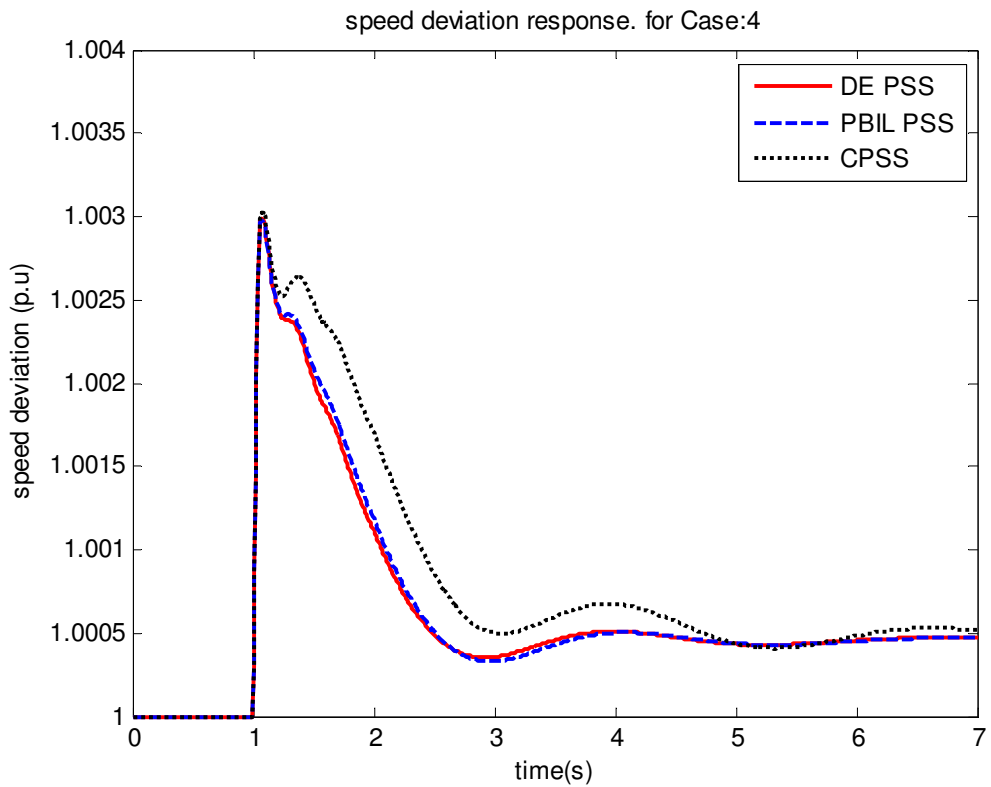


Figure 7.40: Rotor speed response following a 3-phase fault for Case 4

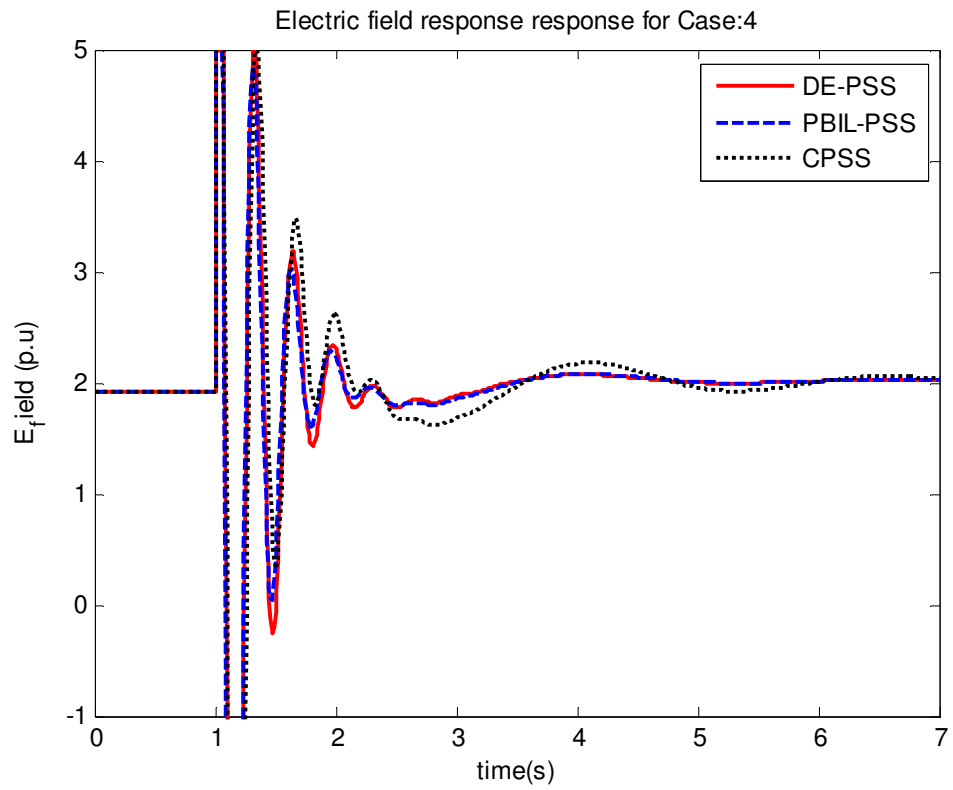


Figure 7.41: Electric field following a 3-phase fault for Case 4

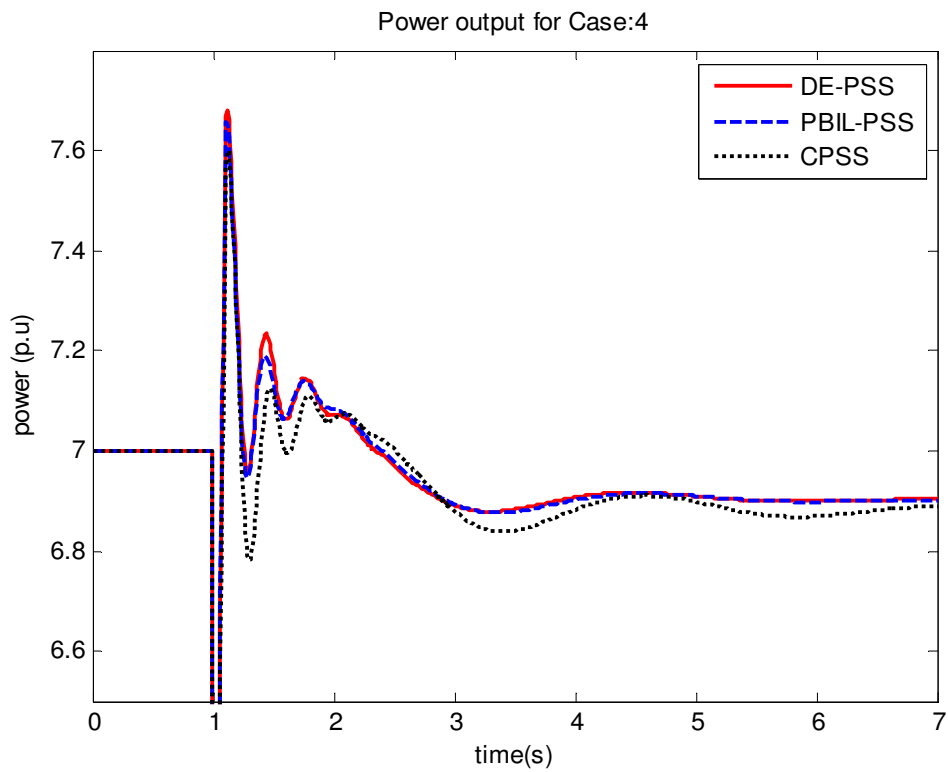


Figure 7.42: Active power output following 3-phase fault for Case 4

7.5 Summary

The application of two proposed Evolution Algorithms (DE and PBIL) for tuning the parameters of PSS has been tested for a Two-Area Multimachine system. The resulting PSSs were compared with those of the CPSS. The modal analysis shows that all the PSSs are capable of improving the dynamic stability of the system. In particular, DE-PSS performs better than PBIL-PSS and CPSS for all cases that have been discussed. These results have been validated in time domain simulations under small disturbance where DE-PSS outperformed both PBIL-PSS and CPSS with faster settling times and less overshoots and undershoots in some cases. As expected, the CPSS performs well for the nominal condition however degrades when the operating conditions change for both small and large disturbances. In transient stability, it was observed that DE displayed slightly higher overshoots than PBIL, which can be attributed to the slight higher gain.

Chapter 8

Application of Self-Adaptive Differential Evolution to Power System Stabilizer Design

In this chapter, a systematic analysis of DE's control parameters is conducted. The effects of the mutation factor, crossover probability, and population size on DE's performance in tuning PSSs are presented. The chapter is therefore divided into three sections. The first section investigates the inherence of the mutation factor and crossover probability on DE. In the second section, the population size is also investigated. The effects of increasing the population size are presented. As a result, the Self-Adaptive Differential Evolution is used to tune a power system stabilizer in SMIB and compared to the Differential Evolution (DE) in the third section.

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8.1 Introduction

Differential Evolution has been shown to be a simple yet powerful algorithm especially for optimally tuning Power System Stabilizers as in this thesis and real-world problems. However like other EAs, DE's performance is closely dependent on its intrinsic control parameters such as the mutation factor, the crossover probability and population size. Inappropriate choice of control parameter values may result in significant deterioration of the algorithm performance and reliability to effectively explore the search space for the global maximum or minimum. DE works generally well for many practical problems when the control parameters are set appropriately [51]. The authors in [18] and [65] have proposed some simple guidelines for the control parameter settings that ensure DE's optimal performance. These guidelines are mainly based on experimental studies and their applicability to other problems may not always yield satisfactory results because the control parameters of DE are dependent on problem characteristics and objective functions. Consequently, they must be adjusted. Very often, these parameters are obtained by trial-and-errors.

In the next section, the effect of the mutation factor ‘F’ and the crossover probability ‘CR’ on the PSS tuning are presented. An empirical rule for choosing these parameters is established based on the results obtained in this research. Then, a self-adaptive scheme is applied to the PSS tuning to solve the tedious process of finding the optimal control parameters. For this section, since we want to evaluate the performance of the self-adaptive DE to the conventional DE, CPSS and PBIL are not included. Only DE and self-adaptive DE are compared.

8.2 Effects of Mutation F and Crossover Probability Cr on PSS Tuning

The sensitivity of differential evolution to its intrinsic control parameters has been a challenge to practitioners since its inception. Quantifying the sensitivity of differential evolution to its control parameters and subsequently extracting empirical rules for choosing optimal intrinsic control parameters for future applications is of great importance [45].

DE’s ability to find the global maximum is mainly dependent on the mutation and crossover processes. As mentioned in chapter 3, the differential mutation allows DE to explore the search space for the global maximum or minimum. This process is controlled by the mutation scale factor F , a positive real number in the range of 0 to 2. ‘ F ’ controls the rate at which the population evolves. Whereas the crossover ensures that the diversity of population is kept to avoid premature convergence to local optima. Hence this process is directly dependent on the value of ‘CR’.

Figure 8.1 and 8.2 below show the effect of F and CR on the algorithm ability to maximize the lowest damping ratio of an SMIB system. In order to determine the optimal F & CR, one parameter is kept constant whilst varying the other, alternatively. In Figure 8.1, CR was kept constant to 0.9 whilst F was varied. DE converges to lowest damping ratios of about 12%, 14%, 16%, 21%, 22% and 20% when F is set to 0.1, 0.3, 0.5, 0.75, 0.9 and 1.0, respectively. It can be observed that at lower value of F, DE’s performance is restricted and unable to explore the search space adequately causing the algorithm to converge fast to a poor solution (possibly local optima). Hence increasing F guarantees a

good search. However as it can be seen in Figure 8.1, DE performance is negatively affected when $F = 1.0$ (but it's still better than $F=0.1$ to 0.5). It can therefore be deduced that values of F close to 0.9 are suitable for PSS tuning. It allows the algorithm to search for the global maximum.

In Figure 8.2 above, F is kept constant at 0.9 and CR is varied. Results show that lower values of CR , for example 0.1 , cause the search to converge to local minima despite the diversity. Hence increasing CR increases the diversity of the population which subsequently increases the chance of finding the global maxima. Therefore, it is deduced that for PSS tuning, CR works best with values ranging between $[0.9 - 0.95]$. Hence, it is not recommended to set $CR = 1.0$ due to the fact that it negatively affects the performance of DE by reducing its convergence rate when applied to PSS tuning.

This has led to the development of self-adaptive DE discussed in the section 8.4. The next section investigates the impact of population size on the optimization.

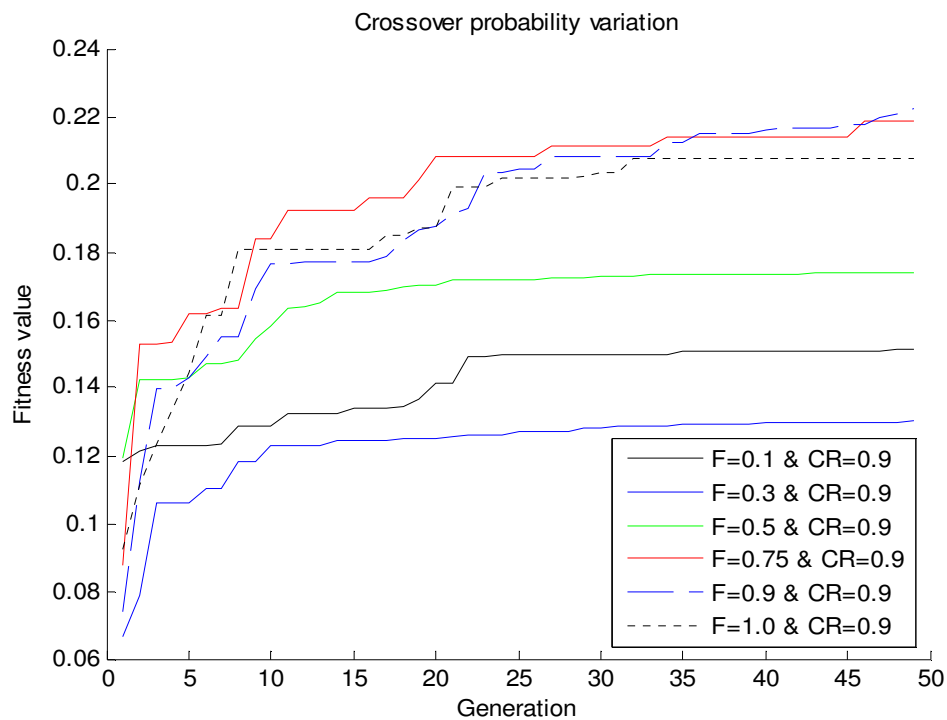


Figure 8.1: Effect of CR probability on DE performance

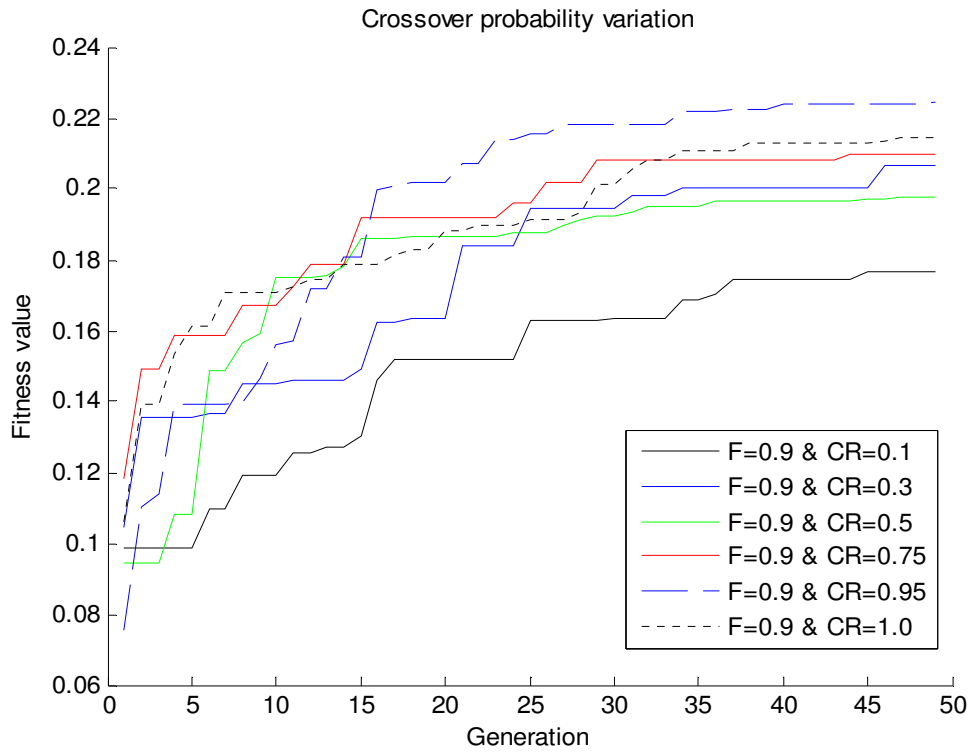


Figure 8.2: Effect of CR probability on DE performance

8.3 Effect of population size on PSS tuning

The effect of population size parameter on the performance of DE has not been given much importance and is often chosen as a constant value. Therefore, in this section, investigations on its effects when tuning PSSs are conducted.

A population is constituted of N_p individuals that represent possible solutions to a problem domain. For many years, the question as to how to choose an adequate population size for a particular domain has puzzled practitioners. If the population size is small, the algorithm may converge fast; but the probability of premature convergence and stagnation may be higher. Stagnation can be defined as a phenomenon whereby the population remains diverse and unconverged, whilst the optimization process no longer progresses [66]. Stagnation may occur virtually without any obvious reasons, whereas in premature convergence the population loses its diversity [63], [66]. To overcome this

drawback, larger population may be used. However large population may require more computational effort.

The author in [16] recommended that a population between $7.D$ and $10.D$ be used for a good and efficient performance. D is the dimensionality of the problem also known as parameters to be optimized. However, like the other control parameters, the population size is problem dependent. With small dimensional problems, the population size bears slight effects; however when the dimension is increased the effect of the population size is more noticeable [16]. In that regards, the authors in [63] conducted an empirical study of population size on DE algorithm. The results showed significant influence on the performance of DE.

The investigations in this research are carried out in a Two-Area Multi Machine system whereby the dimensionality of the problem is 10 parameters. Different population sizes (D , $3.D$, $5.D$, $7.D$, $10.D$, $15.D$ and $20.D$) are considered in this study. Because the optimization is a random based operation, each set of population was run 10 times, independently. The number of generation is set to 200. The analysis is based on the maximum damping obtained from the 10 runs, the 'Mean' which indicate the average of the maximized best damping ratios, the standard deviation, and time domain analysis.

Table 8.1 below summarizes the results obtained from the simulations (Best damping, mean and standard deviation) for different population sizes. It is observed that when the population is increased, the converged value also increases. This is due to a better exploration capability. With a population of $15.D$ (150), DE converges to the best overall damping ratio and has the highest average value. DE's worst performance is recorded when the population is set to D (10). This is expected since there is not much diversity within the population which leads to a premature convergence and possibly stagnation. Hence, this suggests that DE is caught into a local optimum. When DE is set to $10D$ (100), it is observed that the standard deviation is at its highest. Hence a wider variation in the maximum converged values is noticed. When the population is set to 50, DE converged to a best damping of 26.94 %, and an average of 23%. It is also observed that the standard deviation is the smallest. This suggests that the results obtained over the 10

runs are close with each other, therefore consistent. DE (50) converged to the second highest average value.

Table 8.1: Experimental results of DE when varying the population

Population size	Best damping	Mean	Std. Dev
10	0.178	0.161	0.0163
30	0.24390	0.22677	0.022362
50	0.26940	0.230126	0.0110
70	0.259877	0.2214	0.023452
100	0.26714	0.22537	0.032241
150	0.27400	0.24582	0.028197
200	0.26030	0.22307	0.026662

Figure 8.3 below shows the fitness curve of different population sizes. It is observed that when the population size is small (10 & 30), the diversity is lost early, and DE prematurely converges, possibly to a local optimum. However, as the population is increased (50 to 100), DE is able to further explore the search space and converge to better damping ratios. When the population is further increased (150 & 200), the diversity within the population also increases, hence the convergence becomes slower. In this case, one of the solutions would be to increase the number of generation to allow the algorithm to converge to an optimum value.

Population size is often cited as a source of increased diversity [67], [71], [72], but it would be expected to have different impacts on searches with low versus high values of CR given the differing ways in which they explore solution space. It was shown in [73] that the performance of DE improves whilst increasing CR settings and decreasing the population size. This is due to the fact that the acceptance rate of new solutions is always

inversely related when $CR = 0.9$, yet on several problems is positively related when $CR = 0.5$. This suggests that a larger population can promote improved exploration when CR is low, however, it means that the resulting gradual search has fewer opportunities to improve each member of the population. In other words, when CR is high, a large population works against convergence and prolongs the time during which large difference vectors are produced. This results in a greater number of inappropriately large moves due to the differential mutation F being used to try to improve result. Consequently, reducing the value of F may also allow large populations to converge.

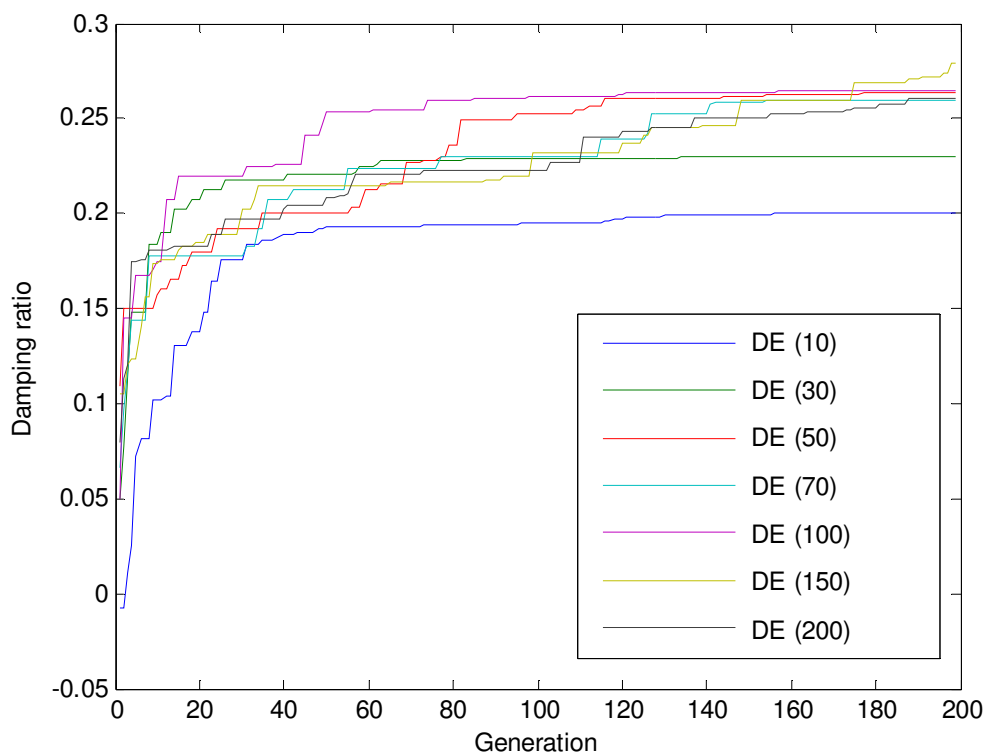


Figure 8.3: Fitness curve for different population size

In [64], the author developed a self-adaptive population algorithm to address the drawbacks observed. However this method was not investigated in this thesis. In the next section, the population was kept constant whereas CR and F were self-adapted when applied to the tuning of the PSS.

8.4 Self-Adaptive Differential Evolution

The performance of DE is affected by its control parameters which are subsequently dependent on the characteristics of the objective function. Hence, inadequate parameter settings may degrade the convergence of the algorithm. Often, a tedious trial-and-error approach is used to determine these parameters [51].

Recently, adaptive or self-adaptive mechanisms have been introduced in [19], [67]-[69], to dynamically and automatically update the parameters to fit the objective function of the optimization problem without user interaction or trial-and-error. The self-adaptive algorithms have shown faster and more reliable convergence performance than the Classic DE for many problems [19], [51], [67]-[69].

Most adaptive algorithms have been developed based on the Classic DE with strategy DE/rand/1, [19], mentioned in Chapter 3 which is known to be robust but less efficient in terms of convergence rate. However not many methods have been developed based on the DE/current-to-best strategy due to the fact that they are usually less reliable and may lead to premature convergence.

Some adaptive strategies have been briefly reviewed in the subsequent paragraphs.

FADE: The Fuzzy Adaptive Differential Evolution (FADE), introduced by Liu and Lampinen in [67], is a fuzzy logic based controllers to adapt DE parameters F_i and CR_i for the mutation and crossover operations whereas the population is kept fixed.

SaDE: The Self-Adaptive DE (SaDE) was proposed in [69]. Two mutation strategies, 'DE/rand/1' (strategy 1) and 'DE/current-to-best/2' (strategy 2), may be used in this algorithm. The self-adaptive strategy consists of applying a probability p_i to first determine the mutation strategy (1 or 2) to be used to generate a mutation vector. This probability is updated after 50 generations. In this strategy, F_i is independently generated for each generation according to a normal distribution with a mean of 0.5 and standard deviation 0.3, defined as follows

$$F_i = randn(0.5,0.3) \quad (8.1)$$

The crossover probabilities CR_i are independently generated according to a normal distribution with mean CR_m and standard deviation 0.1, defined as follows,

$$CR_i = randn(CR_m, 0.1) \quad (8.2)$$

where CR_m is the mean of every 25 successful CR values in the last 25 generations.

jDE: Proposed by Brest *et al.* [19], this strategy is based on DE/rand/1. Similar to other schemes, jDE fixes the population size during the optimization whilst adapting the control parameters F_i and CR_i associated with each individual. In other words, these control parameters are adjusted by means of evolution (see Figure 8.4). Both of them are encoded in the individual levels. The better values of these (encoded) control parameters lead to better individuals which, in turn, are more likely to survive and produce offspring and, hence, propagate these better parameter values [68], [70], [74].

$\vec{x}_{1,G}$	$F_{1,G}$	$CR_{1,G}$
$\vec{x}_{2,G}$	$F_{2,G}$	$CR_{2,G}$
...
$\vec{x}_{NP,G}$	$F_{NP,G}$	$CR_{NP,G}$

Figure 8.4: Self-Adapting encoding aspect

The initialization process sets $F_i = 0.5$ and $CR_i = 0.9$ for each individual. jDE regenerates (with probabilities $\tau_1 = \tau_2 = 0.1$ at each generation) new values for F_i and CR_i according to uniform distributions on $[0.1, 1]$ and $[0, 1]$, respectively.

$$F_{i,g+1} = \begin{cases} F_i + rand_1 * F_u & \text{if } rand_2 < \tau_1 \\ X_{i,g} & \text{otherwise} \end{cases} \quad (8.3)$$

$$CR_{i,g+1} = \begin{cases} rand_3 & \text{if } rand_4 < \tau_2 \\ CR_{i,g} & \text{otherwise} \end{cases} \quad (8.4)$$

where $rand_j, j = 1, 2, 3, 4$, are uniform random values on $[0, 1]$, and $\tau_1 = \tau_2 = 0.1$ represent the probabilities to adjust the control parameters. The newly generated parameter values are used in the mutation and crossover operations to create the corresponding offspring vectors and will replace the previous parameter values if the offspring survive in the selection. It is believed that better parameter values tend to generate individuals which are more likely to survive, and able to go into the next generation.

In this work, the adaptive scheme used is based on the mutation strategy DE/rand/2 which is as follows

$$V_{i,g} = X_{r0,g} + F \cdot (X_{r1,g} - X_{r2,g}) + F \cdot (X_{r3,g} - X_{r4,g}) \quad (8.5)$$

Hence, each individual is encoded with two values of F_i and CR_i parameters as illustrated in Figure 8.5. These control parameters are generated based on (8.3) and (8.4). This approach allows robustness and good convergence. This self-adaptive strategy is applied to tune the PSS and compared with DE in the subsequent sections.

$\vec{x}_{1,G}$	$F_{1,G}^1$	$CR_{1,G}^1$	$F_{1,G}^2$	$CR_{1,G}^2$
$\vec{x}_{2,G}$	$F_{2,G}^1$	$CR_{2,G}^1$	$F_{2,G}^2$	$CR_{2,G}^2$
...
$\vec{x}_{NP,G}$	$F_{NP,G}^1$	$CR_{NP,G}^1$	$F_{NP,G}^2$	$CR_{NP,G}^2$

Figure 8.5: Self-adaptive encoding aspect based on DE/rand/2

8.4.1 Application to PSS tuning: Modal analysis

In this section, jDE has been applied to tune PSS in an SMIB system described in chapter 5. The resulting PSS is compared to the classical DE (CDE) which is based on DE/rand/1 mutation strategy.

Because DE search engine is based on the mutation process, a relatively small population is often used to ensure an optimal performance. Thus, DE control parameters were configured as follows (Figure 8.6 below)

	<u>DE PARAMETERS</u> Population: 30 Generation: 100 Mutation F: 0.9 Crossover CR: 0.9 Termination: Max gen		<u>jDE PARAMETERS</u> Population: 30 Generation: 100 Mutation F: Adaptive Crossover CR: Adaptive Termination: Max gen	
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Figure 8.6: Control Parameter settings

The PSSs were designed over 5 operating conditions. Table 8.2 shows the PSSs' parameters obtained after optimization using jDE and CDE whereas Table 8.3 shows the case considered for the design and testing of the PSSs, with the open-loop and closed-loop eigenvalues and the respective damping ratio in brackets.

The results show that the open-loop system is poorly damped for all cases whereas both PSSs perform adequately. However, in Case 1, jDE has a slightly higher damping ratio of 43.4% whereas CDE has 40.8%. Furthermore there is a noticeable decrease in the frequency of oscillation on the electromechanical modes associated with CDE-PSS. In case 2, a slight drop in damping ratio is observed on the system equipped with CDE-PSS whereas jDE-PSS damping remains constant at 43%. As the tie-lines are further weakened in case 3, CDE-PSS has a damping of 33%. jDE on the other hand has 44.6%. In case 4, when the system is heavily loaded combined with weak transmission lines, jDE-PSS has a damping of 45 % whilst CDE-PSS has 27%. This agrees with the expected performances of CDE with regards to the algorithm performance.

Table 8.2: PSS parameters

	K	T1	T2	T3	T4
CDE-PSS	17.2	0.0102	0.154	4.839	0.272
jDE-PSS	18.9	4.64	1.67	3.21	1.50

Table 8.3: System Open-loop and Closed-loop eigenvalues

Case	Active Power (pu)	Line reactance (pu)	Open-Loop Eigenvalues	CDE-PSS	jDE-PSS
Case 1	1.0	0.3	-0.268±j4.457 (0.060)	-1.521±j3.405 (0.408)	-1.920±j3.979 (0.434)
Case 2	1.0	0.5	-0.118±j3.781 (0.0483)	-1.134±j2.735 (0.383)	-1.566±j3.232 (0.436)
Case 3	1.0	0.7	-0.133±j3.311 (0.0402)	-0.832±j2.319 (0.330)	-1.341±j2.689 (0.446)
Case 4	1.0	0.9	-0.0997±j2.947 (0.0338)	-0.487±j1.694 (0.276)	-1.161±j2.249 (0.458)

8.4.2 Time Domain Step Response Analysis

The small disturbance simulations have been performed for all four cases previously mentioned by applying 10% step change in the reference voltage. The step response for speed deviation of the generator is presented from Figure 8.7 to Figure 8.10.

Figure 8.7 shows the responses of the rotor speed deviations for Case 1. It can be seen that all controllers are able to damp the oscillations and improve the system dynamic stability; jDE-PSS settles within 2.5 seconds whereas CDE-PSS settles around 3 seconds.

In Figure 8.8, it is observed that the speed deviations jDE-PSS displays better performance than that of CDE-PSS with settling time of 3 seconds and 5 seconds, respectively.

Figure 8.9 shows the responses of the rotor speed deviations for case 3 whereby CDE-PSS performance is deteriorating whereas jDE-PSS outperforms its counterpart. Despite the faster settling time, jDE has slightly bigger undershoots which can be attributed to the overcompensation of the stabilizer.

Figure 8.10 shows the speed response of the system for case 4. In this particular case, the system is heavily loaded with weak tie-line. In this scenario, as expected, jDE has the fastest response with settling time of about 4 seconds; whereas CDE-PSS has slower decaying oscillations with a long settling time of about 10 seconds. It is also observed that CDE has the largest overshoots whereas jDE has the biggest undershoots.

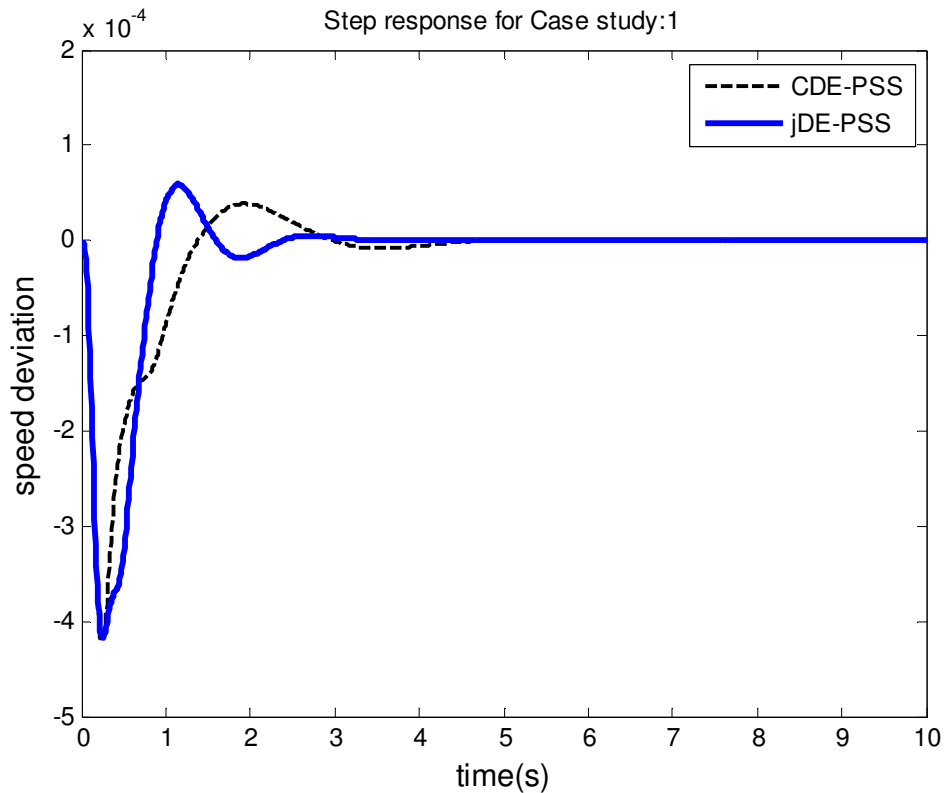


Figure 8.7: Step-response for Case 1

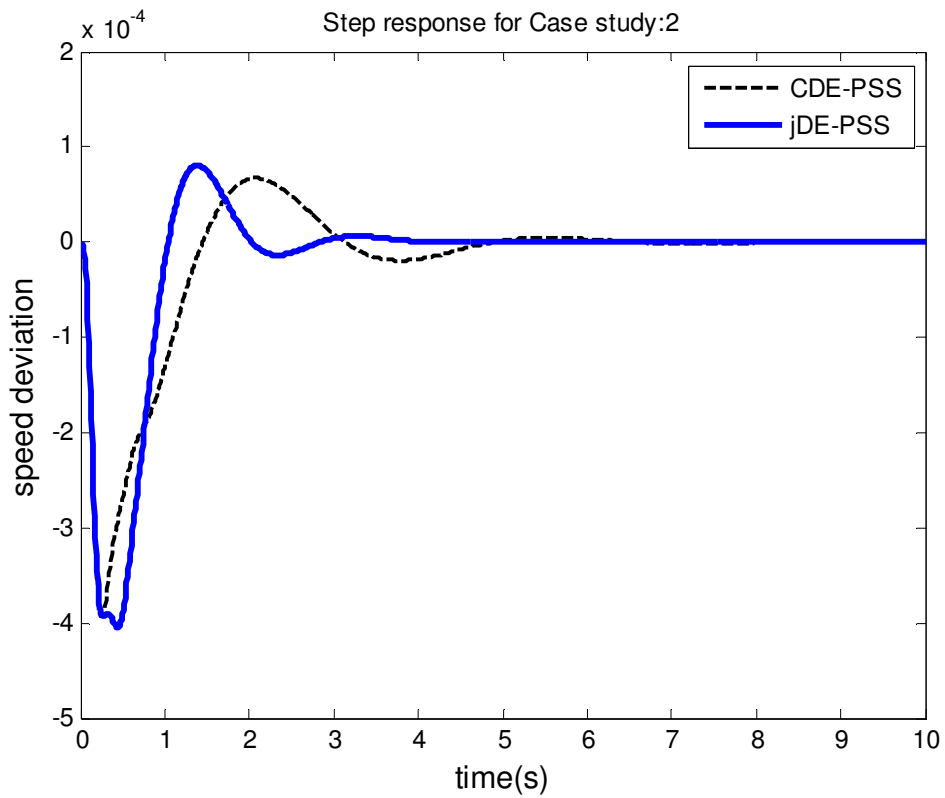


Figure 8.8: Step-response for Case 2

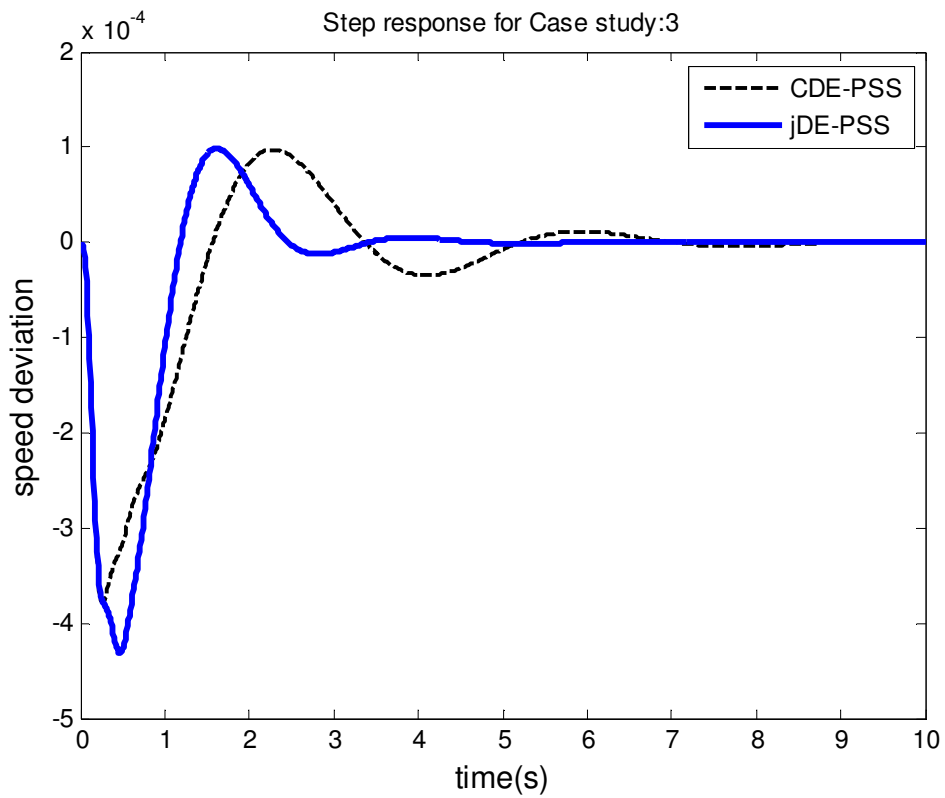


Figure 8.9: Step-response for Case 3

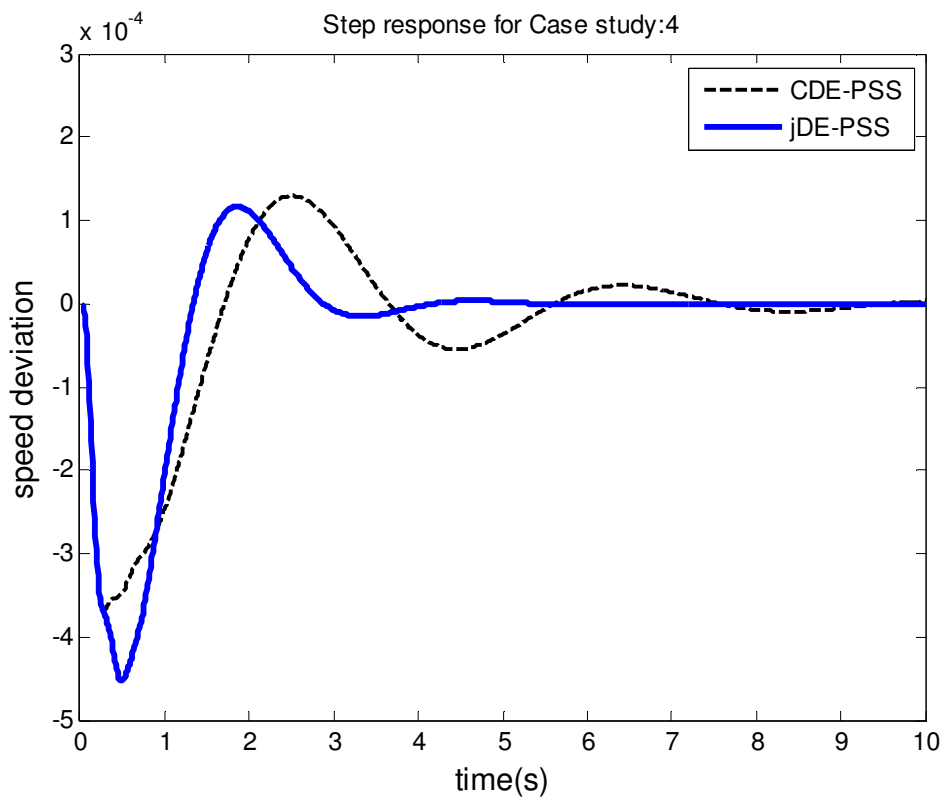


Figure 8.10: Step-response for Case 4

8.4.3 Large disturbances

The robustness of the PSS is further assessed when subjected to severe faults. The fault consists of submitting the system to a 5-cycles 3-phase fault applied after 1 second on line 1 (Figure 5.6). The fault was cleared by disconnecting the line.

Transient Stability was performed for Case 1 to 3. Case 4 was not considered because of the inability of the system to regain stability after clearing the fault. This may be attributed the high angle associated with the power (more than 90°).

8.4.3.1 Case 1: Three-phase fault applied the nominal system

Figure 8.11 to Figure 8.15 are the responses of the rotor speed, terminal voltage, the field voltage and active power following a three-phase short circuit fault on line 1 for the nominal operating condition. Both PSSs perform adequately and rapidly returned the system to steady state operation after the fault is cleared. Despite the rapid response of jDE-PSS, the system exhibits slightly larger overshoots and undershoots for the rotor angle and speed response whereas for the terminal voltage, electric field and power output, CDE-PSS has slightly high overshoots. In terms of settling time, both systems settle around 5 seconds, however jDE is observed to be slightly faster.

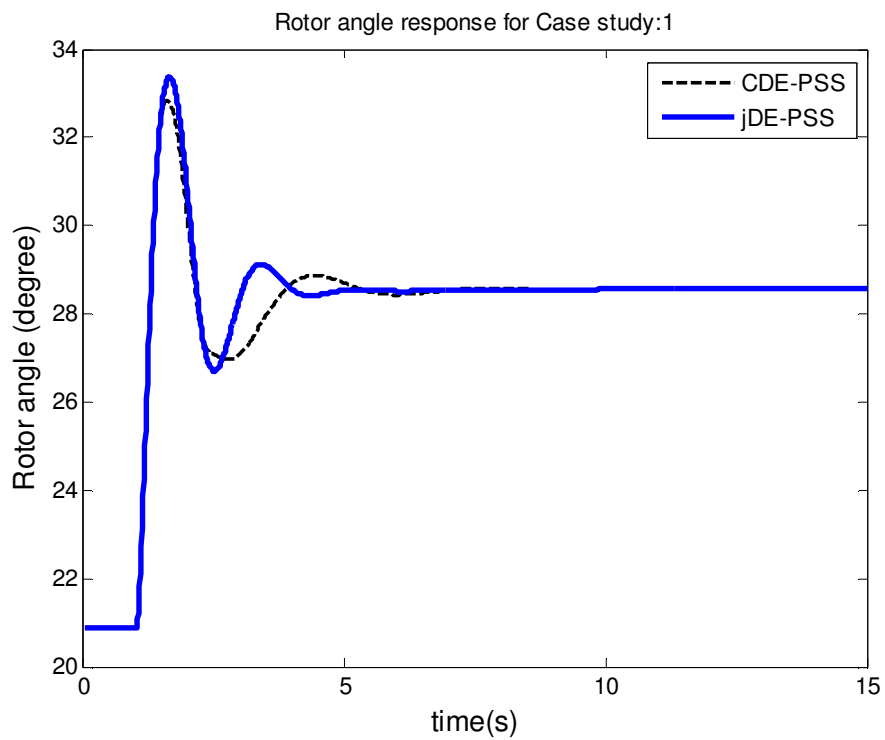


Figure 8.11: Rotor angle following 3-phase fault for Case 1

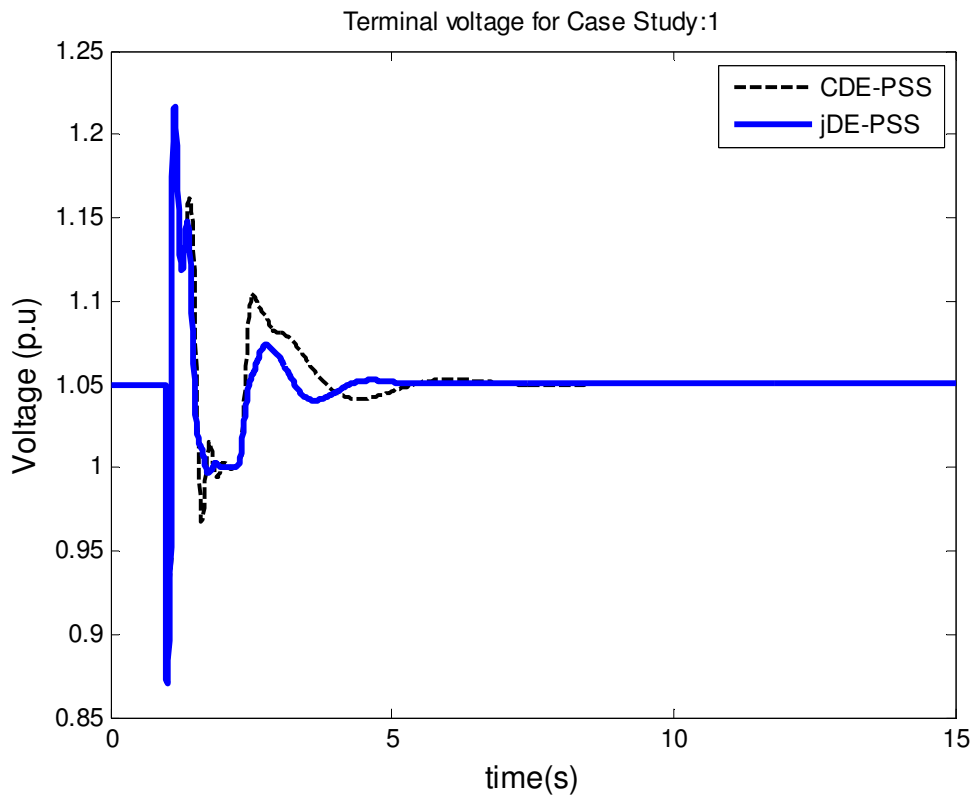


Figure 8.12: Terminal voltage following 3-phase fault for Case 1

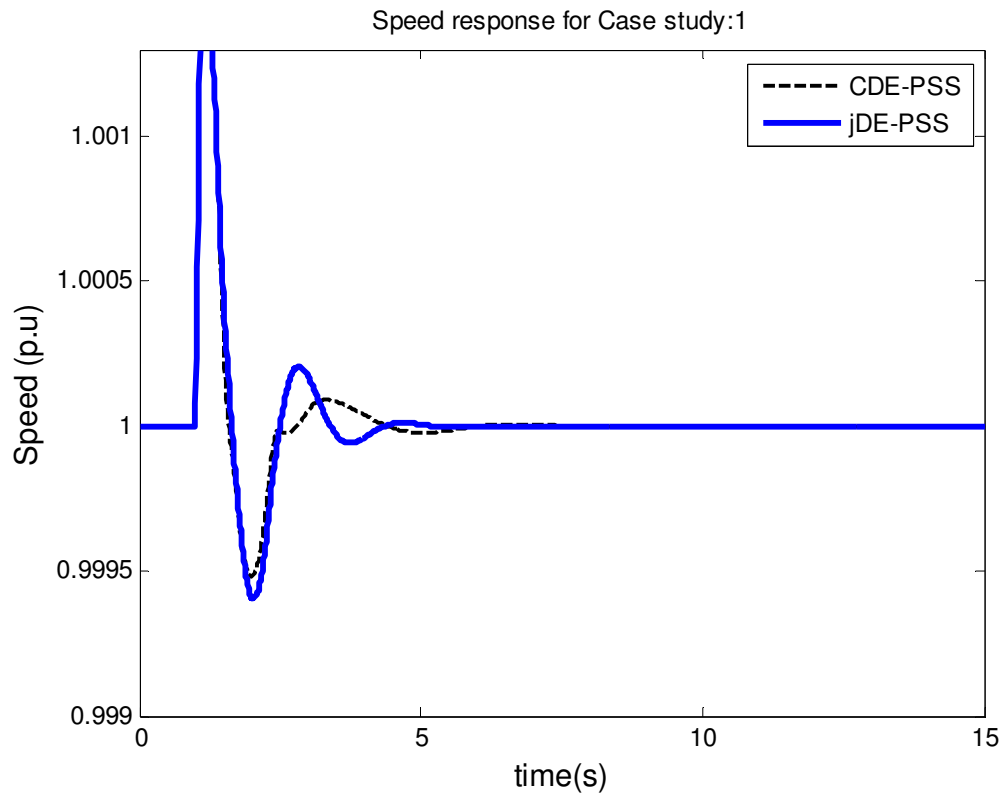


Figure 8.13: Speed response following 3-phase fault for Case 1

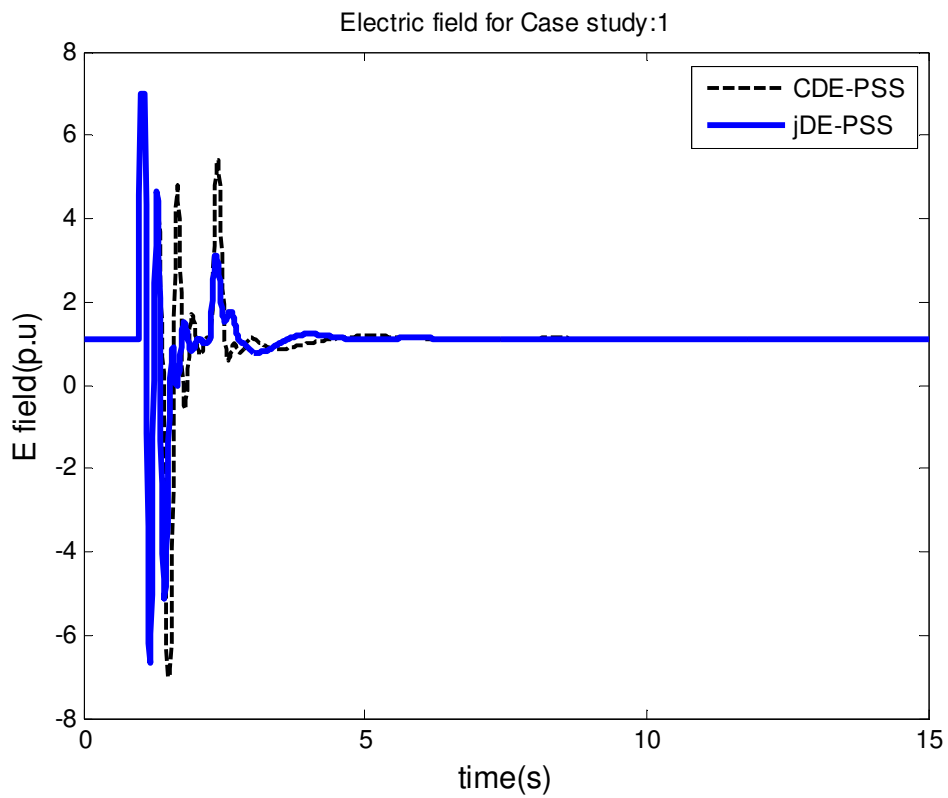


Figure 8.14: Electric field voltage following 3-phase fault for Case 1

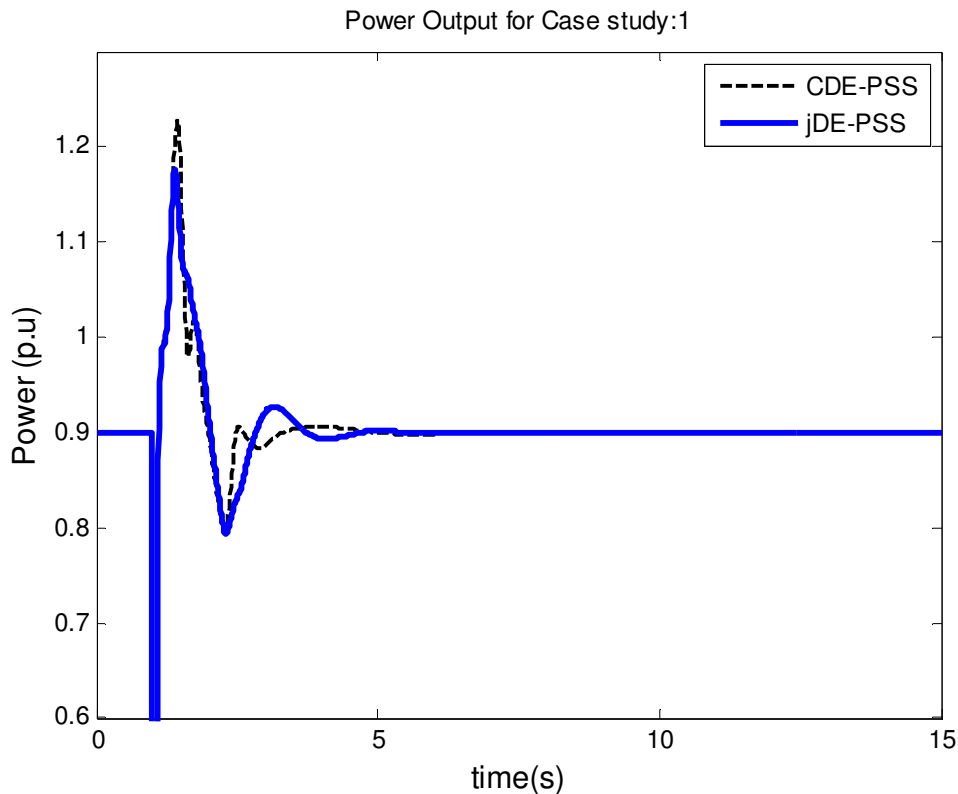


Figure 8.15: Active power output following 3-phase fault for Case 1

8.4.3.2 Case 2: 3-phase fault applied to case 2

The responses of the rotor angle, the terminal voltage at bus 3, the rotor speed, the field voltage and active power are shown in Figure 8.16 to Figure 8.20, respectively. In this case the difference in performance between the two systems is evident. jDE-PSS has a faster response with less overshoots and undershoots for the rotor angle, terminal voltage, speed, field voltage and power. Furthermore, the system has a settling time of around 5 seconds, whereas CDE-PSS settled around 7 seconds for the electric field and power, and 9 seconds for the rotor angle, speed and terminal voltage. The overshoots observed in CDE-PSS may be attributed to under compensation which in turn is a result of inadequate tuning of the intrinsic control parameters.

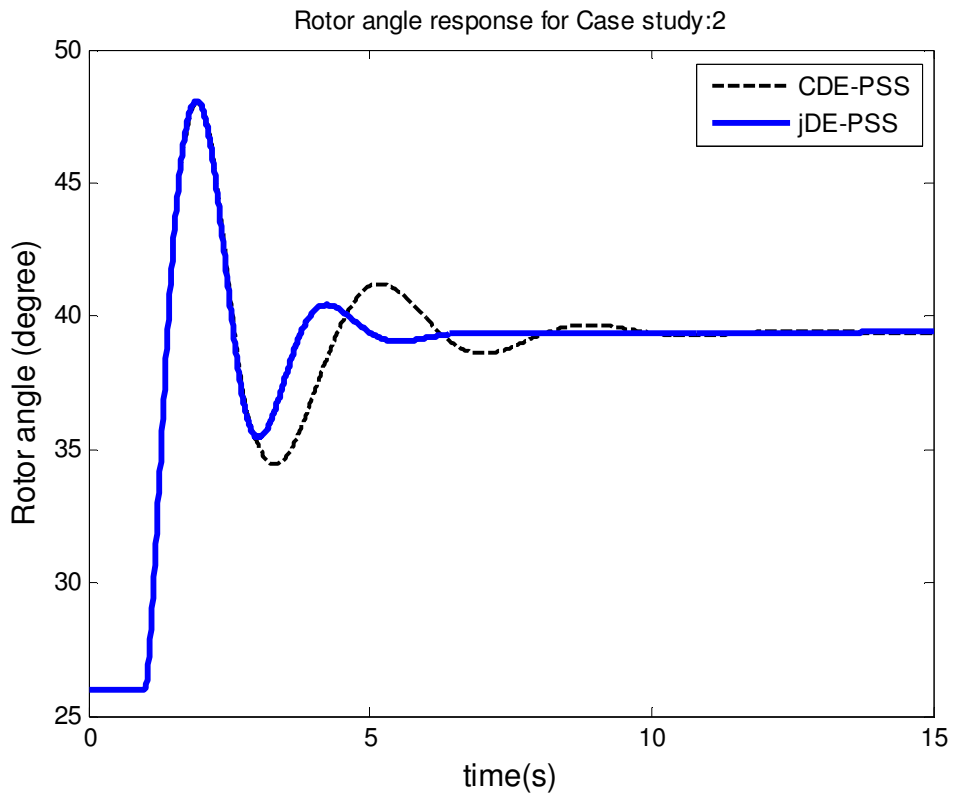


Figure 8.16: Rotor angle following 3-phase fault for Case 2

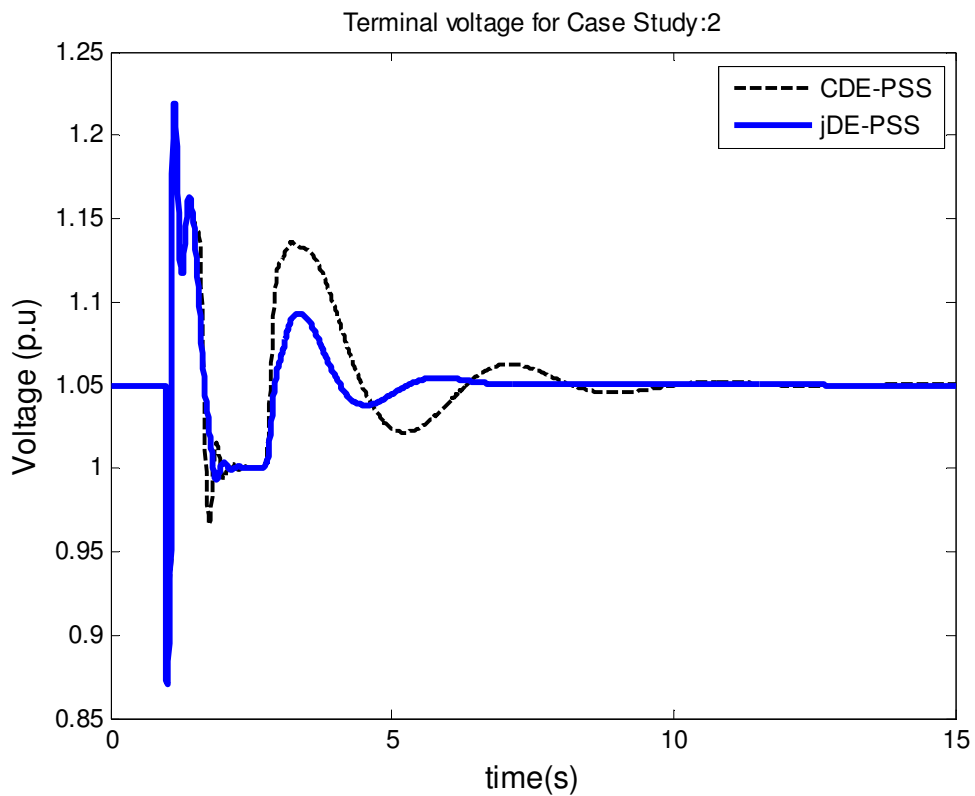


Figure 8.17: Terminal following 3-phase fault for Case 2

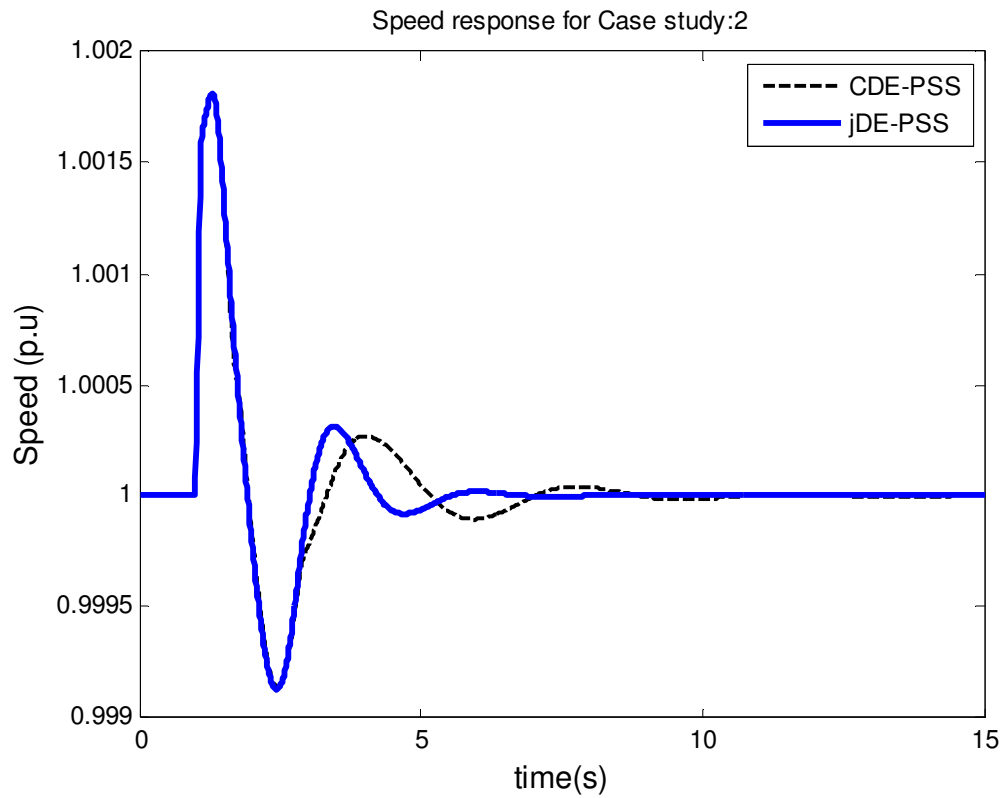


Figure 8.18: Speed response following 3-phase fault for Case 2

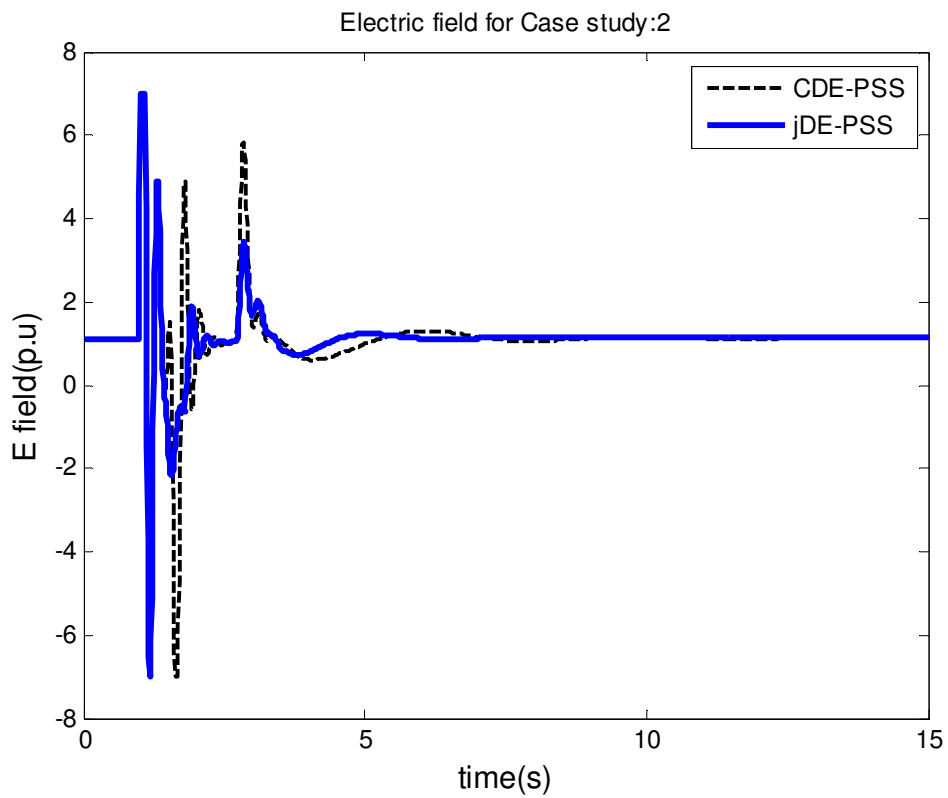


Figure 8.19: Electric field voltage following 3-phase fault for Case 2

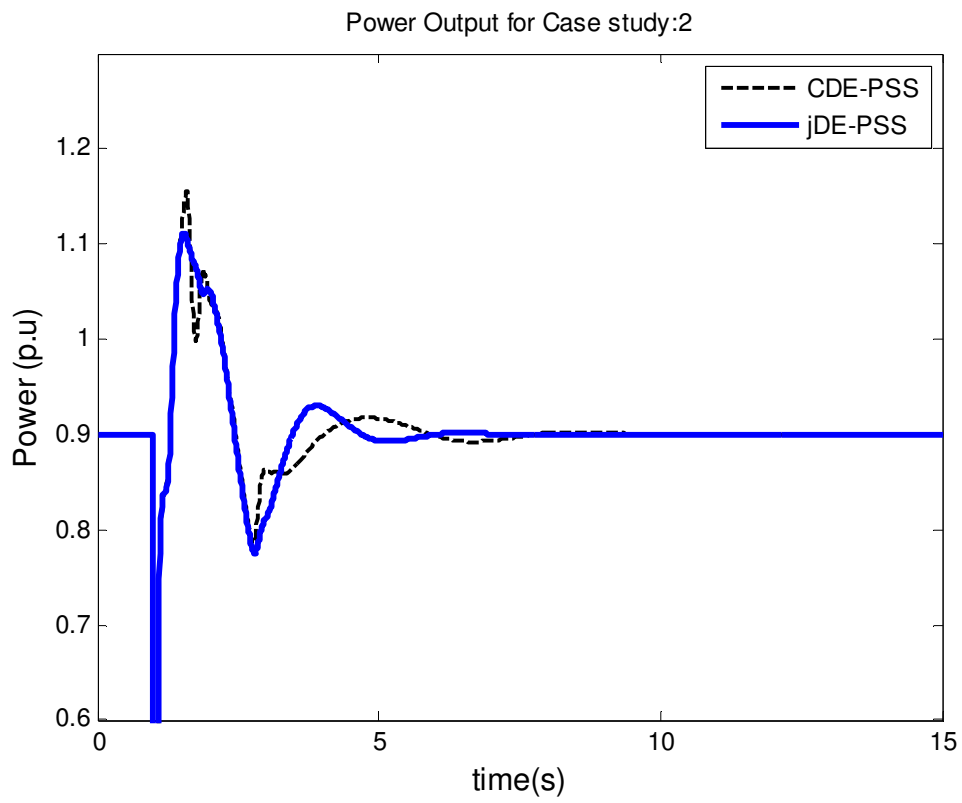


Figure 8.20: Active power output following 3-phase fault for Case 2

8.4.3.3 Case 3: 3-phase fault applied for case 3

The Transient responses of the terminal voltage at bus 3, the rotor speed, the field voltage and active power are shown in Figure 8.21 to Figure 8.25, respectively. In a similar trend to the previous case, CDE-PSS is more oscillatory with larger overshoots and undershoots. The resulting system settles around 13 seconds whereas jDE-PSS in around 7 seconds.

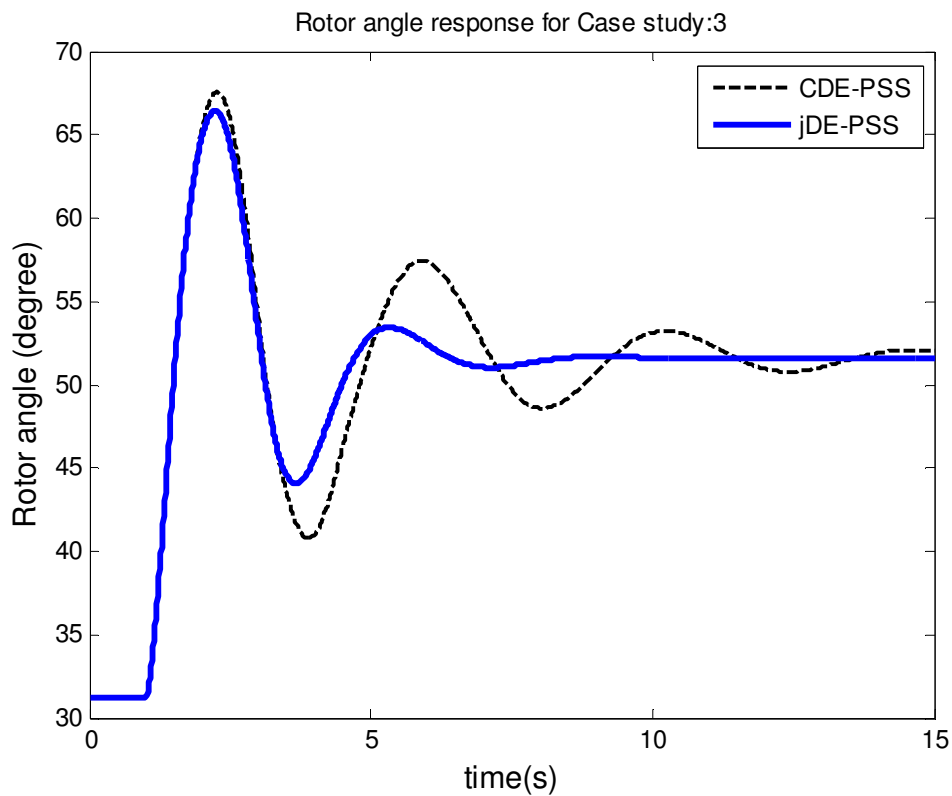


Figure 8.21: Rotor angle following 3-phase fault for Case 3

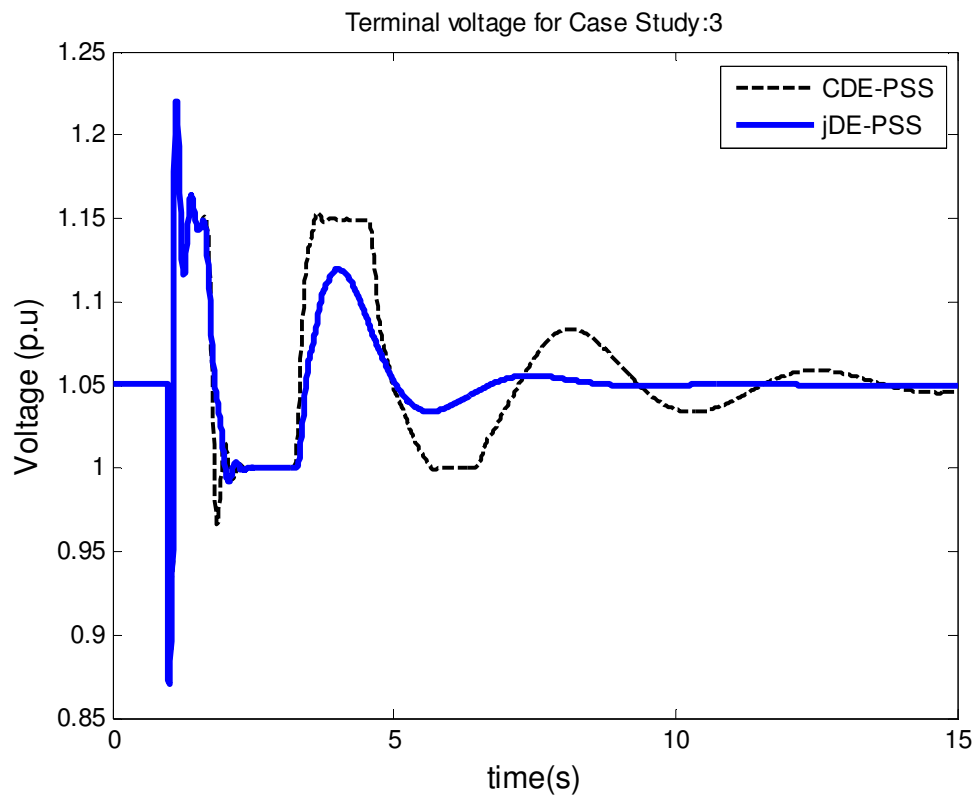


Figure 8.22: Terminal voltage following 3-phase fault for Case 3

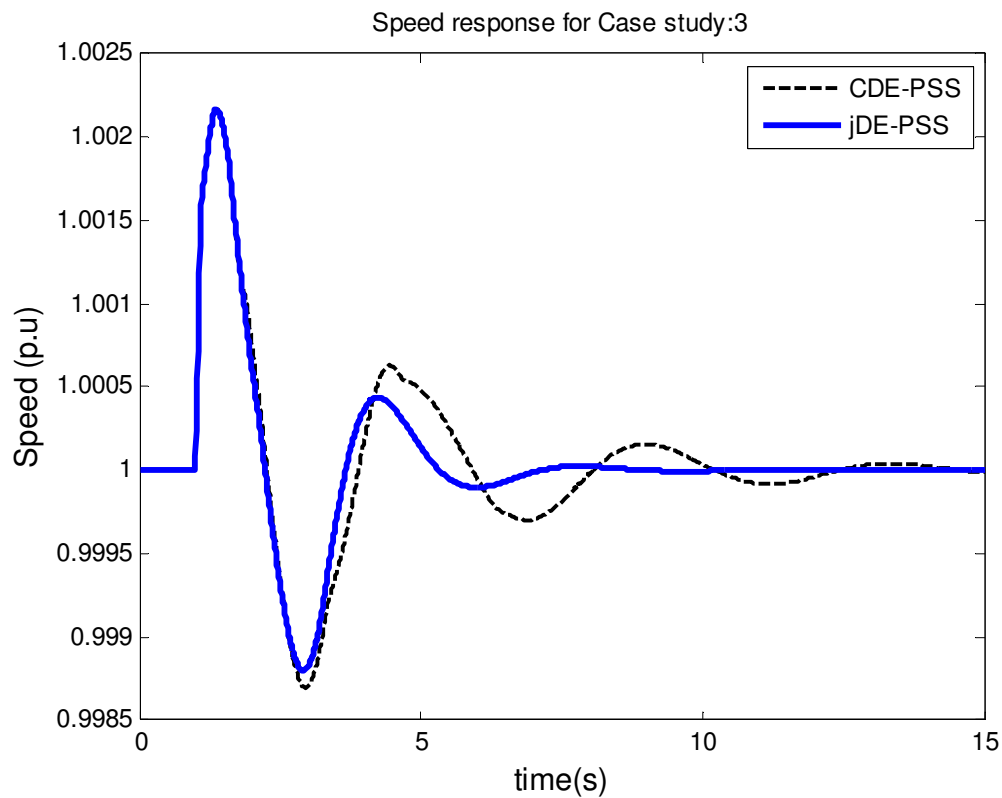


Figure 8.23: Speed response following 3-phase fault for Case 3

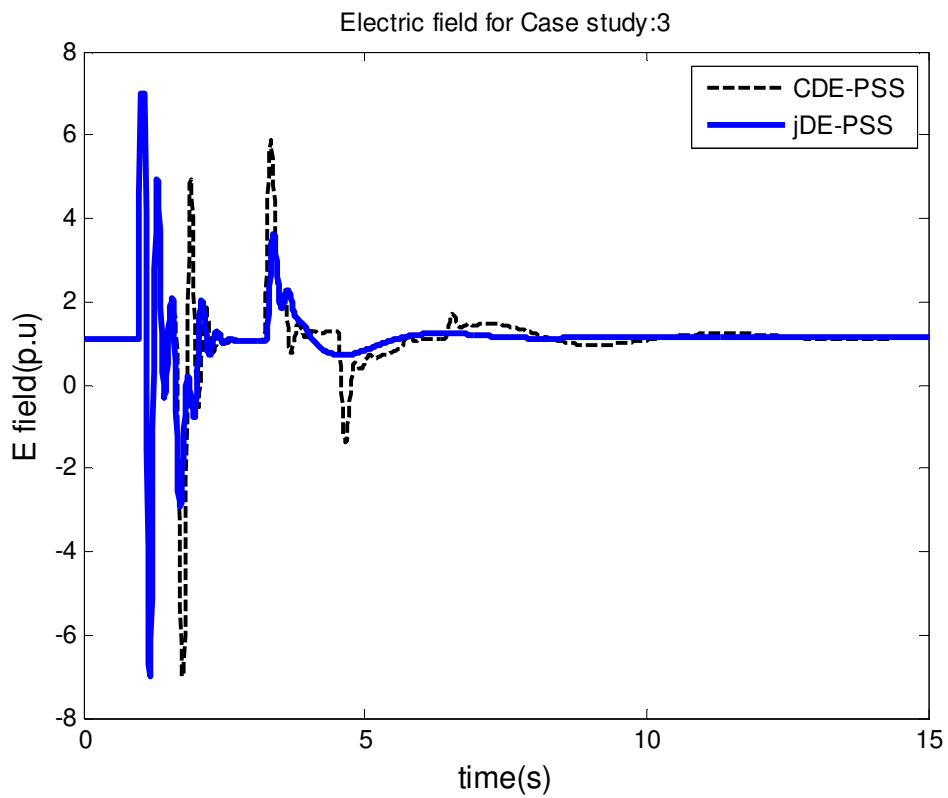


Figure 8.24: Electric field voltage following 3-phase fault for Case 3

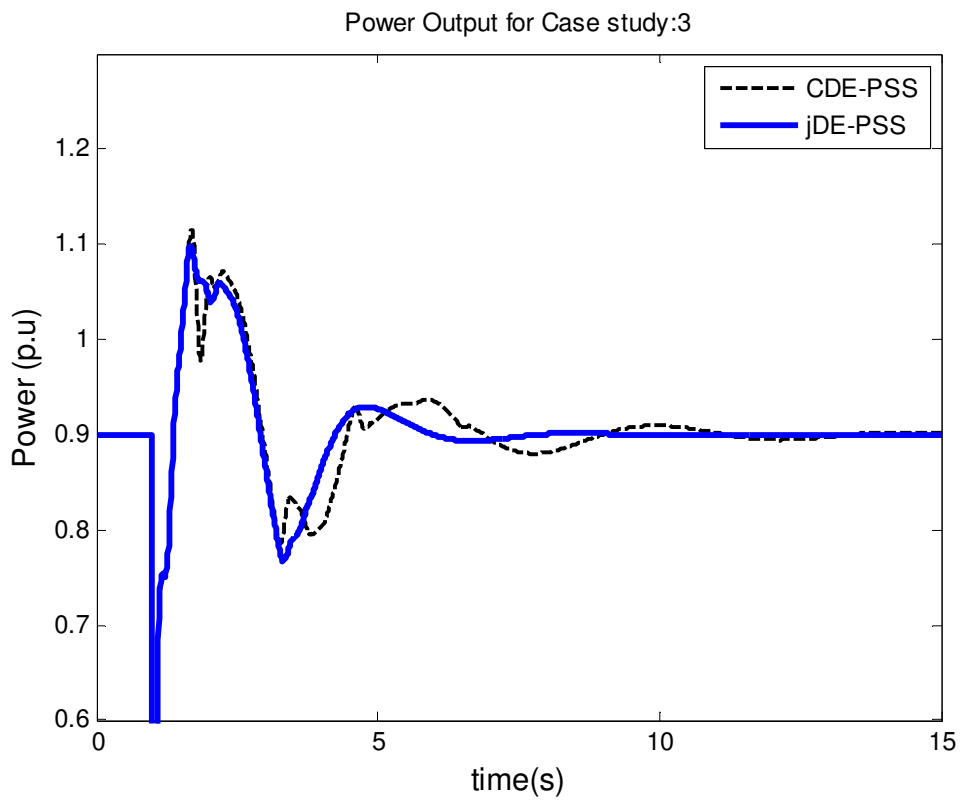


Figure 8.25: Active power output following 3-phase fault for Case 3

8.5 Summary

The effects of DE' intrinsic parameters (mutation factor, crossover probability and population size) have been investigated when applied to PSS tuning. The performance of DE is in fact dependent of the settings of these parameters. Low values of F cause DE to prematurely converge to local optimum whereas values between [0.9, 0.95] ensures good performance of the algorithm.

The crossover probability (CR) is responsible for the diversity within a given population. It is observed that low values of CR restrain DE from exploring the entire search space despite the diversity. This may cause stagnation. However when CR is set between [0.9-0.95], the algorithm is able to both remain diverse and escape stagnation, which subsequently search better for the global optimum. CR should not be set to 1 because it may cause a quick loss of diversity.

The population size is also responsible for the diversity. Small population sizes have revealed that DE prematurely converges to an optimum value whereas large populations have slower convergence rates although yield better results.

The Self-Adaptive DE whereby F and CR are changing and N_p is fixed, was implemented to tune the PSS in SMIB. The results are compared with those of the DE. The modal analysis shows that the self-adaptive outperformed DE-PSS for all the cases considered. These results have been validated in time domain simulations under small disturbance where self-adaptive outperformed DE-PSS with faster settling times and slightly more overshoots and undershoots in some cases. In transient stability, it was observed that self-adaptive displayed slightly less overshoots than DE.

Chapter 9

Conclusions and Recommendations for Further Works

The problem of optimally tuning the power system stabilizer has been addressed in this thesis. Optimization techniques based on Computational Intelligence have been used to find the optimal set of the PSS' parameters that provides adequate damping for a wide range of operating conditions. Thus, in the first part, DE was developed and implemented to tune the PSS. The algorithm performances were evaluated by comparing it to PBIL, and CPSS. An empirical rule was formulated with regard to DE intrinsic parameters that ensured optimal performance of the algorithm. To substitute the trial-and-error tuning approach, an adaptive scheme was applied to DE' control parameters and compared to the Classic version of DE (CDE).

An eigenvalue based objective function was considered and implemented to optimally tune the PSS. The optimization consisted of maximizing the lowest damping ratio that corresponded to the electromechanical modes across all the operating conditions. The tests were carried out on a Single Machine Infinite Bus and Two-Area Multimachine systems.

Some of the primary requirements of good optimization algorithms are based on their efficiency, robustness and simplicity. And when applied to PSS tuning, the resulting PSS must be robust enough to wide variations in the system' operating conditions. Hence, investigations revealed that the CPSS does provide satisfactory performances for the operating conditions around the nominal conditions. But the performance deteriorated as the system loading was increased. Both DE and PBIL, on the other hand ensured stability over the operating conditions considered in this work. Especially, DE-PSS which yielded better results than PBIL-PSS. The modal analysis showed that DE had better damping, and the step-response validated the findings by settling faster than PBIL-PSS and CPSS in both SMIB and Two-Area systems. The PSSs were further tested for transient

disturbance to ensured robustness of the stabilizers when subjected to a three-phase fault. The time domain simulations revealed that all PSSs performed satisfactory. However, CPSS least performed whereas DE and PBIL had similar performance.

Based on the simulations results obtained in this thesis, the follow conclusions were drawn:

- ❖ PBIL is a simpler algorithm with two control parameters to be set, population and learning rate.
- ❖ PBIL is most efficient in terms of memory utilisation since it only keeps two solutions, the best one and the one being evaluated. However PBIL requires more generations to find the optimal solution. Because PBIL need to re-populate at every generation based on the probability vector, it also requires more time. Whereas DE is faster due to the floating point representation of the population which facilitate arithmetic operations such differential mutation. The speed is also attributed to the one-to-on selection type.
- ❖ DE performed better than PBIL in converging to highest minimum damping hence better explored the search space for optimal set of PSS' parameters.
- ❖ The time domain simulation validated the modal analysis whereby DE-PSS settled faster than PBIL-PSS and CPSS when subjected to set-change input in the reference voltage. Both DE and PBIL based PSSs proved to be more robust than the CPSS with similar performances when the systems were subjected to transient fault.
- ❖ Whereas DE has 3, mutation factor, the crossover probability and the population number. It was established that the performance of DE is closely dependent on its control parameters namely, mutation factor (F), crossover probability (CR), and population size. These parameters are susceptible to change according to the problem being optimized. Hence, an empirical rule for choosing DE intrinsic

control parameters for PSS tuning was developed. F and CR effects were first investigated by individually testing varying them and keeping the population constant. The investigations showed that DE performed best when F and CR is between 0.9 and 1. In this thesis, $F=CR=0.95$ was found to be satisfactory for both SMIB and Two-Area. Despite the guidelines provided in literature, these parameters are problem-based and must be carefully chosen. Trial-and-error tuning is often used. This approach is very intricate and time consuming. Hence, self-adaptive was implemented to address that issue.

- ❖ The results in this thesis showed, that jDE-PSS yielded better performance than CDE-PSS in small signal analysis whereas transient analysis revealed much close performance.

The effect of population size on DE's performance was also investigated. Increasing the population size has significant impact on DE. The convergence rate of DE suggests that larger population improves the performance by achieving higher best minimum damping ratios as opposed to small population. However, large population groups are highly diverse and slow in convergence. This might require an increase in number of function evaluations (generations) or readjustment of the mutation F and the crossover CR .

Based on the conclusions drawn from the work, the following recommendations can be made

- ❖ Further investigation need to be carried out in simultaneously tuning the PSS in New-England, 10-machine or even larger system to determine with certitude the robustness of DE.
- ❖ Validate DE-PSS' simulation results by implementing it into a physical system.
- ❖ Investigate the online application of DE in tuning the PSS.
- ❖ It would be desirable to further evaluate the Self-Adaptive DE (jDE), by considering larger systems. Moreover, investigate different types of adaptive schemes applied to DE.

- ❖ A fixed population size was considered in this work whilst adapting the control parameters F and CR which were associated with each individual. Other adaptive techniques that include the population size must be explored.

References

- [1] P. Kundur, *Power System Stability and Control.*: Prentice-Hall, 1994.
- [2] Pal Bikash and Chaudhuri Balarko, *Robust Control in Power Systems.*: Springer Science, 2005.
- [3] S.P.N Sheetekela, "Design of Power System Stabilizers using Evolutionary Algorithms," Electrical, University of Cape Town, Cape Town, Thesis 2010.
- [4] Graham Rogers, *Power System Oscillations.* Boston: Kluwer Academic, 1999.
- [5] PM Anderson and Fouad AA, *Power System Control and stability.* Iowa, USA: The Iowa State University Press, 1994.
- [6] K.A. Folly, "Multimachine Power System Stabilizer design based on a simplified version of genetic algorithms combined with learning," in *Proceeding of the 13th International conference on Intelligent Systems Application to Power Systems*, 2005.
- [7] KA Folly, "Small-signal stability enhancement of Power Systems using PBIL based power system stabilizer ," in *Artificial Intelligence in Energy Systems and Power*, 2006.
- [8] A. Ghosh, G. Ledwich, O. P. Malik, and G. S. Hope, "Power System Stabilizer Based on Adaptive Control Techniques," *IEEE Transaction on Power Apparatus and Systems*, , vol. Vol. PAS-103, no. 8, pp. 1983-1989, August 1984.
- [9] Shaoru Zhang and Fang Lin Luo, "An Improved Simple Adaptive Control Applied to Power System Stabilizer," *IEEE TRANSACTIONS ON POWER ELECTRONICS*, vol. 24, no. 2, FEBRUARY 2009.
- [10] S Chen and OP Malik, "H-infinity optimisation-based power system stabiliser," *IEE Proc.Gener. Transm. Distrib.*, vol. 142, no. 2, pp. 179-184, March 1995.
- [11] T.C Yang, "Applying optimisation method to power system stabiliser design Part 1: single-machine infinite-bus systems," *Electrical Power & Energy Systems*, vol. 19, no. 1, pp. 29-35, 1997.
- [12] S.S. Lee, Y. T. Yoon, J.K Park, S.Y. Li, "Non-linear Adaptive Decentralised Stabilization Control for Multimachine Power Systems," *International Journal of*

Control , Automation and Systems, vol. 7, no. 3, pp. 389-397, 2009.

- [13] O.P. Malik, G.S. Hope, Y.H. Qin and G.Y. Xu, G.P. Chen, "An adaptive Power System Stabilizer based on the self – optimizing pole shifting control strategy," *IEEE Transaction on Energy Conversion*, vol. 8, no. 4, pp. 639 – 645, December 1993.
- [14] A. Abido, and H. Mantaury Y. L. Abdel – Magid, "Simultaneous stabilization of Multimachine Power System via genetic algorithm," *IEEE Trans. Power Sys.*, vol. 14, no. 4, pp. 1428 – 1438, November 1999.
- [15] KA Folly and GK Venayagamoorthy, "Optimal tuning of system stabilizer parameters using PBIL with adaptive learning rate," in *Power and Energy Society General Meeting*, 2010, pp. 1-6.
- [16] K. Price, R. Storn, and J. Lampinen, *Differential Evolution: A practical approach to global optimization*. Berlin Heidelberg: Springer-Verlag, 2005.
- [17] R. Curuana S. Baluja, "Removing the genetic from the standard genetic algorithm," in *Proceedings of the twelfth international conference on machine learning*, Lake Tahoe, July 1995.
- [18] R. Storn and K. Price, "Differential Evolution- a simple and efficient adaptive scheme for global optimization over continuous space," 1995.
- [19] J. Brest, S. Greiner, B. Boskovic, M. Mernik, and V. Zumer, "Self-adapting control parameters in differential evolution: a comparative study on numerical benchmark problems," *IEEE Trans. on Evolutionary Computation* , pp. 646-657, 2006.
- [20] S. Baluja, "Population-based incremental learning: A method for integrating genetic search based function optimization and competitive learning," Carnegie Mellon University, Technical report CMU-CS-94-163 1994.
- [21] G. Rogers, "Demystifying power system oscillations," *IEEE Computer Applications in Power*, vol. 9, no. 3, pp. 30 - 35, 1996.
- [22] Antal Soos, "An Optimal Adaptive Power System Stabilizer," Department of Electrical and Computer Engineering, University of Calgary, Calgary, Thesis 1997.
- [23] P. Kundur, M. Klein, G.J. Rogers, and M.S. Zywno, "Application of Power System Stabilizers for enhancement of overall system stability," *IEEE Transactions on*

Power Systems, vol. 4, no. 2, pp. 614-626, May 1989.

- [24] K.R. Padiyar, *Power System Dynamics Stability and Control*. Bangalore: Wiley, 1996.
- [25] E.V. Larsen and D.A. Swann, "Applying Power System Stabilizers, Part I; General concepts, Part II; Performance objectives and Tuning concepts, Part III; Practical considerations," *IEEE Trans. on Power Apparatus and Systems*, vol. PAS-100, pp. 3017-3046, 1981.
- [26] Antal Soós and O.P. Malik, "An H₂ Optimal Adaptive Power System Stabilizer," *IEEE TRANSACTIONS ON ENERGY CONVERSION*, vol. 17, no. 1, pp. 143-149, March 2002.
- [27] A. Hariri and O.P. Malik, "Adaptive-Network-Based Fuzzy Logic Power System Stabilizer," in *IEEE WESCANEX PROCEEDINGS*, 1995, pp. 111-116.
- [28] Zhang S and Fang Lin Luo, "An Improved Simple Adaptive Control Applied to Power System Stabilizer," *IEEE TRANSACTIONS ON POWER ELECTRONICS*, vol. 24, no. 2, pp. 369-375, FEBRUARY 2009.
- [29] A. Hariri and O.P. Malik, "A Fuzzy Logic Based Power System Stabilizer with Learning Ability," *IEEE Transactions on Energy Conversion*, vol. 11, no. 4, pp. 721-727, December 1996.
- [30] T. Basar and P. Bernhard, *H-Infinity Optimal Control and Related Minimax Design Problems*. Boston: Birkhäuser Boston, 1995.
- [31] KA Folly, "Robust Controller Design for Small-Signal Stability Enhancement of Power Systems," in *IEEE AFRICON*, 2004, pp. 631-636.
- [32] K.A. Folly, N. Yorino, and H. Sasaki, "Improving the Robustness of H-Infinity PSSs Using the Polynomial Approach," *IEEE Trans. on Power Systems*, vol. 13, no. 4, pp. 1359-1364, 1998.
- [33] J Reddy and J.K Mendiratta, "H-Infinity Loop Shaping Based Robust Power System Stabilizer for Dynamic Equivalent Multi-machine Power System," in *IEEE International Energy Conference*, 2010, pp. 582-587.
- [34] KA Folly, "On the prevention of Pole-zero cancellations in H infinity Power System Controller Design: A comparison," *SAIEE Africa Research Journal*, vol. 99, no. 4,

December 2008.

- [35] Y.N. Yu and C. Siggers, "Stabilization and Optimal Control Signals for a Power System," *IEEE Transactions on Power Apparatus and Systems*, vol. 90 , no. 4, pp. 1469 - 1481, 1971.
- [36] Y.N. YU, KHIEN VONGSURIY, and L.N. WEDMAN, "Application of Optimal theory to Power System," *IEEE TRANSACTIONS ON POWER APPARATUS AND SYSTEMS*, vol. 89, no. 1 , pp. 55-62, 1970.
- [37] L.J. Cai and I. Erlich, "Simultaneous Coordinated Tuning of PSS and FACTS Controller for Damping Power System Oscillation in Multi-Machine Systems," in *IEEE Proceedings: Power Tech Conference*, vol. 2, 2003.
- [38] N. Mithulanathan, C.A Canizares, J Reeve, and G. J Rogers, "Comparison of PSS, SVC, and STATCOM controllers for damping power system oscillations," *IEEE trans on power systems*, vol. 18, no. 2, pp. 786-792, May 2003.
- [39] J. Nocedal and S.J. Wright, *Numerical Optimization*. New York: Springer-Verlag, 1999.
- [40] Z. Michalewicz, *Genetic Algorithms+Data Structure=Evolution Programs.:* 3rd ed. Springer-Velag; , 1996.
- [41] K.A. Folly, "Multimachine power system stabilizer design based on a simplified version of Genetic Algorithm combined with Learning," in *International Conference on Intelligent Systems Application to Power Systems (ISAP)* , 2005.
- [42] Pinaki Mitra and et.al, "Comparative of Population based Techniques for Power System Stabilizer Design," in *15th International Conference on Intelligent System Applications to Power Systems ISAP '09*, 2009.
- [43] H. Mühlenbein and D. Schlierkamp-Voosen, "The science of breeding and its application to the breeder genetic algorithm BGA," *IEEE Trans on Evolutionary Computation*, vol. 1, no. 4, pp. 335-360, 1994.
- [44] H. Mühlenbein and D. Schlierkamp-Voosen, "Predictive models for the Breeder Genetic Algorithm, I.: continuous parameter optimization," *Evolutionary Computation*, vol. 1, no. 1, pp. 25-49, spring 1993.
- [45] A. Qing, *Differential Evolution: Fundamental and applications in Electrical*

Engineering.: John Wiley & Sons, 2009.

- [46] T. Mulumba and K.A. Folly, "Design and comparison of Multi-machine Power System Stabilizer base on Evolution Algorithms," in *Universities' Power Engineering Conference (UPEC), Proceedings of 2011 46th International* , Soest, 2011, pp. 1 - 6.
- [47] Z. Wang, C.Y Chung, K.P. Wong, and C.T. Tse, "Robust power system stabilizer design under multi-operating conditions using differential evolution," *IET Gener. TransmDistrib.*, vol. 2, no. 5, pp. 690 – 700, 2008.
- [48] H. Shayeghi, A. Safari, and H.A. Shayanfa, "Multimachine Power System Stabilizers Design Using PSO Algorithm," *International Journal of Electrical and Electronics Engineering*, vol. 4, no. 4, pp. 226-233, 2009.
- [49] M.A. Abido, "Robust Design of Multimachine Power System Stabilizers Using Simulated Annealing," *IEEE Trans on Energy Conversion*, vol. 15, no. 3, 2000.
- [50] R. Storn and K. Price, "Differential Evolution: a simple and efficient heuristic for global optimization over continuous spaces," *Journal of Global Oprimization*, vol. 11, pp. 341-359, 1997.
- [51] J. Zhang and A.C. Sanderson, *Adaptive Differential Evolution: A robust approach to multimodal problem optimization*. Berlin Heidelberg: Springer, 2009.
- [52] Considine G.D. and P.H. Kulik, Eds., *Van Nostrand's Scientific Encyclopedia*, 9th ed. New York, USA: John Wiley & Sons Inc., 2002.
- [53] J.E. Smith A.E. Eiben, *Introduction to evolutionary computing.*: Springer, 2003.
- [54] R. Storn, "On the usage of Differential Evolution for function optimization," in *Fuzzy Information Processing Society*, 1996, pp. 519-523.
- [55] S. Rahnamayan and et. al, "Opposition-Based differential evolution," *IEEE Trans. on Evolutionary computation*, vol. 12, pp. 64-79, 2008.
- [56] H. Chi Y. Ao, "Experimental Study on Differential Evolution Strategies," in *Global Congress on Intelligent Systems*, 2009.
- [57] Science daily. (2012, June) Science daily. [Online]. http://www.sciencedaily.com/articles/c/chromosomal_crossover.htm
- [58] DE Goldberg, *Genetic algorithms in search, optimization and machine learning*.

Reading, MA: Addison-Wesley, 1989.

- [59] S. Yang and X. Yao, "Population-Based Incremental Learning with Associative for Dynamic Environments," *IEEE Trans. on Evolutionary computation*, vol. 12, no. 5, pp. 542-561, October 2008.
- [60] S.L. Ho and S. Yang, "A population-Based Incremental Learning Method for robust optimal Solution," *IEEE Trans. on Magnetics*, vol. 46, no. 8, pp. 3189-3192, August 2010.
- [61] S.Y. Yang and et. all, "A new implemetation of population based incremental learning method for optimizations in Electromagnetics," *IEEE Trans. on magnetics*, vol. 43, no. 4, pp. 1601-1604, April 2007.
- [62] KA Folly, "Robust Controller Design Based on a Combination of genetic Algorithms and Competitve Learning," in *Proceedings of International Joint Conference on Neural Networks*, Orlando, Florida, USA, August 2007.
- [63] R. Mallipeddi and P. N. Suganthan, "Empirical Study on the Effect of Population Size on Differential," *IEEE Congress on Evolutionary Computation*, pp. 3663 - 3670, 2008.
- [64] Jason Teo, "Exploring dynamic self-adaptive populations in differential evolution," *Soft Comput*, pp. 673–686, August 2005.
- [65] J. Vesterstroem and R. Thomsen, "A comparativestudy of differential evolution, particle swarm optimization, and evolutionary algorithms on numerical benchmark problems," *IEEE congress on Evolutionary Computation*, pp. 1980-1987, June 2004.
- [66] Lampinen J and Zelinka I., "On Stagnation of Differential Evolution Algorithm," in *Proc. of MENDEL, 6th Int. Mendel Conf. on Soft Computing*, Brno, Czech Republic, 2000.
- [67] J. Lui and J. Lampinen, "A Fuzzy adaptive differential evolution algorithm.," *Soft Computation: Fusion Found Method Appl.*, pp. 448-462, 2005.
- [68] J. Brest, Borko Boskovic, and Viljem Zumer, "An improved self-adaptive Differential Evolution Algorithm in Single objective constrained real-parameter optimization," in *Congress on evolutionary computation*, 2010, pp. 1-8.
- [69] P.N. Suganthan A.K. Quin, "Self-Adaptive Differential Evolution Algorithm for

- numerical optimization," in *Congres on Evolutionary Computation*, 2005, pp. 1785-1791.
- [70] U.K. Chakraborty, *Advances in Differential Evolution.*: Springer-Verlag, 2008.
- [71] E. Mezura-Montes, J. Velázquez-Reyes, and C. A. and Coello, "A comparative study of differential evolution variants for global optimization," in *in GECCO*, Seattle, Washington, USA, 2006, pp. 485–492.
- [72] J. Brest and M. S. Maucec, "Population size reduction for the differential evolution algorithm," *Applied Intelligence*, vol. 29, no. 3, pp. 228–247, 2008.
- [73] J. Montgomery, "Crossover and the different faces of differential evolution searches," in *IEEE Congress on Evolutionary Computation (CEC)*, 2010.
- [74] S. M. Islam, "An adaptive differential Evolution algorithm with novel mutation and crossover strategies for global numerical optimization," *IEEE Trans. on systems, man, and cubernetics*, vol. 42, no. 2, pp. 482-500, 2012.

Research Publications

1. T. Mulumba, K. A. Folly, O.P. Malik “Tuning of PSS parameters using Differential Evolution”, *20th Southern African Universities’ Power Engineering Conference (SAUPEC 2011)*, July 13-15, 2011, Cape Town, South Africa
2. T. Mulumba, K.A Folly, “Power System Stabilizer design: Comparison analysis between Differential Evolution and Population based Incremental Learning”, *20th Southern African Universities’ Power Engineering Conference (SAUPEC 2011)*, July 13-15, 2011, Cape Town, South Africa
3. T. Mulumba, K. A. Folly, “Design and comparison of Multi-machine Power System Stabilizer base on Evolution Algorithms”, *The 46th international Universities’ Power Engineering Conference (UPEC 2011)*, September 5- 8, 2011, Soest, Germany
4. T. Mulumba, K. A. Folly, “Application Evolution Algorithms to Power System Stabilizers design: Impact of the Population size in Differential Evolution and Population Based Incremental Learning”, *10th IEEE Africon 2011*, September 13-15, 2011, Victoria Falls, Livingstone, Zambia.
5. T. Mulumba, K. A. Folly, “Application of self-adaptive Differential Evolution to tuning PSS parameters”, *IEEE-PES Power Africa 2011*, July 9-13, 2011, Johannesburg, South Africa

Appendix

Appendix A: This section contains higher oscillations modes of DE-PSS, CPSS and PBIL-PSS. Further simulation results (modal analysis and step responses) of DE with different population sizes in chapter 8 have also been included.

Appendix B: This section gives the data for the SMIB system presented with results in Chapter 5, section 5.1. The generator parameters, transmission line data as well as the controller (AVR) are given under this section. Also presented under this section are equations that model the generator used in the simulations.

Appendix C: This section presents the data used in the Multi-machine simulations of Chapter 5, section 5.2. The models used for the generators, the AVR and the speed governor are given in this section.

Appendix A

A.1 SMIB High Frequency modes of oscillations

Table A.1 : AVR mode for SMIB

Case	No PSS	CPSS	PBIL-PSS	DE-PSS
1	-6.7205±j19.969 (0.3190)	5.0169±j22.845 (0.2145)	-4.5783±j22.762 (0.1972)	-6.1665±j22.512 (0.2642)
2	-6.6572±j20.191 (0.3131)	-5.5135±j23.226 (0.2310)	-5.2960±j23.1643 (0.2229)	-6.0897±j23.0875 (0.2550)
3	-6.6728±j20.110 (0.3149)	-5.0693±j23.026 (0.21500)	-4.6721±j22.9456 (0.1995)	-6.1079±j22.7322 (0.2595)
4	-6.6263±j20.282 (0.3105)	-5.5142±j23.337 (0.2300)	-5.3099±j23.2782 (0.2224)	-6.0548±j23.2072 (0.2525)
5	-6.6416±j20.202 (0.3123)	-5.1037±j23.144 (0.2153)	-4.7334±j23.0658 (0.2010)	-6.0706±j22.8744 (0.2565)

A.2 Self-Adaptive and DE High frequency modes of oscillation for SMIB

Table A.2 : AVR modes for Self-Adaptive DE (SMIB)

Case	DE-PSS	jDE-PSS
1	-5.54±j22.6 0.238	-4.60±j22.4 0.201
2	-5.55±j22.9 0.236	-4.75±j22.7 0.205
3	-5.56e±j23.1 0.234	-4.83±j22.9 0.206
4	-5.56±j23.2 0.233	-4.88±j23.0 0.207

A.3 Modal Analysis of population sizes**Table A.3 : Inter-area mode eigenvalues for different population sizes**

Population size	Case 1	Case 2	Case 3
DE-PSS (10)	-0.959 + j4.91 (0.192)	-0.871 + j4.81 (0.178)	-0.684 + j3.86 (0.174)
DE-PSS (30)	-1.19 + j4.72 (0.244)	-1.09 + j4.63 (0.229)	-0.815 + j3.75 (0.212)
DE-PSS (50)	-1.17 + j4.57 (0.248)	-1.06 + j4.48 (0.231)	-0.820 + j3.53 (0.2263)
DE-PSS (70)	-1.16 + j4.63 (0.242)	-1.06 + j4.55 (0.227)	-0.788 + j3.67 (0.210)
DE-PSS (100)	-1.03 + j4.84 (0.207)	-9.31e-001 + j4.75 (0.192)	-0.755 + j3.84 (0.193)
DE-PSS (150)	-1.15 + j4.63 (0.241)	-1.06 + j4.54 (0.226)	-0.784 + j3.67 (0.209)
DE-PSS (200)	-1.16 + j4.56 (0.247)	-1.06 + j4.47 (0.231)	-0.770 + j3.53 (0.213)

A.4 Step responses for PSS designed with different population sizes

The time domain simulations were performed to validate the results of modal analysis. The PSSs are assessed by their ability to damp low frequency oscillations following a 10% step change in the reference voltage of the generator that has the largest participation on the inter-area mode, which is generator 2.

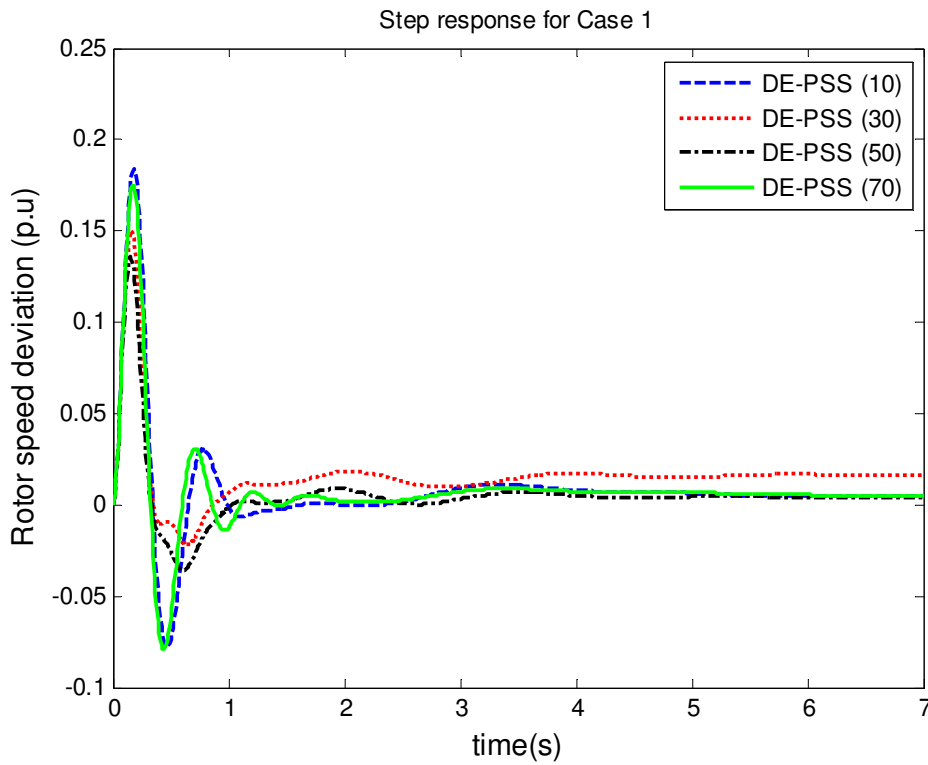


Figure A.1: Step responses for case 1 for population of 10, 30 50 & 70

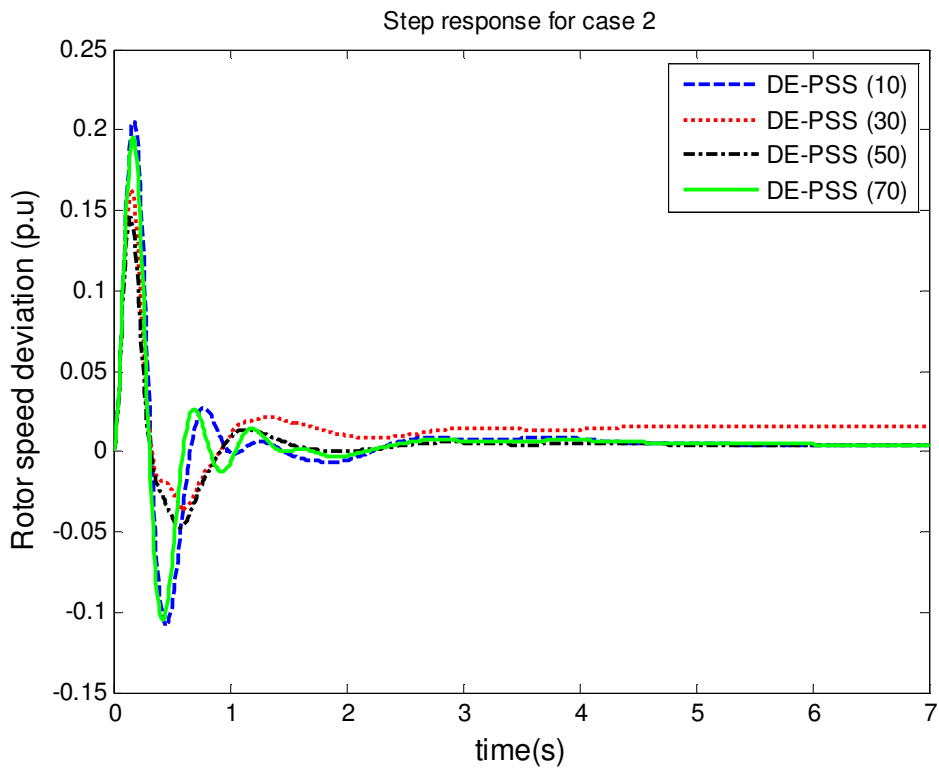


Figure A.2: Step responses for case 2 for population of 10, 30 50 & 70

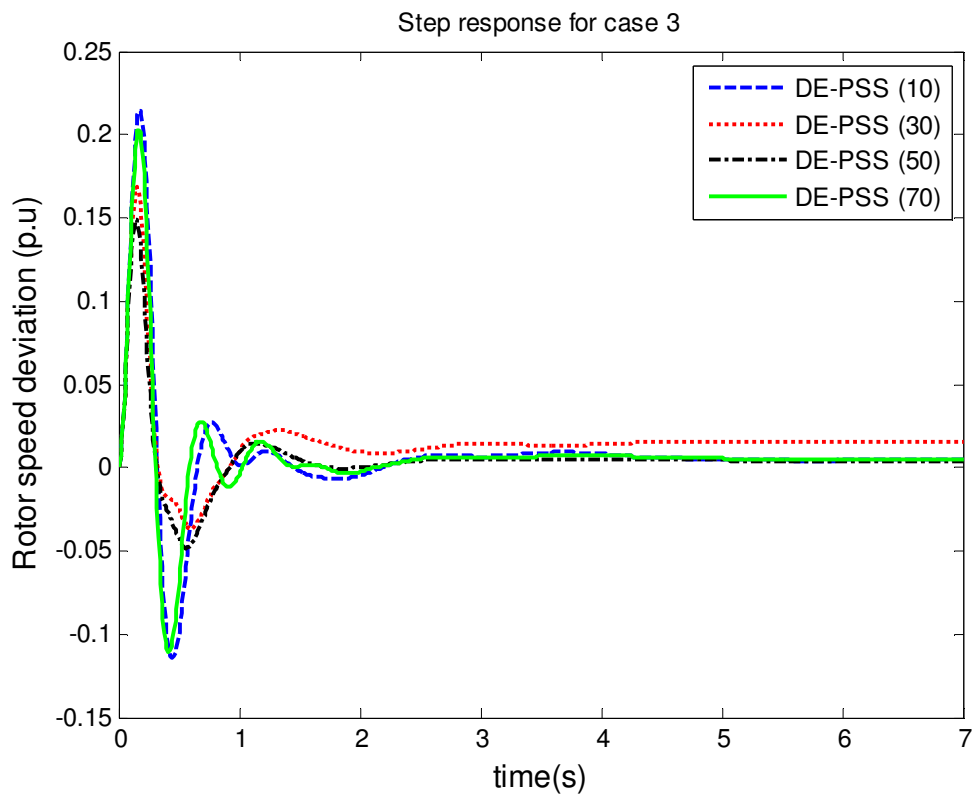


Figure A.3: Step responses for case 3 for population of 10, 30 50 & 70

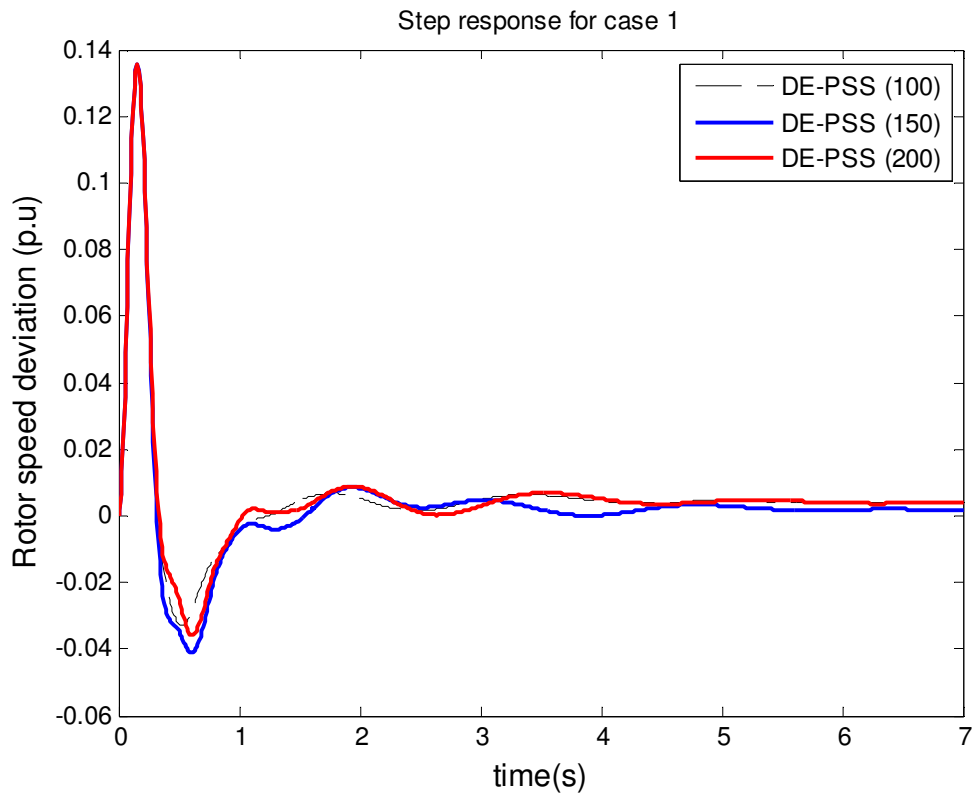


Figure A.4: Step responses for case 1 for population of 100, 150 & 200

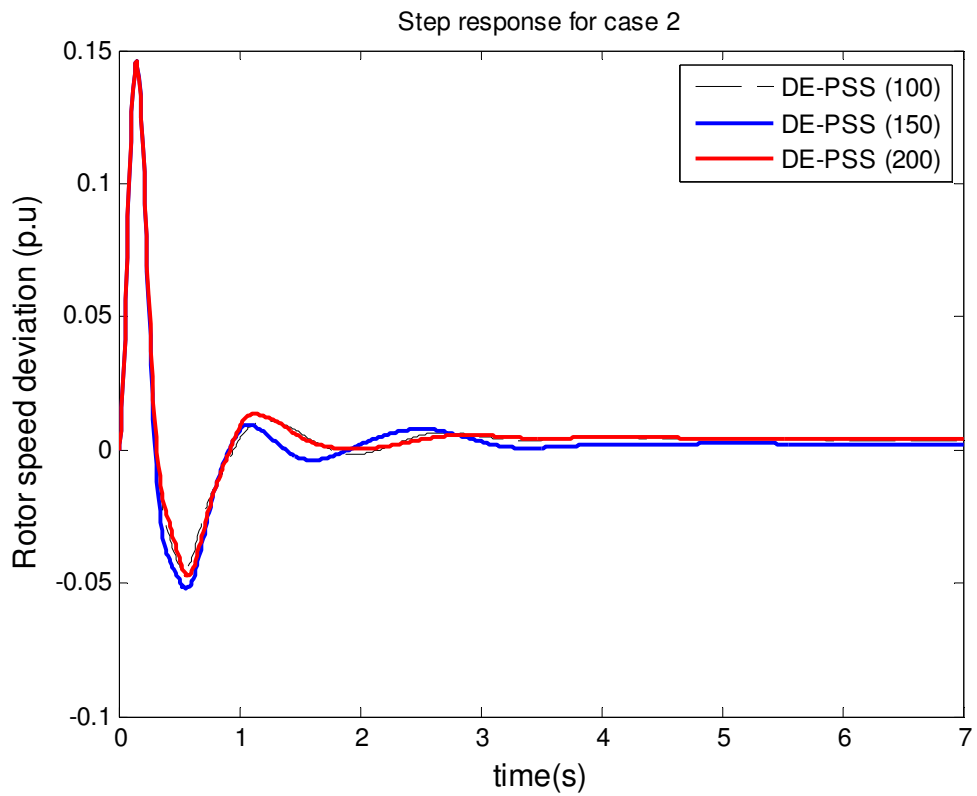


Figure A.5: Step responses for case 2 for population of 100, 150 & 200

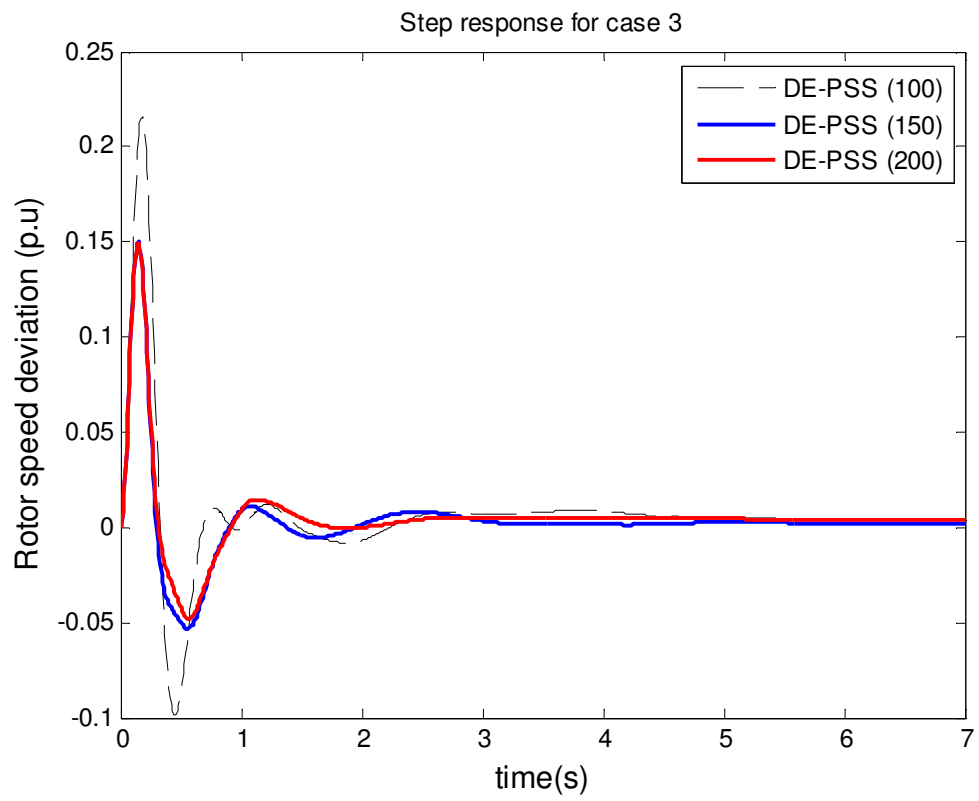


Figure A.6: Step responses for case 3 for population of 100, 150 & 200

Appendix B

SMIB System

B.1 Model Equations

Simple AVR structure

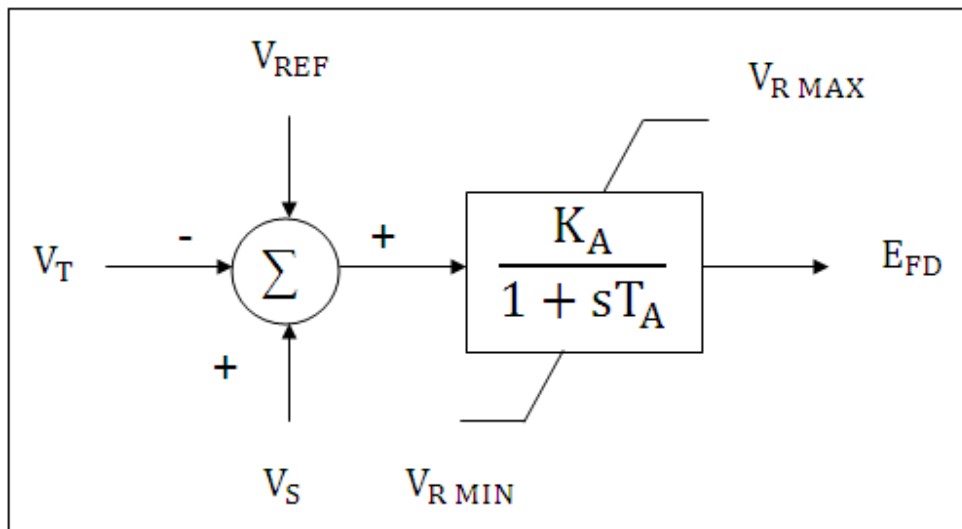


Figure B.1: AVR block diagram used in the SMIB system

Where:

T_A is the regulator time constant

K_A is the exciter gain

V_T is the terminal voltage

V_S is the PSS control signal

V_{REF} is the reference voltage

$V_{R\ MAX}$ is the maximum limit of the AVR signal

$V_{R\ MIN}$ is the minimum limit of the AVR signal

E_{FD} is the electric field signal

For the Single machine to infinite bus system, the following generator model was used; it's a 6th order sub-transient model.

The generator equations are:

$$\frac{d\Delta\bar{\omega}_r}{dt} = \frac{1}{2H} (\bar{T}_m - \bar{T}_e - K_D\Delta\bar{\omega}_r)$$

$$\frac{d\delta}{dt} = \omega_o\Delta\bar{\omega}_r$$

The rotor circuit equations are represented as follows:

$$\frac{d\psi_{fd}}{dt} = \frac{\omega_o R_{fd}}{L_{adu}} E_{fd} - \omega_o R_{fd} i_{fd}$$

$$\frac{d\psi_{1d}}{dt} = -\omega_o R_{1d} i_{1d}$$

$$\frac{d\psi_{1q}}{dt} = -\omega_o R_{1q} i_{1q}$$

$$\frac{d\psi_{2q}}{dt} = -\omega_o R_{2q} i_{2q}$$

The rotor currents are expressed by the following:

$$i_{fd} = \frac{1}{L_{fd}} (\psi_{fd} - \psi_{ad})$$

$$i_{1d} = \frac{1}{L_{1d}} (\psi_{1d} - \psi_{ad})$$

$$i_{1q} = \frac{1}{L_{1q}} (\psi_{1q} - \psi_{aq})$$

$$i_{2q} = \frac{1}{L_{2q}} (\psi_{2q} - \psi_{aq})$$

The electrical torque is expressed by the following:

$$T_e = \psi_{ad} i_{1q} - \psi_{aq} i_{1d}$$

B.2 SMIB System Data

For the Single machine to infinite bus system, a 6th order sub-transient generator model was used with a simple exciter with the following data

Generator data

Rating: 991MVA, 22kv, 50Hz

$$X_l = 0.15; r_a = 0.0; X_d = 2.0; X'_d = 0.245;$$

$$X''_d = 0.2; T'_{do} = 5.0; T''_{do} = 0.031;$$

$$X_q = 1.91; X'_q = 0.42; X''_q = 0.2;$$

$$T'_{qo} = 0.66; T''_{qo} = 0.061; H = 2.875$$

Automatic Voltage Regulator parameters

$$K_a = 200 \text{ and } T_a = 0.03$$

Transmission Line parameters

$$r = 0$$

$$b_c = 0$$

Power system stabilizer structure

The parameters K_s , T_1 - T_4 were designed using conventional method and evolutionary algorithms.

T_w of 2.5s was used.

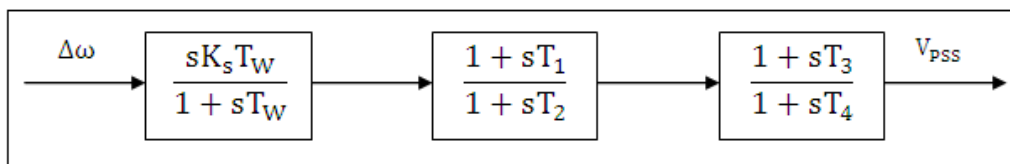


Figure B.2 : PSS block diagram used in SMIB

Appendix C

Multi- Machine System

C.1 System Model Equations

$$\frac{d\Delta\bar{\omega}_r}{dt} = \frac{1}{2H} (\bar{T}_m - \bar{T}_e - K_D\Delta\bar{\omega}_r)$$

$$\frac{d\delta}{dt} = \omega_o\Delta\bar{\omega}_r$$

The rotor circuit equations are represented as follows:

$$\frac{d\psi_{fd}}{dt} = \frac{\omega_o R_{fd}}{L_{adu}} E_{fd} - \omega_o R_{fd} i_{fd}$$

$$\frac{d\psi_{1d}}{dt} = -\omega_o R_{1d} i_{1d}$$

$$\frac{d\psi_{1q}}{dt} = -\omega_o R_{1q} i_{1q}$$

$$\frac{d\psi_{2q}}{dt} = -\omega_o R_{2q} i_{2q}$$

The rotor currents are expressed by the following:

$$i_{fd} = \frac{1}{L_{fd}} (\psi_{fd} - \psi_{ad})$$

$$i_{1d} = \frac{1}{L_{1d}} (\psi_{1d} - \psi_{ad})$$

$$i_{1q} = \frac{1}{L_{1q}} (\psi_{1q} - \psi_{aq})$$

$$i_{2q} = \frac{1}{L_{2q}} (\psi_{2q} - \psi_{aq})$$

The electrical torque is expressed by the following:

$$T_e = \psi_{ad} i_{1q} - \psi_{aq} i_{1d}$$

C.2 System Data

Generator data

All the generators used in the system are identical with the following parameters:

Rating: 900MVA, 22kV, 60Hz

$$\begin{aligned} X_d &= 1.8; & X'_d &= 0.3; \\ X'_d &= 0.25; & T'_{do} &= 8.0; & T''_{do} &= 0.03; & X_q &= 1.7; \\ X'_q &= 0.55; & X''_q &= 0.24; & T'_{qo} &= 0.4; & T''_{qo} &= 0.05; \\ H &= 6.5; & K_D &= 0; & A_{sat} &= 0.0654; & B_{sat} &= 0.5743; \end{aligned}$$

All the reactances are in per unit, while the time constants are in seconds

Automatic Voltage Regulator (AVR) Parameters

$$K_a = 200 \text{ and } T_a = 0.05 \quad T_r = 0.01$$

Turbine Governor block diagram and Parameters

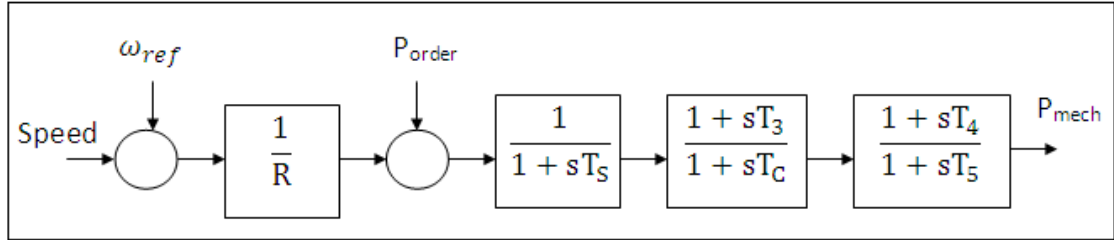


Figure C.1: Speed governor block diagram used in the simulations for multi-machine

$$\begin{aligned} \text{set point } (\omega_f) &= 1 & \text{Gain } \left(\frac{1}{R}\right) &= 25 & T_{\max} &= 1 & T_s &= 0.1 & T_c &= 0.5 \\ T_3 &= 0.0 & T_4 &= 1.25 & T_5 &= 5.0 \end{aligned}$$

Power System stabilizer Structure

The following PSS structure was used, the parameters K_s , T_1 , T_2 , T_3 and T_4 were designed using the evolutionary algorithms, while the washout time constant T_w of 10s was used.

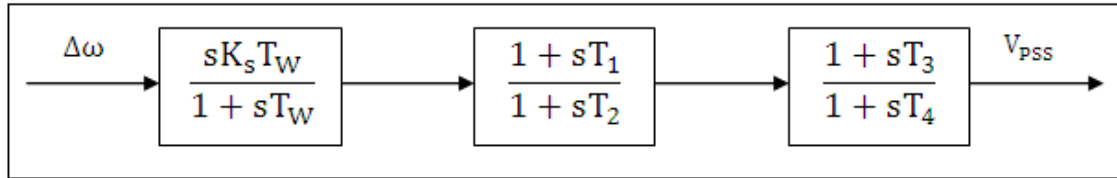


Figure C.2: Block diagram for the PSS used in the Multi-machine system

All the reactances are in per unit, while the time constants are in seconds. The generators active power output and line reactance were varied to simulate different operating conditions.