

**A SYSTEM FOR THE
ACQUISITION AND DIGITAL
ANALYSIS OF LOWER LIMB
FLOW WAVEFORMS.**

BY : L. SMITH

Supervisors : Dr. W.L. Capper
Assoc. Prof. A.E. Bunn
Prof. E.J. Immelman

**THESIS SUBMITTED IN PARTIAL FULFILMENT OF THE
REQUIREMENTS FOR THE DEGREE OF MASTER OF
SCIENCE IN BIOMEDICAL ENGINEERING**

**UNIVERSITY OF CAPE TOWN
MARCH 1994**



The copyright of this thesis vests in the author. No quotation from it or information derived from it is to be published without full acknowledgement of the source. The thesis is to be used for private study or non-commercial research purposes only.

Published by the University of Cape Town (UCT) in terms of the non-exclusive license granted to UCT by the author.

DECLARATION

I **Leonard Smith** hereby declare that this dissertation is my own, unaided work. It is being submitted for the degree of Master of Science at the University of Cape Town. It has not been submitted before in any form for any degree or examination at any other university

Signed

:

Date

:

Signed by candidate

ACKNOWLEDGEMENTS

The author wish to thank all those who have aided in the work presented herein. A special work of gratitude is expressed to my friends and parents for their moral and emotional support during my stay at university.

ABSTRACT

A PC based waveform acquisition and analysis system has been developed for use in aorto-iliac arterial assessment. A Motorola DSP56001 based system containing dual Analog to Digital converters is used to sample phase quadrature demodulated signals from a commercially available continuous wave Doppler unit. The Power Spectral Density is calculated using an autoregressive model from which the mean velocity waveform is calculated. This waveform is used to calculate the damping factor, vessel compliance and runoff resistance of a simple electrical model of the lower limb arterial circulation using a non-linear regression technique of curve fitting in the time domain. A pilot study using the system shows a significant separation ($p < 0.001$ Mann Whitney U-test) between the damping factors of a normal control group (quartile range = 0.15 - 0.25; median = 0.19) and a patient group with angiographically determined aorto-iliac arterial disease (quartile range = 0.45 - 0.89; median = 0.49).

CONTENTS

ACKNOWLEDGEMENTS	i
ABSTRACT	ii
LIST OF FIGURES	v
LIST OF ABBREVIATIONS	vi
CHAPTER ONE	7
1.1 INTRODUCTION	7
1.2 THESIS OBJECTIVES	8
CHAPTER TWO	9
2.1 LITERATURE SURVEY	9
2.1.1 SAMPLING AND SIGNAL PROCESSING.....	9
2.1.1.1 AR SPECTRAL ESTIMATION	12
2.1.2 BLOOD VELOCITY WAVEFORM ANALYSIS.....	15
CHAPTER THREE	21
3.1 HARDWARE	21
3.1.1 ULTRASOUND UNIT.....	21
3.1.2 MICRO-CONTROLLER ENVIRONMENT.....	22
3.1.3 HOST PLATFORM.....	22
3.2 SOFTWARE IMPLEMENTATION	22
3.2.1 MICRO-CONTROLLER ENVIRONMENT.....	22
3.2.2 HOST PLATFORM.....	27
3.2.3 CODE DEVELOPMENT FOR MATLAB.....	29
CHAPTER FOUR	32
4.1 INTRODUCTION	32
4.2 METHODS	32
4.3 RESULTS	33
CHAPTER FIVE	40
5.1 DISCUSSION	40
5.1.1 SYSTEM HARDWARE.....	40
5.1.2 SYSTEM SOFTWARE.....	40
5.1.3 SYSTEM MODEL.....	40
5.1.4 RESULTS.....	41
5.2 CONCLUSION	41
APPENDIX I	43
DOPPLER PHYSICS	43
APPENDIX II	44

FREQUENCY DOMAIN PROCESSING	44
APPENDIX III	46
DSP SOURCE CODE LISTING (0.0000-0.1372)	46
PC SOURCE CODE LISTING (1.0000-3.0067)	71
MATLAB SOURCE CODE LISTING (4.0000-4.0807)	105
REFERENCES	120

LIST OF FIGURES

Linear filter as a tapped delay line	12
Tri-phasic shape of a normal velocity waveform	16
Electrical analogue of the arterial system	18
System data flow diagram	21
DSP program loop	23
DSP interrupt schedule	24
FrameR interrupt	25
Spectral process flow	26
Spectral distribution	26
Program module control	28
Discrete system model	31
Processed data for normal group	33
Processed data for patient group	34
Subject data distribution	35
Doppler Physics	43
Frequency domain processing	44

LIST OF ABBREVIATIONS

AIC	Akaike's information criterion
ALU	Arithmetic logic unit
AR	Autoregressive
ASPI	Ankle systolic pressure index
BGI	Borland Graphic Interface
CAT	Criterion of autoregressive transfer function
CW	Continuous wave
DMA	Direct memory access
DSP	Digital signal processing
FDD	Floppy disk drive
FFT	Fast Fourier transform
FPE	Final prediction error
FZC	First zero crossing
HCR	Host Control Register
HDD	Hard disk drive
HSR	Host Status Register
IBM	International Business Machines
IIR	Infinite impulse response
IPL	Interrupt priority level
kb	kilobyte
LTA	Laplace transform analysis
LTD	Laplace transform damping
LTE	Laplace transform elasticity
LTR	Laplace transform resistance
Mb	Megabyte
MHz	Megahertz
MIPS	Million instructions per second
ms.	millisecond
MSE	Mean square error
PC	Personal computer
PCA	Principal component analysis
PI	Pulsatility index
PSD	Power Spectral Density
SBI	Spectral broadening index
VDU	Video display unit
VGA	Video Graphics Array

CHAPTER ONE

1.1 INTRODUCTION

Blood flow velocity within arteries varies from slow moving cells near the arterial walls to fast moving cells near the vessel center. When an artery is bathed in ultrasound, reflections occur from the moving cells. The reflected signal thus contains Doppler shifted frequencies which are proportional to the velocities of these cells. It follows that with varying velocities, a spectrum of representative frequencies is obtained. Once the reflected signal is demodulated, spectral analysis can extract the individual frequencies. If a mean frequency is chosen from the spectrum, a resultant mean blood velocity waveform can be constructed for a given time interval.

The waveform shape is dependent not only on the pulsatile pump (cardiac input) but also on the characteristics of the arterial system (e.g. arterial wall compliance, lumen diameter) and the fluid dynamics (e.g. viscosity, density). Studies have shown that the blood velocity waveform shape can be predicted if the characteristics of the arterial system are known. It is postulated that the reverse is also true i.e. if the shape of the blood velocity waveform is known, then the characteristics of arterial system can be extracted mathematically.

Two factors are crucial in this process. The first is that an optimum method of data acquisition and spectral analysis be applied to extract the blood velocity waveform. The second is that a representative model of the arterial system be used to extract the arterial parameters. This study concentrated on the former. A system has been developed which is able to measure, analyse and store the information contained in the reflected signal component. Work in the area of system modelling and parameter extraction is continuing as part of an ongoing project at the University of Cape Town's department of Biomedical Engineering.

Work from this thesis has been accepted for publication by *Automedica*, an international journal for automation in the medical sciences (Capper, Smith and Immelman, 1994).

1.2 THESIS OBJECTIVES

The thesis objectives are :

- I** To develop software which will allow a Digital Signal Processing (DSP) unit to sample the output from a suitable continuous wave (CW) Doppler demodulator, to calculate the spectral content and to display a plot thereof in real time.
- II** To develop software to store, load and plot blood velocity waveforms.
- III** To develop software which will allow an operator to analyse acquired blood velocity waveforms in terms of the pulsatility index (PI) and an equivalent 3rd order electrical model.
- IV** To use the system in a pilot study to measure the parameters of normal subjects and patients presenting with lower limb vascular disease. The latter group will also be assessed by other clinical means such as arteriography.

CHAPTER TWO

2.1 LITERATURE SURVEY

2.1.1 SAMPLING AND SIGNAL PROCESSING

The continuous wave ultrasound unit, with user selectable frequency of five or ten megahertz (MHz), which was used during this project uses phase quadrature demodulation to output two demodulated signals which differ in phase by 90° . The frequency shifts are within the audio spectrum. Accurate, stable spectral estimation of these Doppler shifted frequencies is of vital importance.

If the demodulated signal is divided into segments, each $N\Delta t$ long (N = number of data points in each slice and sampling rate = $1/\Delta t$) then the frequency content of the signal may be calculated for these N points using the Fourier transform. However, if one considers the random nature of the signal under investigation it becomes evident that the Fourier transform may be inadequate. In the field of signal processing this problem is partly overcome by calculation of the power spectral density (PSD). The power spectral density of a stationary time signal is defined according to the Wiener-Khintchine theorem as the Fourier transform of the autocorrelation function. The solution is best illustrated if one considers the periodogram approach where the spectrum is averaged over a number of short consecutive segments taken from the time signal.

We have seen that the capture and processing of data in real time establishes the need to isolate time frames within a continuous process. The difficulty in calculating the PSD is therefore that the complete time domain function is not available and an estimate of the autocorrelation function has to be made.

A possible solution is to utilise a long time window but the problem is complicated by the non-stationary nature of the Doppler Ultrasound signal. The signal can be considered to be quasi-stationary for a period of no greater than 10ms (Evans and Schlindwein 1989). To apply the PSD theorem the time window is therefore limited to this time duration.

As it follows intuitively that the signal is not zero outside this window the autocorrelation function needs to be adjusted. Spectral methods, such as the periodogram which incorporates a Fast Fourier Transform (FFT) calculation, do not allow for this phenomenon and therefore have a relatively large variance. The parametric spectral techniques, such as autoregressive modelling, use the available autocorrelation values to form a prediction of the points outside the given time window.

An alternative for improving the PSD estimate is to apply a weighting function to the time signal before calculating the autocorrelation coefficients. This has the effect of giving a greater significance to the initial autocorrelation coefficients when the data overlap is a maximum. This process of assigning a relative weighting to the autocorrelation coefficients is commonly employed and can also be found when inspecting the following equation for a biased autocorrelation estimation.

Biased autocorrelation $R_{xx}(m)$ of discrete time sequence $x(n)$ containing N samples is :

$$R_{xx}(m) = \frac{1}{N-m} \sum_{n=0}^{N-|m|-1} x(n)x(n+m) \quad (2.1)$$

for $m = 0, 1, 2, \dots, N-1$

When the PSD is calculated using the periodogram approach (magnitude squared of the Fourier transform) it is common practice to apply a weighting window to the time signal before calculating the PSD. The influence is slightly more intuitive. Digital time domain waveforms assume a periodic extension of the sampled window. Consider an example of a sampled sine wave. If the sampled window does not exactly overlap a whole number of signal wavelengths, an artificial discontinuity is introduced which manifests as undesirable sidelobes in the frequency domain. By applying a weighting window these discontinuities are suppressed at the expense of signal distortion. Examples of time weighting windows are the Hamming- and Blackman windows.

Kaluzynski (1989a) has researched the different spectral estimation methods as applied to Doppler ultrasound. His studies included the periodogram, autoregressive (AR) modelling, the maximum likelihood method and the

Wigner-Ville distribution. He concluded that for Doppler spectra the method of maximum likelihood and AR spectral estimation produced the best results.

The selection of the order of the AR process is presented in a follow up study done by Kaluzynski (1989b) (See also Evans and Schlindwein 1990). The study described in this thesis concentrates on AR spectral estimation, because of its relative superiority as a method of spectral estimation and the improvement in spectral resolution obtained (Kaluzynski 1989a).

The principle behind AR spectral analysis is a simple yet powerful one. Consider a system with input $p(z)$, output $x(z)$ and system transfer function $H(z)$, such that

$$H(z) = \frac{x(z)}{p(z)} \quad (2.2)$$

Assume that we had measured the output signal, $x(z)$, and we propose to calculate the spectral content thereof. AR spectral analysis seeks to find that system, $H(z)$, which when driven by Gaussian white noise, $p(z)$, delivers an output equal to $x(z)$. Note that the process is firstly a means of system identification. Using our acquired knowledge of the system, $H(z)$, we can estimate autocorrelation points outside that region from which we have taken direct measurements (With a unit function as input and the system transfer function, $H(z)$, known we can calculate the output to any number of points by application of the discrete difference equation.). Spectral analysis is completed simply by the application of the Wiener-Khintchine theorem i.e. the Fourier transform of the autocorrelation sequence.

Evans and Schlindwein (1989) proposed a real-time autoregressive spectrum analyser based on a TMS 320C25 digital signal processing processor. The AR spectral estimation technique that was employed proved computationally very efficient. Evans and Schlindwein highlight the inherent trade off between sampling frequency, window length and processing power. During system optimisation careful consideration is given to allow all frequencies under consideration to develop in the given time frame to improve the statistical properties of the estimate.

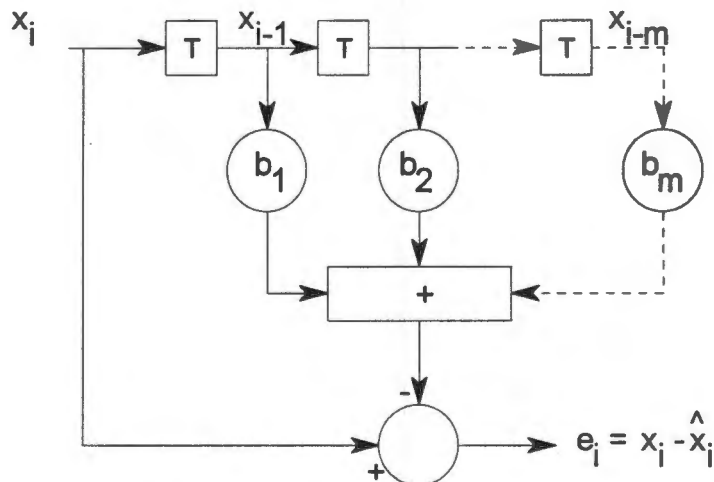
Evans and Schlindwein (1990) applied four different criteria for the selection of the order i.e. first zero crossing (FZC), final prediction error (FPE), Akaike's information criterion (AIC) and the Criterion autoregressive transfer-function

(CAT). The study concluded that in 98 % of the 1280 frames analysed the order selected was ten or less and that for a time domain window of 256 samples the FPE would suffice as selection criteria for the process order.

2.1.1.1 AR SPECTRAL ESTIMATION

As can be seen from the preceding paragraphs AR spectral estimation plays an important role during waveform analysis. Although the mathematical foundation of this technique is well documented it is important to consider briefly the underlying theory behind this approach (Readers who are not familiar with parametric estimation techniques may wish to skip this section).

The Durbin algorithm used during the AR process is a recursive algorithm for performing least square estimates (Giordano and Frank 1985). It allows the received signal to be estimated from a sequence of previously received signal samples. Figure 2.1 serves to illustrate.



Linear filter as a tapped delay line
figure 2.1

Assume that we have an all pole system. A signal is observed and it consists of the transmitted signal plus noise. An estimate of the received signal may be obtained from delayed versions of the received signal linearly weighted by prediction coefficients. If we minimise the mean square error between the received signal and its estimate then we are left with a set of equations whose solution is the prediction coefficients.

Let x_i and s_i denote the complex low-pass equivalent of the received and transmitted signals taken at the i th sample respectively. The received signal is expressed as

$$x_i = s_i + n_i \quad (2.3)$$

for $i = 1, 2, \dots, N$

where n_i is assumed to be stationary white noise and N is the number of signal samples. Assuming that the signal is stationary, x_i , may be predicted from x_{i-1} , x_{i-2} , ..., x_{i-M} . It follows that

$$\hat{x}_i = \sum_{m=1}^M b_m x_{i-m} \quad (2.4)$$

where b_m are the coefficients of the linear predictor, M is the filter order and \hat{x}_i is the predicted value of x_i .

$$\begin{aligned} P_M &= \varepsilon \left[|e_i|^2 \right] \\ &= \varepsilon \left[\left| x_i - \sum_{m=1}^M b_m x_{i-m} \right|^2 \right] \end{aligned} \quad (2.5)$$

where e_i is the error signal at the i th sample, ε is the ensemble operator and P_M the mean square estimate of the error signal.

$$e_i = x_i - \hat{x}_i \quad (2.6)$$

Since the estimate of the received signal attempts to remove the predictable portion of the received signal, the error signal is approximately a white noise process. Equation (2.6) can be expressed as

$$e_i = \sum_{m=0}^M a_m x_{i-m} \quad (2.7)$$

where a_m are the prediction error filter coefficients given by

$$a_0 = 1, a_m = -b_m \quad m \neq 0 \quad (2.8)$$

Substituting of (2.8) into (2.5) leads to

$$\begin{aligned}
P_M &= \sum_{m=0}^M \sum_{j=0}^M a_m a_m^* \varepsilon[x_{i-m} x_m^*] \\
&= \sum_{m=0}^M \sum_{j=0}^M a_m a_j^* r_{j-m}
\end{aligned} \tag{2.9}$$

where

$$r_{j-m} = \varepsilon[x_{i-m} x_{i-j}^*] \tag{2.10}$$

is the ensemble average autocorrelation coefficients. Minimisation of P_M with respect to the predictor coefficients a_l^* leads to the set of linear equations.

$$\frac{\partial}{\partial a_l^*} P_M = 2 \sum a_m r_{l-m} = 0 \quad l=1, \dots, M \tag{2.11}$$

Equation (2.11) can be re-expressed in a form known as the Yule-Walker equations.

$$\sum_{m=1}^M b_m r_{l-m} = r_l \quad l=1, \dots, M \tag{2.12}$$

The Durbin algorithm (Code implementation shown in **Appendix III L4.0392**) is an order recursive method for solving (2.12). That is, it solves the coefficients of an M -order predictor recursively from the coefficients of an $(M-1)$ -order predictor.

Assuming the input samples have zero mean Eq. (2.12) may be solved by first calculating the autocorrelation estimates.

$$R_l = \frac{1}{N} \sum_{r=0}^{N-l-1} x_r x_{l-r}^* \quad l=0, 1, \dots, M \tag{2.13}$$

For a stationary process the covariance matrix $\{r_l\}$ is Toeplitz, so that the coefficients b_m , are recursively computed as

$$b_{M,p} = b_{M-1,p} - b_{M,M} b_{M-1,M-p}^* \tag{2.14}$$

for $p=1, 2, \dots, M-1$ where

$$b_{M,M} = \frac{A}{B}$$

$$A = R_0 - \sum_{l=1}^{M-1} b_{M-l,l} R_{M-l} \quad (2.15)$$

$$B = R_0 - \sum_{l=1}^{M-1} b_{M-l,l}^* R_l \quad (2.16)$$

If the Fourier transform of the prediction coefficients is defined as

$$A(f) = \sum_{i=0}^M a_i e^{-j2\pi i f T} \quad (2.17)$$

then the PSD may be estimated as

$$\hat{P}(f) = \frac{c}{A(f) A^*(f)} \quad (2.18)$$

where c is a constant and T the sampling frequency.

A more explicit representation of the spectrum in terms of an all pole (AR) model can be obtained by substituting (2.17) into (2.18) using $a_0 = 1$.

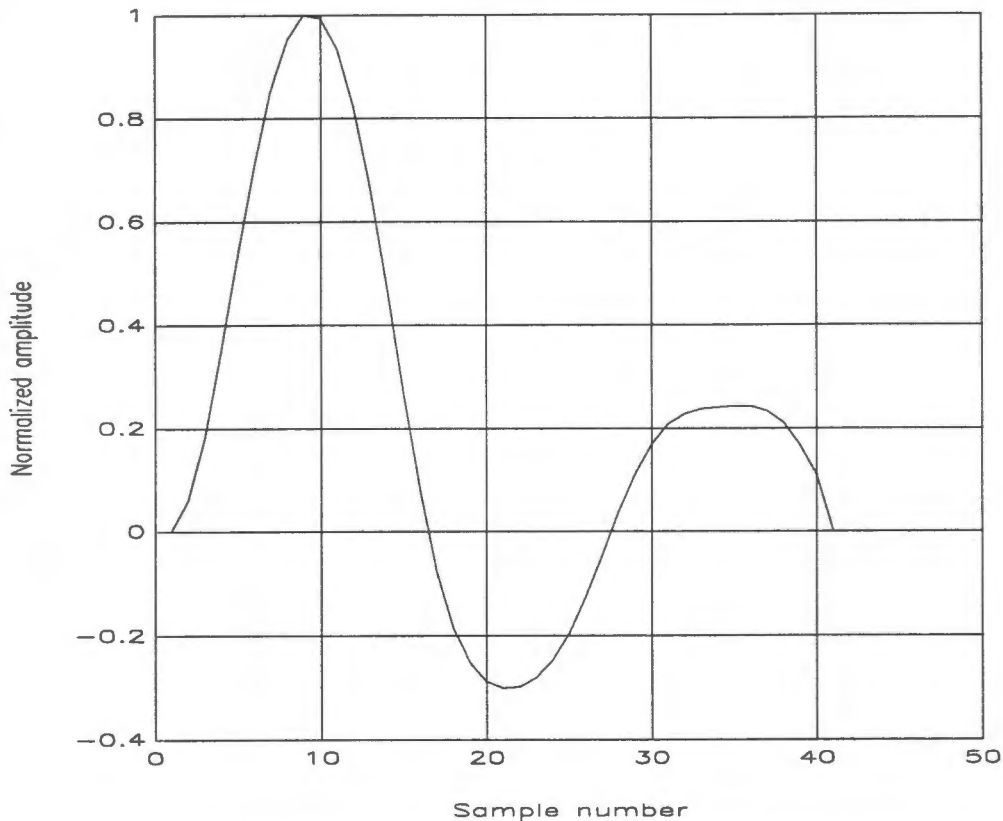
$$\hat{P}(f) = \frac{c}{\left| 1 + \sum_{i=1}^M a_i e^{-j2\pi i f T} \right|^2} \quad (2.19)$$

The power spectral estimate is then obtained using Eq. (2.14) with $b_m = -a_m$.

2.1.2 BLOOD VELOCITY WAVEFORM ANALYSIS

Following the process of PSD estimation the blood velocity waveform is constructed through the calculation of a representative blood velocity value for each consecutive time frame. The two methods most commonly employed are the mean- and peak velocity processors (Skidmore et al 1989). The mean velocity is calculated through the identification of that point, in the frequency domain, which has an equal amount of energy to the left and right. Note that the absolute value thus calculated has only a relative importance because of the angle dependence as explained in Appendix I. A criticism of the peak velocity process is the inherently noisy environment in which measurements are performed. This can be clearly understood when considering the turbulent nature of blood flow.

The velocity waveform thus calculated has, given a normal arterial pathology, a characteristic tri-phasic shape. The waveform shape changes in the event of arterial disease and/or under conditions which act as a disturbance to normal flow.



**Tri-phasic shape of a normal velocity waveform
figure 2.2**

Early studies have shown the loss of pulsatility of a velocity waveform in the presence of arterial disease (Fitzgerald et al 1971). The pulsatility index (PI) as a quantitative measure of disease was first described in terms of the signal's Fourier components, but the technique was later simplified (Gosling and King 1974) to the ratio of the peak to peak and mean velocities.

The variation of the velocity profile in the post-stenotic field gives rise to a broadening of the spectral estimate. The use of the spectral broadening index (SBI) as a quantitative index of turbulent flow has however not been popular. This is partly due to the dependence of the frequency variation on the probe angle and the inherent shortfall of spectral methods applied. It is shown (Wijn et al 1987) that the SBI depends strongly on the insonation of the vessel and

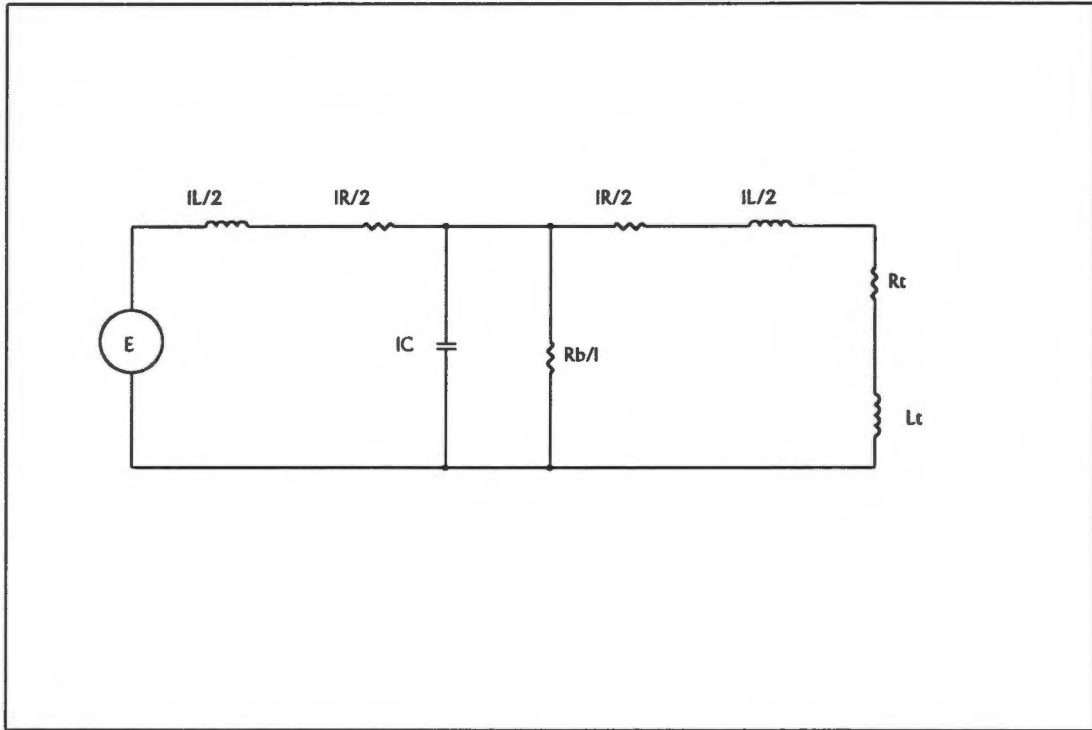
geometry of the ultrasound beam. In intra-operative measurements a large variation is possible due to near field (Fresnel region) operation of the Doppler probe.

The transfer function approach (Brown et al 1978) relates the frequency transforms of measurements taken at two separate sites, namely the proximal end of the femoral and popliteal arteries. The former is considered as the input to the arterial segment and the popliteal waveform the output. A second order system was formulated to describe the blood flow in the arterial segment and the roots of the characteristic equation solved. The pole positions are indicative of the segment's physical state. A severe criticism of this approach is that the input waveform does not have a sufficiently broad bandwidth to be used for system identification and the non-linearity in the arterial segment can lead to the prediction of an unstable system, which is clearly not the case. As such the use of this method is not popular in clinical application.

The method of principal component analysis (PCA) is similar to Fourier analysis in that the waveform is broken down into its principal components, but a distinct difference exists in that the components are not necessarily sinusoidal. A minimum number of components is sought to best describe the velocity waveform. As this complex method has little advantage over the simple PI it too has remained dormant in the research environment.

More complex methods, such as the Laplace transform analysis (LTA) (Skidmore 1979), have been developed which relate the waveform shape to the arterial physiology. In LTA a simple elastic tube model was developed to represent the arterial tree from the heart down to the periphery. The electrical analogy of this model was then used to develop an equation which relates the flow of current (analogous to blood flow) to a stimulating voltage (analogous to cardiac pressure pulse).

Figure 2.3 shows the simple electrical analogue model of the arterial system. L , R , and C represent the inertia of blood, resistance to flow, and wall compliance, respectively, of the main aorto-iliac trunk of effective length l . R_b represents the resistance of all vessels which run off the main trunk. R_t and L_t are the resistance to flow and inertia in the terminal branch which represent the runoff resistance distal to the inguinal ligament. E represents the impulse delivered by the heart.



**Electrical analogue of the arterial system
figure 2.3**

With numerous simplifications the model can be reduced to a third order system where parameters relate to arterial wall elasticity, distal impedance and proximal arterial diameter. If the input voltage to the model is assumed to be an impulse then the pole positions on the Argand diagram can be used as an indication of arterial disease.

The system transfer function may be written as

$$H(s) = \frac{k}{(s^2 + 2\alpha s + \omega_0^2)(s + \gamma)} \quad (2.20)$$

where :

$$k = \text{constant} \quad (2.21)$$

$$\alpha = \frac{R}{2L} + \frac{1}{2CR_b} = \omega_0 LTD \quad (2.22)$$

$$\omega_0 = \frac{1}{l} \sqrt{\frac{2}{CL}} = LTE \quad (2.23)$$

$$\gamma = \frac{R_t}{L_t + \frac{LI}{2}} \quad (2.24)$$

and

$$LTR = |\gamma| \quad (2.25)$$

The parameters of the model are calculated from a measured flow waveform by first calculating the spectral content of the waveform and then using a curve fitting procedure in the frequency domain. This model attempted to isolate the different physiological variables and as such seems a very attractive option but several authors have questioned the validity of this approach (Law et al 1984 and Johnston 1984). It was shown by these authors that the flow waveform is a function of peripheral resistance but that the model does not account for continuous flow during the cardiac cycle, as might be the case with distal vasodilation.

A follow-up study by Skidmore et al 1989 conceded that further model refinement should be done but proved that the presence or absence of superficial femoral occlusion did not affect the prediction of iliac disease. The LTA proved a success in detecting stenoses of greater than 50 % with a sensitivity of 92 %. Two areas of concern are stated. First, the LTR variable was proven to be without diagnostic value and secondly that the curve fitting procedure used has difficulty in dealing with positive or negative offsets at the origin i.e. the problem of continuous flow during a given cardiac cycle.

Capper et al (1986) employed an analog system to obtain the mean frequency waveform for analysis. The Laplace Transform parameters and pulsatility index are calculated and compared with angiographically determined aorto iliac disease and pull through measurements. From this pilot study LTD performed better than both the PI and the ankle systolic pressure index (ASPI = systolic pressure at ankle / systolic pressure at brachial artery) in separating significant stenosis from mild disease.

Meara (1984) presented a non-linear regression technique to analytically solve the characteristic equation as proposed by Skidmore. An inverse Laplace transform is applied to Skidmore's frequency based equations. The discrete time equivalent equation thus obtained is used in this study and will be discussed more comprehensively in following sections.

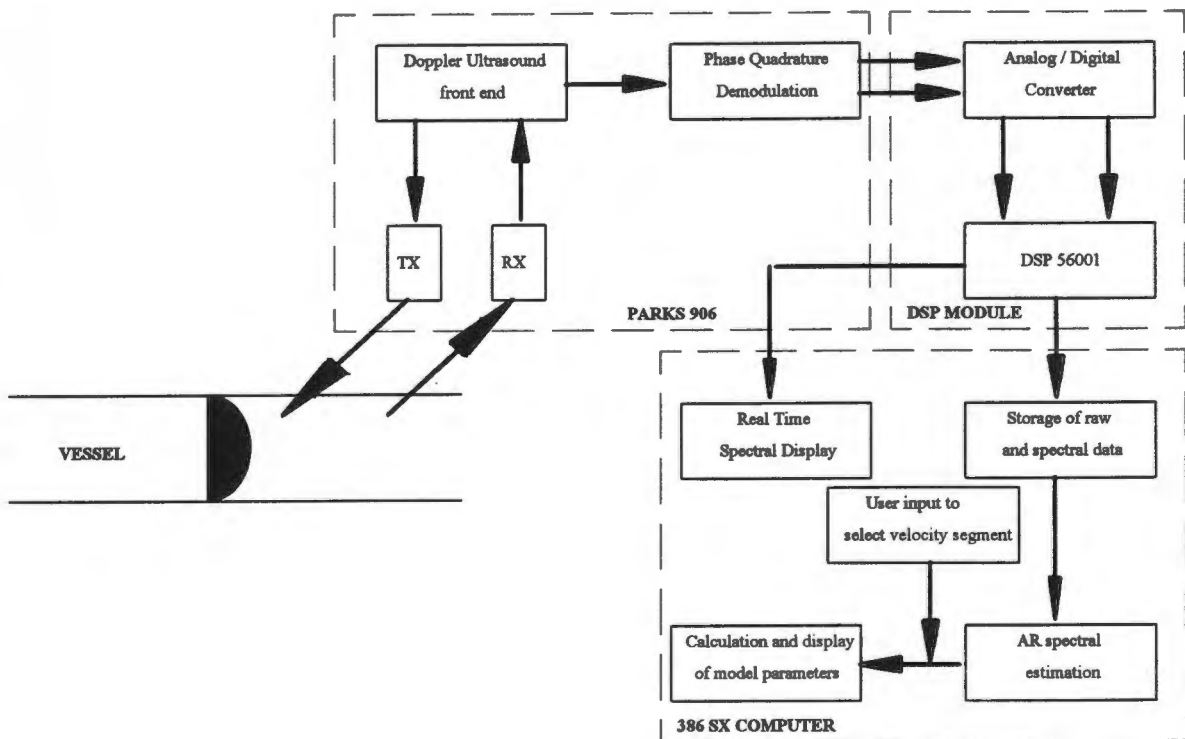
Other parametric methods such as autoregressive (AR) models have proved useful in classification of disease (Alessio et al 1983), being specific in discrimination between signals relating to different pathological states but they have the inherent shortcoming that the parameters extracted have no known physiological meaning.

A great deal of research has been done at the UCT/Groote Schuur Department of Biomedical Engineering and at the Faculty of Medicine, University of Stellenbosch (Bunn and Guelke 1983), into flow in compliant tubes in terms of system parameters such as tube compliance, wall losses and fluid inertia. The approach adopted by these authors is that of an electrical transmission line analogy with specific reference to wave propagation as a function of frequency. This approach has been used to explain cochlea function (Bunn and Guelke 1983, Guelke and Bunn 1984) and the input impedance of the lungs (Capper et al 1990). Applying this resource pool of knowledge to femoral studies necessitated using models which have a direct relationship to arterial pathology.

CHAPTER THREE

3.1 HARDWARE

As real-time implementation and processing is a high priority the hardware has been chosen to facilitate computational power and speed. A schematic representation of the development system is shown.



System data flow diagram
figure 3.1

3.1.1 ULTRASOUND UNIT

The CW Doppler input to the system is from a PARKS 906 bi-directional unit. This unit serves to present two demodulated phase quadrature signals to the DSP unit from which the forward and reverse flow components are calculated. Integrated in the PARKS 906 is the demodulation from the high frequency carrier to baseband audio.

3.1.2 MICRO-CONTROLLER ENVIRONMENT

The DSP board, designed to fit into an IBM compatible PC bus, fulfils the digital processing needs of the system and as such plays an integral part in real-time implementation. The basic composition is a four-channel multiplexed input, 12-bit 1MHz A/D (analog to digital) converter, on board memory and a Motorola DSP56001 fixed point processor. The dual natured architecture of the processor (separate program and data memory space) and the parallel nature in which the program controller, the address generation unit and the ALU operates facilitates benchmarks such as a 1024 point complex FFT calculation in 3.2 ms.

3.1.3 HOST PLATFORM

A PCM 386SX performs the co-ordinating function of the system. This 16MHz PC utilises the following configuration; a 40Mb HDD, 1.2Mb FDD, 2Mb 0 wait state Ram, 16-bit VGA graphics card with 512kB Video Memory. A multi-frequency colour VGA monitor with a resolution of 640 X 480 X 256 colours serves as the VDU.

3.2 SOFTWARE IMPLEMENTATION

3.2.1 MICRO-CONTROLLER ENVIRONMENT

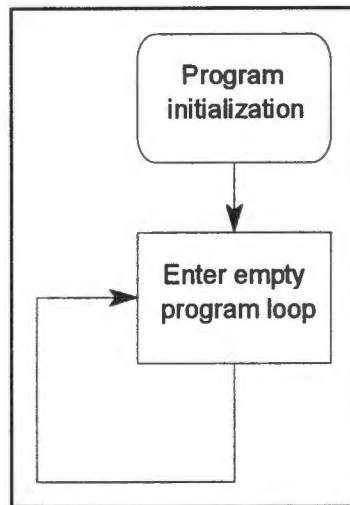
Real or near real time spectral analysis requires considerable processing power which is made feasible with high speed DSP processors. The Motorola DSP56001 fixed point processor used in this project is rated at 10 MIPS (Million instructions per second). To ensure that the greatest speed advantage is attained, software development is done in Motorola DSP56001 assembler.

A complete listing of the DSP assembler code developed is shown in Appendix III. Note the line numbering (Lx.xxxx) system employed. Program comments and debugging lines start with a '%' for MATLAB files, ';' for the assembler files and '/' for the C source code. An explanation of the program philosophy follows.

The process is entirely interrupt driven as a consistent timing sequence is needed. To this end three separate interrupts are used (Int A - L0.0130, Int B - L0.0149, FrameR Int - L0.0169). Int A and Int B are driven by an external source (8254 programmable interval timer) and the FrameR interrupt is derived from the DSP processor itself. Consider carefully the IPL (Interrupt Priority Level) given to them. The DSP processor has three distinct interrupt priority

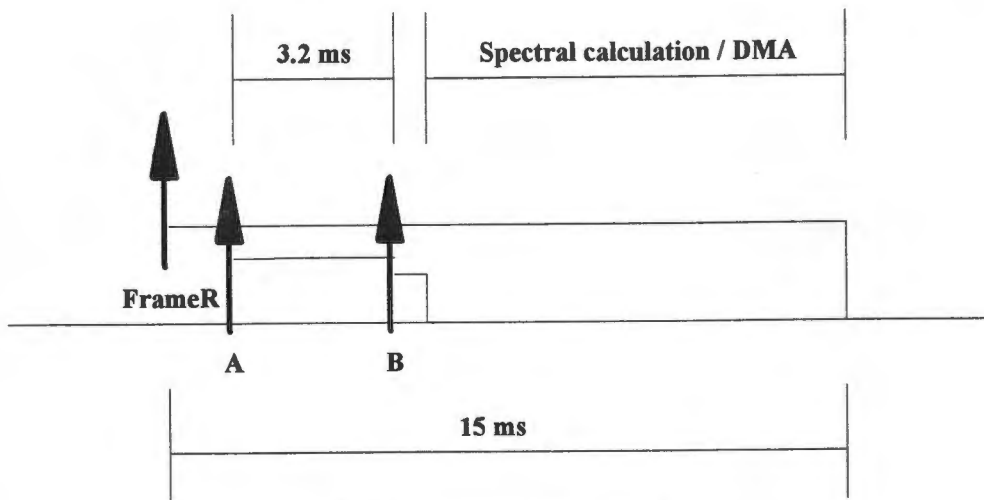
levels. Levels zero through two may be selected by the user. Masking of all lower level IPL's are employed with IPL3 (e.g. a processor reset) having the highest priority.

After program initialisation the program enters an empty loop which allows the interrupts to take control. The following flow diagram serves to illustrate the global perspective.



**DSP program loop
figure 3.2**

As the sampling frequency is known a priori the FFT-, Hamming window- and carrier lookup table are calculated during program initialisation. The carrier table is used during the demodulation phase of the phase quadrature channels (Appendix II). At this time the interrupt priority is set to IPL3 (L0.1178) which disables all user selectable interrupts. The IPL levels are then set as follows FrameR - IPL2, Int A - IPL0, Int B - IPL0 and the host port to IPL1 (L0.1186). Before the main program loop (L0.1210) is entered the interrupt priority is lowered to IPL2 to allow FrameR to start. The sequence which follows is illustrated below.



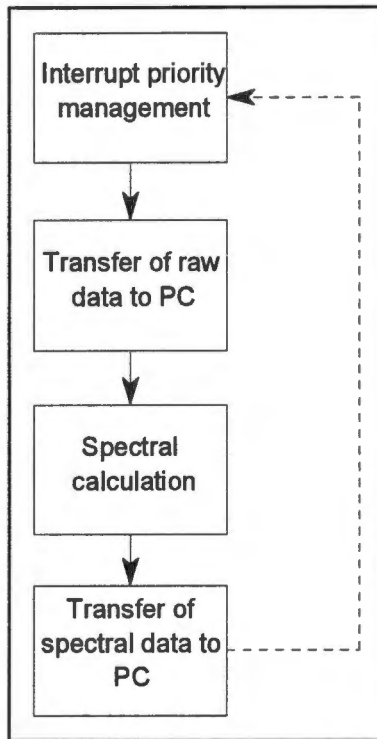
DSP interrupt schedule
figure 3.3

The FrameR interrupt, which is driven by the internal DSP clock, has a period of 15ms. The interrupt initiates the sampling process by setting the 8254 programmable interval timer (which controls Int A and Int B) and unmasking the lower priority interrupts. This enables Int A to start the sampling sequence. The FrameR interrupt then enters an idle state waiting for the IPL to be raised to IPL2 (L0.0175).

Int A samples both input channels into X and Y memory space respectively. During this process the software gain control is introduced through the right shifting of sample data (L0.0135 and L0.0143).

Note that when an interrupt occurs the IPL is automatically raised to disallow any interrupt of the same or lower IPL to interrupt. This process is reversed on interrupt exit (e.g. during the sampling process the IPL alternates between IPL0 and IPL1 when the Int A sequence is first entered, executed and then returned from). When Int B occurs after 3.2ms (i.e. between sampling interrupts) the sampling process is effectively stopped. To prevent Int A from restarting after Int B has finished, we change the exit priority (to IPL2) as stored on the system stack (L0.0150 - L0.0156).

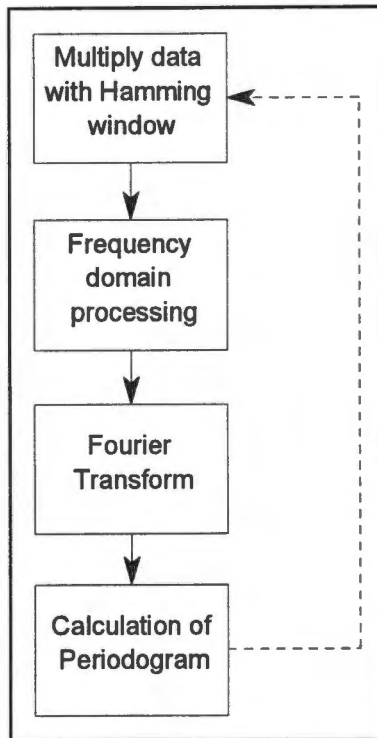
This allows the FrameR interrupt to continue and start the spectral calculation and DMA processes.



FrameR interrupt
figure 3.4

To optimise the DMA transfer process (L0.1275) the raw data is transferred as integer values. No signal resolution is lost during this process as the A/D converters are themselves 12-bit entities. The handshaking between the host and the DSP is implemented through the use of flags in the HCR (Host Control Register) and the HSR (Host Status Register). This is an important process (See DSP macro at L0.1275) which ensures that both the PC and the DSP is at all times aware of what the other party is doing. Consider as an illustration of a potential pitfall the transfer of raw data when the PC expects the spectral values.

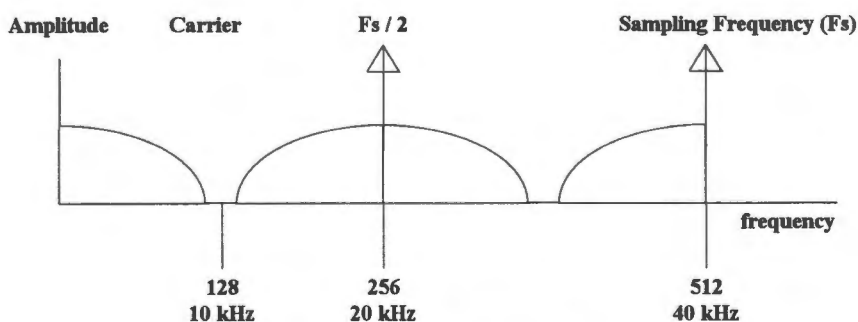
The spectral calculation (L0.1360 - L0.1467) comprises the following steps :



Spectral process flow
figure 3.5

Both direct and quadrature channels are multiplied with a Hamming window to suppress unwanted sidelobes (L0.0356). Frequency domain processing (L0.0382 and L0.0405) follows. This technique (Appendix II) combines the information embedded in the separate input channels into a single memory space. The data is now ready for the application of a single FFT operation which contains all the relevant directional information.

The calculation of the PSD starts with the zero padding from 128 to 512 data points. This process interpolates between data points in the frequency domain. Note that no information is lost during this process. Consider the following diagram :



Spectral distribution
figure 3.6

The replication of the positive and negative frequency bands (Nyquist theorem) and the influence of the frequency domain processing (Appendix II) leaves us with 128 positive and 128 negative data values, offset from the carrier frequency, to display. An FFT and magnitude squared (Periodogram) operation completes the process. The spectral values are then right shifted as only a 8 bit resolution is needed for the spectrogram display. (256 discrete levels proved sufficient for visual discrimination). The advantage gained in this step is the greater speed of 8 bit DMA.

3.2.2 HOST PLATFORM

The PC fulfils a co-ordinating role and provides the user interface. Software development is done in Turbo C and control over the DSP card is maintained from within this environment.

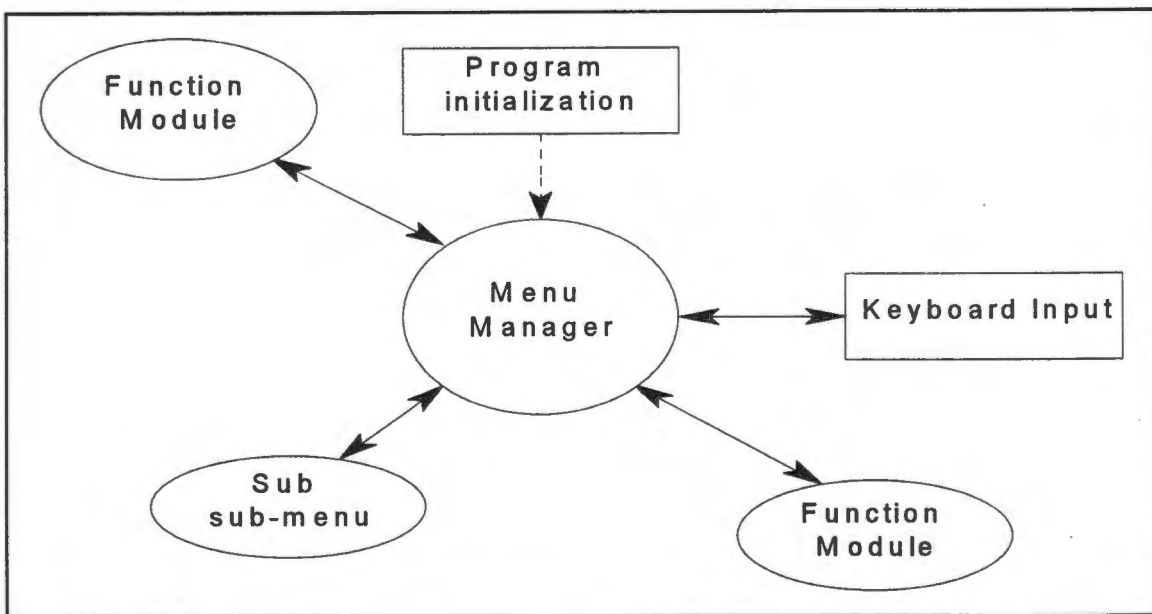
The relative inability of a PC as a real time instrument provided a considerable challenge in ensuring that the DSP be utilised to its full potential. The main culprit proved to be the limited graphics capabilities of the Paradise video adapter. From the outset a decision was made to keep the adapter in graphics mode (640 X 480 X 256 colours) as opposed to continually switching between text and graphics mode. Extensive use is made of the Borland's Graphic Interface (BGI). As the selected mode is not supported by the standard BGI drivers a locally developed software driver specific to the Paradise adapter is used (Paradise.bgi).

The selected mode allows the manipulation of the colour palette and the implementation of a look-up table in which sixteen amplitude ranges are represented by unique colour/intensity combinations. Four colours (Blue, Green, Yellow and Red) are used; each having four different intensity levels. This corresponds to the sixteen amplitude levels. The look-up table is defined with index into the palette starting at colour number sixteen so as not to interfere with the standard IBM defined range for EGA and VGA; zero to fifteen.

To ensure portability and provide a platform for future research work a protocol was established by which data capture and storage could be done. As an aid to this development research done by Evans and Schlindwein (1989) provided valuable input.

A scrolling display provides the user with a real time visual feedback. Upon successful capture the option is provided for permanent storage. A retrieval for comparison or further inspection can be done at any given stage. A record is kept of raw data and the calculated spectrum. This has the advantage of allowing future studies to compare different methods available for spectral analysis and waveform processing.

The system software architecture evolves around a recursive menu manager function (L1.0296 - L1.0324) which, after system initialisation, is called with argument 0 (L2.0076). The significance of this integer is best explained if one considers the MenuWindow struct (L3.0027). All displayable menu items are numbered and provision is made for calling sub levels menus and/or system functions. Menu 0 (L1.0024) is the main menu of the application and from here, given the correct user input, sub menu or implicit function calls are done. As an example consider L1.0028 where the either the Go function or the Menu_Man function (with an integer argument (L1.0029)) is called. This approach facilitates the addition of functionality with a minimum programming effort. A graphical system flowchart is given below.



Program module control
figure 3.7

When considering system data the most important to note is that two multi dimensional arrays are kept for the forward and reverse channels respectively. A global variable row acts as an index to both and is incremented after each

time frame is captured, using the C % (modulus) operator, to allow for a circular buffer structure (L2.0736). All functions have been thoroughly documented (see Appendix III) and can be best understood using this reference provided.

3.2.3 CODE DEVELOPMENT FOR MATLAB

Numerical analysis of the captured data requires complex and intensive computational power. This function is fulfilled by MATLAB. The package supplied by THE MATHWORKS CORPORATION has, in addition to numerous built in functions, a programming structure which allows the user the flexibility to develop his/her own algorithms.

It is proposed that once the waveform analysis is secure and concise, coding will be done in C or DSP assembler and incorporated into the stand alone system. Code development for MATLAB (L4.0000 - L4.0807) is done in a C-like structure which should simplify any cross coding process.

The steps needed to perform waveform analysis (L4.0010 - L4.0029) on given data are as follows :

1. Load raw data as captured (L4.0591 - L4.0598)
2. Calculate AR spectral estimation (**L4.0049 - L4.0090**)
3. Determine blood velocity waveform (L4.0092 - L4.0112)
4. Apply infinite impulse response (**IIR**) Yule Walker filter (**L4.0740 - L4.0770**)
5. Select a velocity segment to analyse (L4.0686 - L4.0710)
6. Calculate discrete model parameters (L4.0114 - L4.0130)

Note : **Steps printed in bold** are optional.

The raw data is stored in a text file to allow for import into MATLAB. Although this is not optimal in terms of disk space used, DOS based applications are limited in their sharing of data. To partly compensate use any shareware application to compress data files which are not currently used.

The spectral estimation is performed using the well known Durbin algorithm (L4.0392 - L4.0440). The model order selection used is 15. Research done by various authors is used as a guide in this regard. As an alternative to this

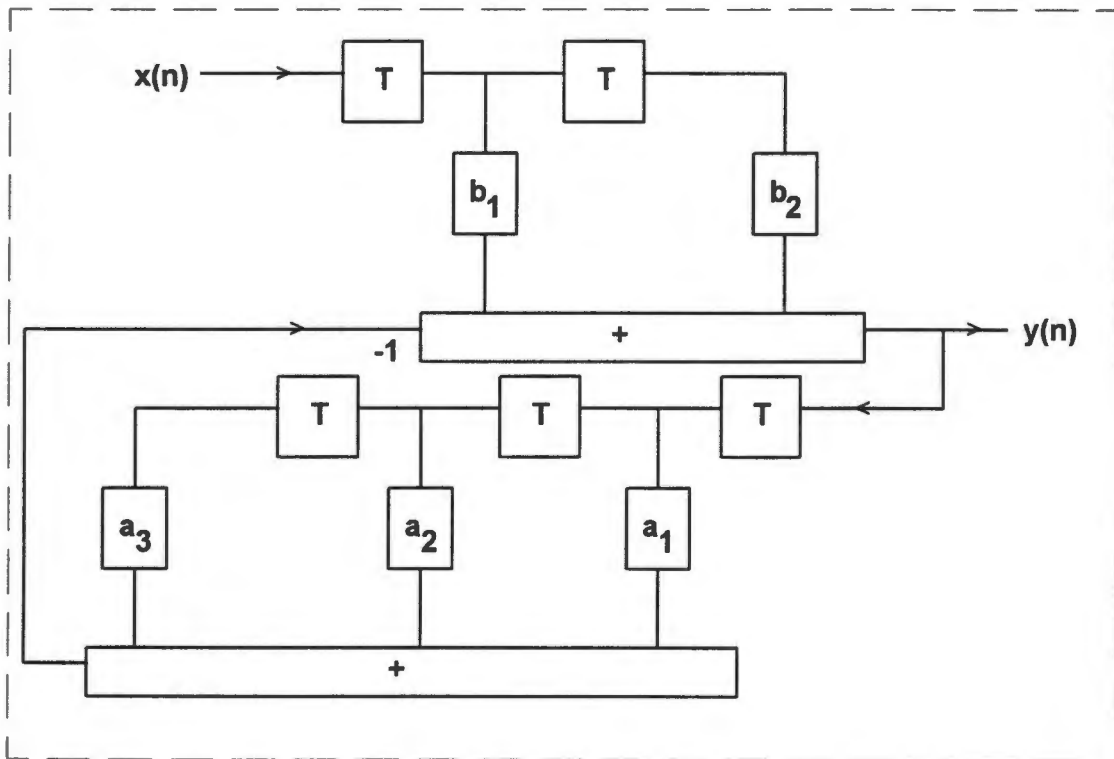
approach the periodogram calculated in real time may be used as a start off point. Pseudo 3-D plots of the calculated spectra are shown in the following chapter.

The mean blood velocity waveform follows from the spectrum. The manner in which we estimate the flow profile is to consider, for each spectral window, that point which has equal amounts of energy to the left and the right. Taking each point thus calculated we construct the waveform profile for the entire period under consideration.

The low pass filtering of the waveform is optional. It has greater importance if Prony's method of system identification is used (L4.0625 - L4.0684). This method is an attractive option for system identification as it assumes an impulse input. Unfortunately this approach is very susceptible to noise variations, a fact stated by Meara et al (1984).

Recall that the objective of the waveform analysis is to identify a system which when driven with an impulse input has a given waveform as output. We therefore need to extract a waveform segment which starts at zero to identify a causal system. (Unfortunately this is not always possible as some physiological phenomena dictate a continuous flow throughout the cardiac cycle.) Using user input as a guide, the start and end of a velocity segment is chosen. The segment amplitude is normalised to unity after subtracting any DC offset component; thus forcing a zero value start. This does not influence the pulsatility of the waveform and is accepted as such until more complete system models are defined.

The discrete equivalent of the continuous s-plane system proposed by Meara et al 1984 is



Discrete system model

figure 3.8

The system transfer function is :

$$\begin{aligned}
 H(z) &= \frac{b_2 z^{-2} + b_1 z^{-1}}{a_3 z^{-3} + a_2 z^{-2} + a_1 z^{-1}} \\
 &= \frac{Y(z^{-1})}{X(z^{-1})}
 \end{aligned}
 \tag{3.1}$$

Calculating the system parameters given the above model is done in an iterative manner. The difference equation form of the transfer function (L4.0136) is fed to the Nelder-Mead algorithm to calculate the unknown parameters. The algorithm calculates the system output and compares the fitted curve to the selected segment using a mean square error (MSE) as criteria. If not within a given tolerance level the parameters are adjusted and the process repeated.

An important point to consider is the initial guess value of the five unknowns. The theory of applied mathematics warns against local minima, a scenario where the algorithm finds a point other than the absolute minimum of the MSE. To overcome this difficulty the initial guess values are taken from the frequency domain Laplacian fit (L4.0132 - L4.0308), a process which in its own right is used for system identification (see Capper et al 1986).

CHAPTER FOUR

4.1 INTRODUCTION

A small pilot study was done on normal individuals and patients on whom peripheral angiograms have been performed. Mean blood velocity waveforms were gathered from twelve normal limbs and twenty-one limbs of patients presenting with lower limb disease. The normal subjects were all healthy non-smokers with no history or signs of arterial occlusive disease. The patients on the other hand all suffered from lower limb arterial disease ranging from about 30% to almost complete occlusion of these vessels. All the test procedures were completed by the technical staff of the Vascular Laboratory at Groote Schuur Hospital.

The chapter presents the results of the investigation and concludes with screen captures, as presented by our system, showing the results of three individuals.

4.2 METHODS

The subjects were all tested in the supine position. A short time period was taken to explain the procedure to them. The emphasis was to reassure the patients of the very low degree of discomfort which they will experience and as such help them to relax and to allow them to reach a steady hemodynamic state. The laboratory technologist applied the ultrasound gel at the site of the femoral artery pulse, slightly distal to the inguinal ligament. (This procedure ensures close coupling of the transducer probe and the measuring site.) With the computer display as visual feedback they captured three (3) seconds of waveform measurements. The waveform analysis was then done off-line, a procedure which takes approximately six minutes per patient.

During the statistical analysis of the subject data, use is made of the Mann Whitney U-test as a quantitative measure of the separation between data groups. The formulae for this index is presented by Mendenhall et al (1986) as

$$Z = \frac{U_A - \left(\frac{n_1 n_2}{2}\right)}{\sqrt{\frac{n_1 n_2 (n_1 + n_2 + 1)}{12}}} \quad (4.1)$$

with

$$U_A = n_1 n_2 + \frac{n_1 (n_1 + 1)}{2} - W_A \quad (4.2)$$

where

n_1 = number of observations in sample A (Normals)

n_2 = number of observations in sample B (Patients)

W_A = rank sum of sample A

After calculation of the Z parameter the p value is read from the appropriate statistical map (see reference).

4.3 RESULTS

The following tables show the processed data for the control group of twelve normals and the data for the patient group consisting of twenty one individuals with angiographically determined aorto-iliac arterial disease.

ID	Frequency domain analysis (Laplace)				Time domain analysis (Meara)			
	LTE	LTD	LTR	MSE	LTE	LTD	LTR	MSE
n1	18.73	0.161	8.434	0.267	17.86	0.162	9.048	0.116
n2	19.12	0.270	12.91	0.527	16.15	0.174	2.533	0.440
n3	20.25	0.293	7.670	0.977	17.93	0.259	6.875	0.335
n4	17.74	0.128	3.758	0.827	16.77	0.128	3.130	0.353
n5	19.17	0.155	4.916	0.682	19.65	0.237	7.527	0.435
n6	17.35	0.146	3.742	0.662	18.06	0.185	4.701	0.583
n7	20.91	0.353	7.650	1.597	19.74	0.371	8.086	0.145
n8	18.89	0.309	6.289	0.939	18.28	0.310	6.168	0.066
n9	19.63	0.227	5.817	1.290	17.84	0.214	5.527	0.366
n10	17.67	0.162	3.588	0.782	16.19	0.115	2.486	0.341
n11	21.21	0.217	4.862	1.112	18.22	0.193	3.734	0.380
n12	18.28	0.167	3.581	0.463	17.22	0.128	2.600	0.096

Processed data for normal group

Table 4.1

As seen from Tables 4.1 and 4.2 (below) the fit proposed by Meara has on average a 64%(((MSE' ave. Meara) - (MSE ave. Laplace))/ (MSE ave. Meara))

advantage over the Laplacian fit in terms of the MSE (Where ave. = geometric mean). Therefore, using the time domain data the statistical properties of the LTD data segment are calculated as :

Mean : 0.206
Median : 0.189

ID	Frequency domain analysis (Laplace)				Time domain analysis (Meara)			
	LTE	LTD	LTR	MSE	LTE	LTD	LTR	MSE
p1	21.77	0.630	3.034	1.139	14.56	0.421	2.328	0.375
p2	26.16	0.799	7.790	1.953	12.76	1.000	41.64	0.820
p3	15.34	0.920	11.70	0.908	14.01	0.403	4.189	0.101
p4	11.76	0.879	34.85	0.674	14.50	0.482	6.107	0.101
p5	29.50	0.386	12.03	1.404	29.81	0.437	13.77	0.315
p6	19.60	0.438	37.10	1.248	19.67	0.419	30.81	0.142
p7	24.01	0.587	8.606	0.149	24.88	0.611	8.853	0.051
p8	16.14	0.875	457.5	1.109	15.99	0.837	442.9	0.106
p9	18.89	0.309	6.289	0.939	18.28	0.310	6.168	0.066
p10	30.24	0.485	14.93	0.231	28.85	0.456	13.87	0.016
p11	13.20	0.703	5.906	0.315	11.34	0.769	7.024	0.113
p12	9.450	0.972	110.2	0.651	10.04	0.998	38.8	0.053
p13	9.510	0.982	101.8	0.641	9.188	0.935	94.58	0.064
p14	18.05	0.512	4.308	0.595	17.47	0.453	3.873	0.045
p15	18.68	0.516	5.268	0.444	15.70	0.497	4.765	0.136
p16	12.69	0.813	62.59	1.884	19.50	0.831	12.43	0.255
p17	24.86	0.485	12.42	1.886	16.82	0.223	4.661	0.670
p18	12.39	0.885	2661	1.042	12.00	0.887	2696	0.149
p19	32.71	0.836	8.621	0.730	16.45	1.000	34.24	0.207
p20	13.86	0.487	714.6	2.074	12.65	0.487	722.5	0.659
p21	12.05	0.645	72.23	1.211	15.30	0.478	9.539	0.281

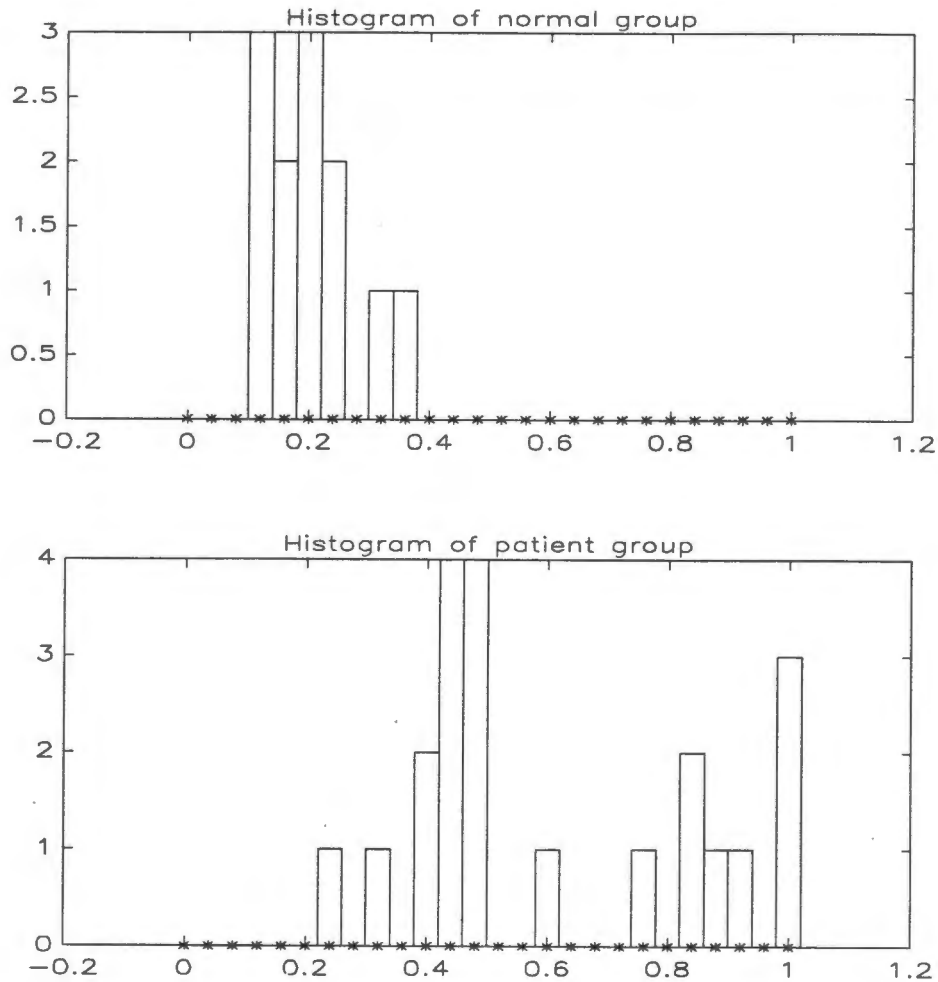
**Processed data for patient group
Table 4.2**

Using the time domain data the statistical properties of the LTD data segment are calculated as :

Mean : 0.616
Median : 0.487

The median and quartile range of the damping factor for the normal group is 0.19, 0.15-0.25 respectively, while that for the patient group is 0.49, 0.45-0.89 respectively. Separation between the two groups was tested using the Mann Whitney U-test and was found to be highly significant ($p < 0.001$). The

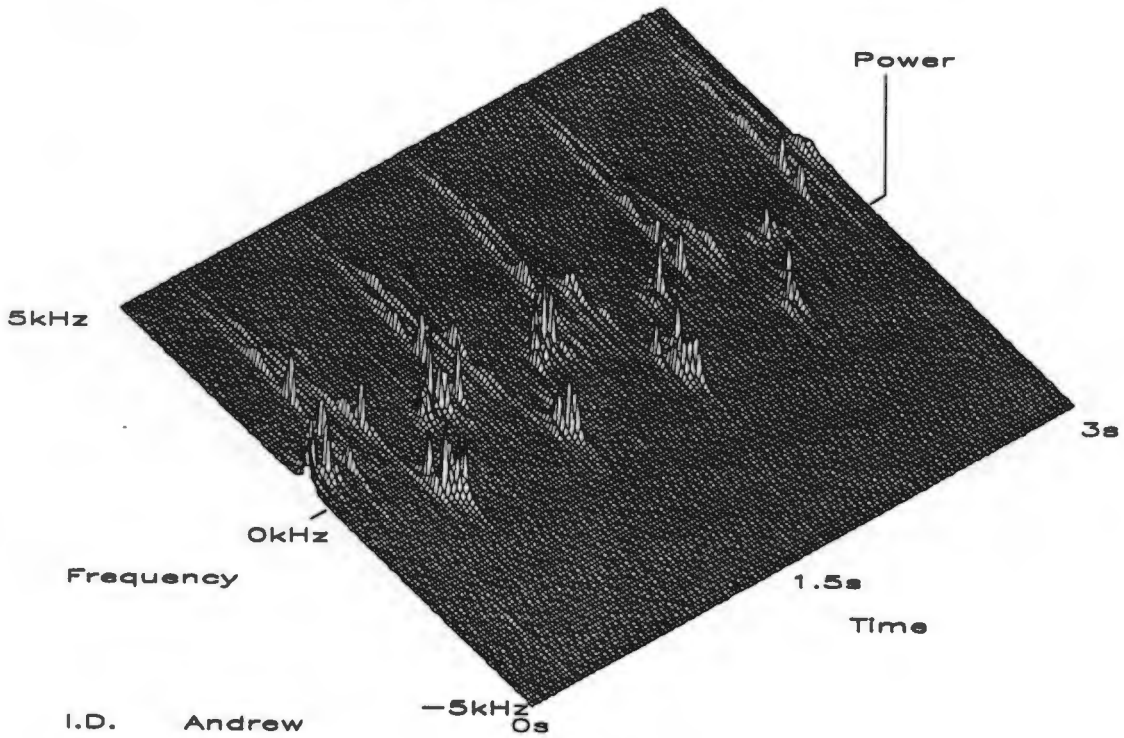
Laplace Transform Damping probability distributions of the two groups are shown in figure 4.1. (Where Y-axis = patient number, X-axis = LTD value).



Subject data distribution
figure 4.1

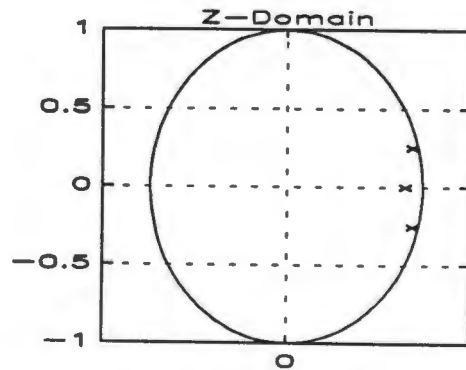
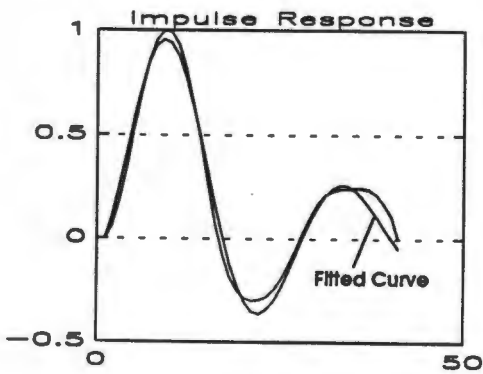
The following graphical displays are of screen captures as presented to the clinician. The first display is of the forward and reverse frequencies as captured. The blood velocity waveform and the curve fitted to this curve are shown in the graph titled **Impulse Response**. The system poles and zeros are shown on the right, both for the discrete z-domain and the continuous s-domain. The system parameters that were calculated are shown together with the values for the LTD, LTR, LTE and the PI. (The figure in brackets adjacent to the PI is the PI for the original unfitted curve)

FORWARD AND REVERSE FREQUENCIES



I.D. Andrew

figure 4.2



	LTE	LTD	LTR	PI (5.941)
Neider	17.86	0.162	9.048	6.397
ZDom	Zeros		Poles	
	0.335		0.923+j0.252	
	Infinity		0.923-j0.252	
	Infinity		0.871	
SDom	Zeros		Poles	
	-72.08		-2.905+j17.62	
	Infinity		-2.905-j17.62	
	Infinity		-9.048	
MSE	0.116			
I.D.	Andrew			

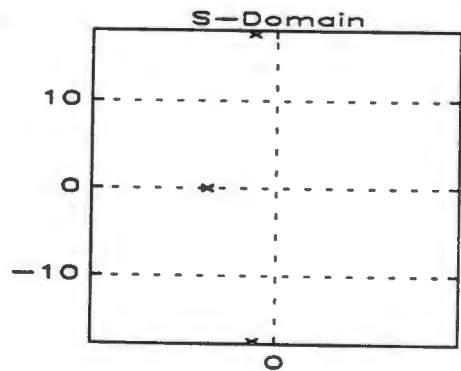


figure 4.3

The graph shown in figure 4.2 and 4.3 is of a normal healthy 25-year old male. The individual is active in long distance running at a competitive level.

FORWARD AND REVERSE FREQUENCIES

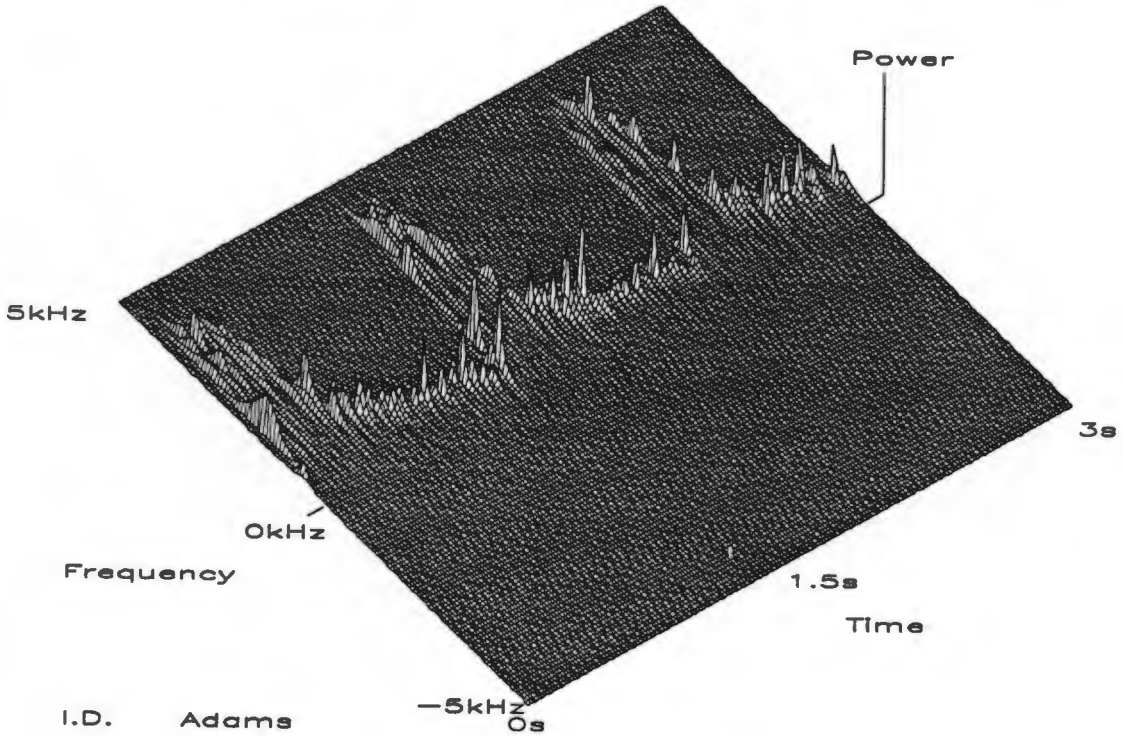
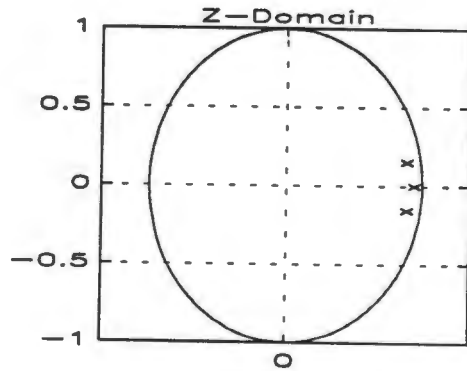
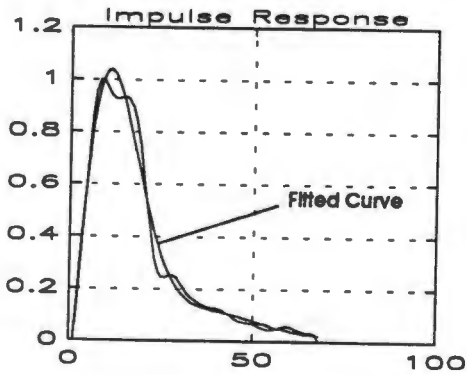


figure 4.4



	LTE	LTD	LTR	PI (3.192)
Nelder	13.4	0.525	3.926	3.314
ZDom Zeros			Poles	
	0.748		0.942	
	Infinity		0.885+j0.154	
	Infinity		0.885-j0.154	
SDom Zeros			Poles	
	-19.14		-3.926	
	Infinity		-7.048+j11.4	
	Infinity		-7.048-j11.4	
MSE	0.131			
I.D.	Adams			

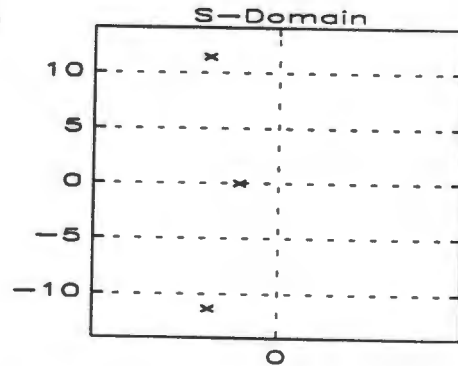


figure 4.5

Figures 4.4 and 4.5 show the results of a 78 year old female (left leg). The peripheral arterial assessment from the department of Radiology reads as follows :

Angiography showed a small infrarenal aneurysm of the abdominal aorta. There was mild disease of all the major vessels of the left leg with a short segment occlusion of the right femoral artery at the adductor hiatus with reconstitution of the popliteals. The trifurcations were patent bilaterally but all crural vessels on both sides were totally occluded shortly below the origin. She was assessed as having unreconstructable disease.

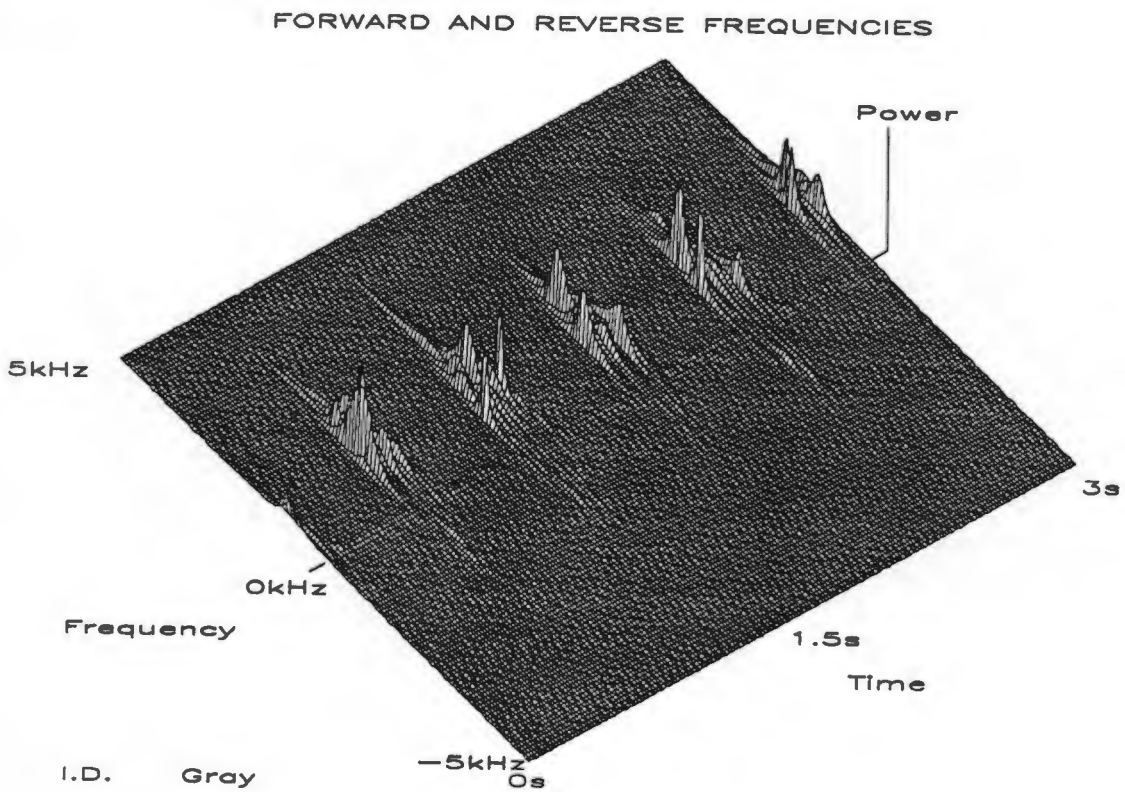


figure 4.6

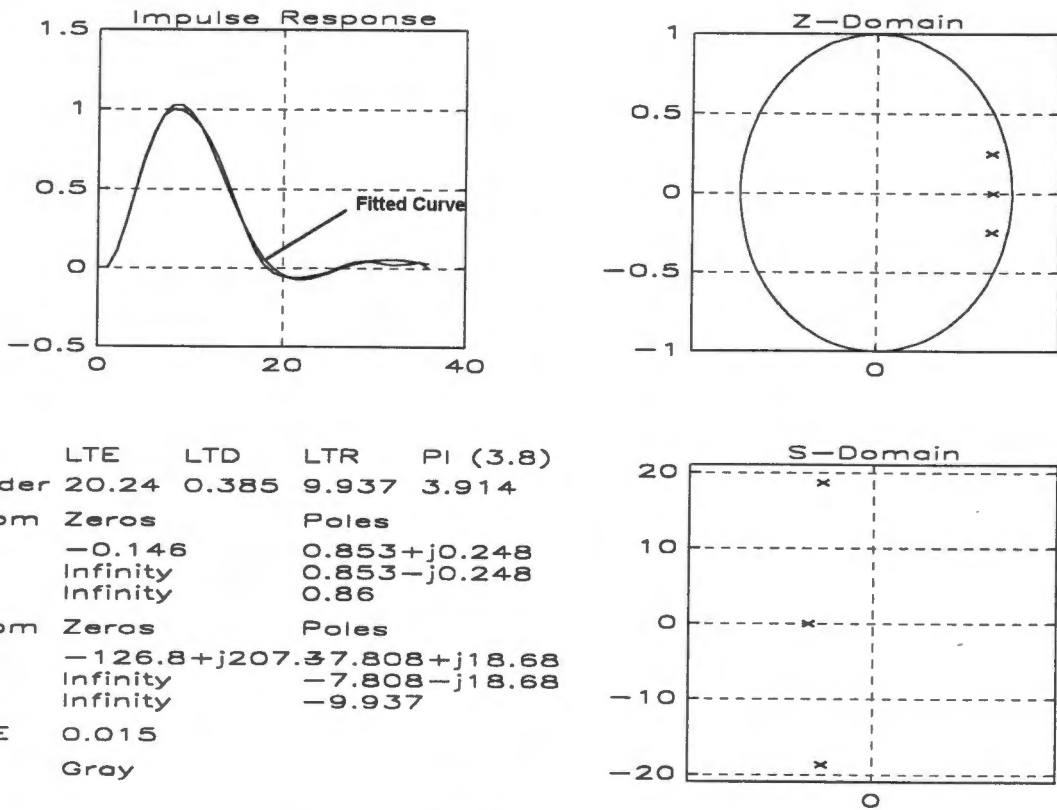


figure 4.7

Figures 4.6 and 4.7 show the results of a 80 year old female (left leg). The peripheral arterial assessment from the department of Radiology reads as follows :

- Aorta : Mild infrarenal atheromatous disease. Mild right proximal renal artery stenosis.*
- Right : Mild common iliac disease. Severe internal iliac disease. Severe profunda/SFA disease with short segment distal occlusion of latter. Heavily diseased popliteal and tight proximal stenosis. Diseased single vessel (peroneal) run-off to ankle.*
- Left : Mild iliac disease. Severe profunda/SFA disease with thigh grade distal SFA stenosis. Diseased trifurcations. Single vessel (peroneal) run-off to ankle.*

CHAPTER FIVE

5.1 DISCUSSION

5.1.1 SYSTEM HARDWARE

The study identified the need for the integration of high speed peripherals. The system hardware that is used has proved to be of an acceptably high quality in its present configuration. With the current trend in operating systems towards graphical user interface based systems the acquisition of a 486-based PC as the host platform should be given consideration.

5.1.2 SYSTEM SOFTWARE

The segmented architecture of the Intel microprocessors used in personal computers today is a major disadvantage if execution speed is of concern. With the application languages and compilers currently only allowing for real mode 16-bit executables all large code and data objects need additional processor overhead to cater for special pointer arithmetic as segment boundaries are crossed. As more sophisticated programming tools become available the application developer will be able to operate the 386-based microprocessors in its protected mode configuration. This will do away with the *segment:offset* pointer arithmetic used in real mode and provide a 32-bit flat address space, with a theoretical limit of 4Giga byte per application. Should the need arise one may recompile the developed code for this mode of operation. This process should prove seamless as the application code in its present state adheres to the ANSI C standard.

5.1.3 SYSTEM MODEL

The analogous electrical circuit model used in this study relates the circuit current (equivalent to mean blood flow) in the terminal branch to the input voltage (equivalent to pressure). The mathematical formula is simplified by assuming the input voltage to be an impulse. The characteristic equation is expressed in the form of poles on the Argand diagram, the poles being the points where the equation becomes infinite. The model has three poles. Two of the poles are a complex conjugate pair, while the third is real. The pole positions express the dynamic behaviour of the system. The two complex poles are used to calculate the damping factor, which relates to the resistance

proximal to the site of measurement and a natural resonant frequency, which is indicative of arterial compliance. The real pole on the other hand relates to the runoff resistance distal to the site of measurement

5.1.4 RESULTS

The use of blood flow waveform analysis to characterise the arterial system has great potential in the non-invasive assessment of arterial disease. This study has highlighted the importance of waveform selection for impulse response analysis. Firstly, the impulse response assumes that the waveform starts at zero. This is not the case when the peripheral resistance is low or when the heart rate is high, as these situations result in a continuous forward flow at the end of diastole. Secondly, fluctuations in the heart rate cause the flow velocity at the end of diastole to differ from that at the beginning of systole. Further work is presently underway to address these problems.

The damping factor has been shown to be a good indicator of lower limb arterial disease, and it is able to separate normals from patients with arterial disease ($p < 0.001$ Mann Whitney U-test). The parameter which is related to runoff resistance is, as expected, highly dependent on peripheral vasodilatation. In the case of distal disease, the decrease in resistance caused by vasodilatation compensates for the increased resistance due to stenoses in all but the very severe cases.

A caution raised by Campbell et al (1984) regarding waveform reproducibility, with specific reference to operator experience, proved most applicable in our study where, although familiar with Doppler CW units the technologist had to be cautioned on the intrusive manner in which measurements were done. With patients that are particularly obese and/or patients with severe occlusions the notion is to apply excessive pressure to obtain waveform readings. As a consequence this acts as a disturbance to the existing flow and inaccurate results are obtained.

5.2 CONCLUSION

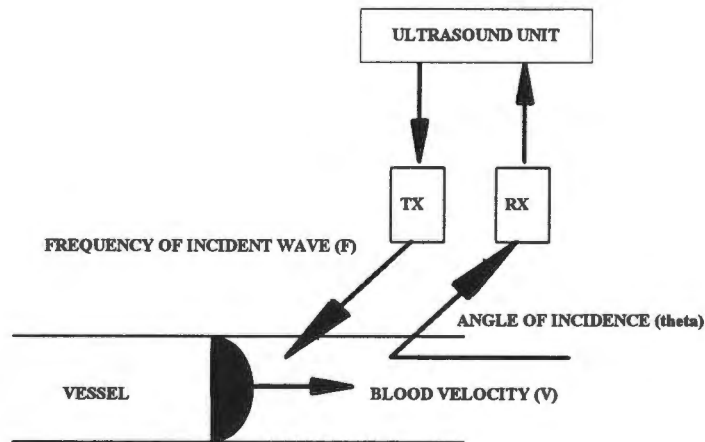
The study is considered to have reached the objectives set. The advantage of having done the initial phase one development has enabled the author to clearly identify potential problems and make recommendations for future research. Proposals which have been made at work group meetings and which may be investigated as post-graduate projects in their own right include research into

the spectral estimation techniques applied to the flow variation, improvement of the Skidmore arterial model and the application of the core unit to different vessel groups. Research in progress involves the gathering of a more comprehensive database and the refinement of the method of waveform selection.

APPENDIX I

DOPPLER PHYSICS

The relative change in frequency due to reflections of acoustic waves off moving structures is well known. A schematic representation of the application to arterial flow is shown below.



Doppler Physics

The change in frequency due to the movement of blood is :

$$\Delta F = \frac{2VF \cos(\theta)}{C}$$

where :

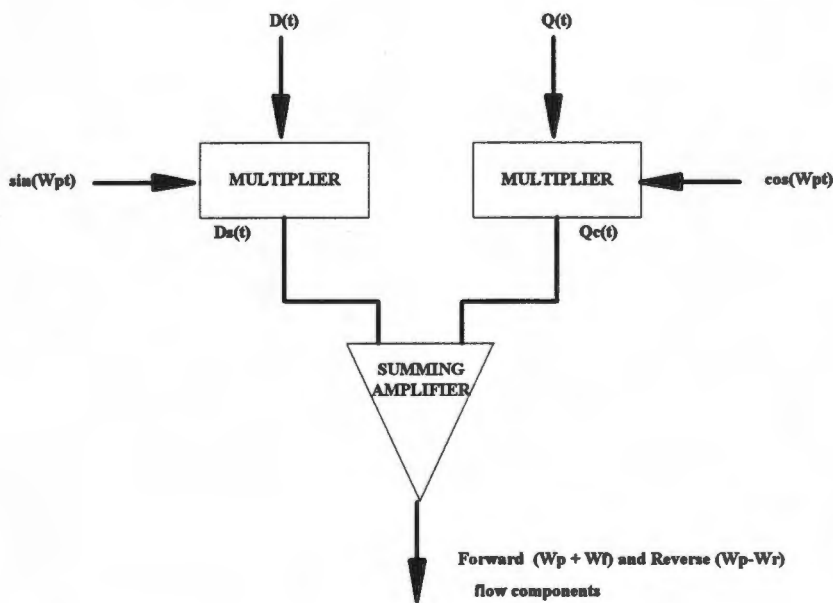
C = velocity of ultrasound in the medium

The transmitted frequency, F , is usually selectable at either 5MHz for deeper vessels or 10MHz for superficial vessels. The Doppler frequency shift falls within the audio band for the range of blood cell velocities found in blood vessels. Since blood cells close to the vessel walls move at a lower velocity than those near the vessel centre a range of Doppler shift frequencies is obtained from any given vessel. These are resolved using spectral analysis of the composite Doppler signal. (Note that it is very important to insonate the total cross-section of the blood vessel during examination).

APPENDIX II

FREQUENCY DOMAIN PROCESSING

In order to distinguish between forward and reverse flow components phase quadrature processing has been used. Frequency domain processing shifts the Doppler frequencies to a pilot frequency. The forward and reverse flow components are then identified from their offset from this reference. The following block diagram serves to illustrate.



Frequency domain processing

If we consider the simplistic example of the simultaneous uniform flow of blood in both a forward and reverse direction resulting in frequency shift of W_f and W_r , respectively (Atkinson and Woodcock, p69),

$$D(t) = \frac{1}{2} [A \cos \Phi_c + B_f \cos(W_f t + \Phi_f) + B_r \cos(W_r t - \Phi_r)]$$

$$Q(t) = -\frac{1}{2} [A \sin \Phi_c + B_f \sin(W_f t + \Phi_f) + B_r \sin(W_r t - \Phi_r)]$$

where :

- B_f, B_r = amplitude of forward and reverse flow respectively
 $W_f \Phi_f$ = frequency shift and phase due to forward flowing blood
 $W_r \Phi_r$ = frequency shift and phase due to reverse flowing blood
 $A \cos \Phi_c$ = DC component due to the demodulation of the carrier

then mathematically the operation alters the direct $D(t)$ and quadrature $Q(t)$ channels as follows :

$$D_s(t) = \frac{1}{2} [A \cos \Phi_c \sin W_p t + B_f \cos(W_f t + \Phi_f) \sin W_p t + B_r \cos(W_r t - \Phi_r) \sin W_p t]$$

$$Q_c(t) = -\frac{1}{2} [A \sin \Phi_c \cos W_p t + B_f \sin(W_f t + \Phi_f) \cos W_p t + B_r \sin(W_r t - \Phi_r) \cos W_p t]$$

$$D_s(t) + Q_c(t) = \frac{1}{2} A (\sin W_p t \cos \Phi_c + \cos W_p t \sin \Phi_c) + \frac{1}{2} B_f [\sin W_p t \cos(W_f t + \Phi_f) + \cos W_p t \sin(W_f t + \Phi_f)] + \frac{1}{2} B_r [\sin W_p t \cos(W_r t - \Phi_r) - \cos W_p t \sin(W_r t - \Phi_r)] = A \sin(W_p + \Phi_c) + B_f \sin[(W_f + W_p)t + \Phi_f] + B_r \sin[(W_p - W_r)t + \Phi_r]$$

It thus follows that the forward and reverse frequencies lie on either side of W_p , the pilot frequency. Note that the greater the reverse frequency the lower its presence on an absolute frequency scale. The clutter component can be easily eliminated due to its fixed frequency location.

APPENDIX III

DSP SOURCE CODE LISTING (0.0000-0.1372)

0.0000 ; *****

0.0001 ; ACQUISITION AND ANALYSIS OF

0.0002 ; LOWER LIMB FLOW WAVEFORMS

0.0003 ; *****

0.0004 ; ASM for Motorola DSP 56001

0.0005 ;

0.0006 ; Copyright(C) L.SMITH

0.0007 ; Department of Biomedical Engineering

0.0008 ; University of Cape Town

0.0009 ; Observatory

0.0010 ; Cape Town

0.0011 ; New South Africa

0.0012 ; (021) 471250 x 235

0.0013

0.0014 ; *****

0.0015

0.0016 ; SAMPLES QUADRATURE INPUTS FROM BI-DIRECTIONAL

0.0017 ; DOPPLER UNIT. VALUES ARE TRANSFERRED VIA DMA TO

0.0018 ; HOST PC.

0.0019 ;

0.0020 ;

0.0021 ; NOTE THE FOLLOWING :

0.0022 ; 1.The ADC is connected to D23..D12. That means that

0.0023 ; the samples taken are in the most significant 12 bits.

0.0024 ;

0.0025 ; 2. With 1 sign bit and 23 significant bits the maximum

0.0026 ; numerical value received by the PC is 8388608.

0.0027 ;

0.0028 ; 3.N Samples are taken per channel at 40kHz sampling rate.

0.0029 ;

0.0030 ; 4.DMA transfer at approx 200 kbytes/s.

0.0031 ; (24 bit transfers are done)

0.0032 ;

0.0033 ; 5.R0 is used to keep the loading addresses of

0.0034 ; channel0 and channel2

0.0035 ;

0.0036 ; 6.IRQA is connected to counter 0 and IRQB to counter 1.

0.0037 ; (Hardware selection on DSP board)

0.0038 ;

0.0039 ; 7.Fractional x and / is NOT the same as integer x and !!

0.0040 ; Care MUST therefore be taken to ensure that overflow is detected

0.0041 ; during INT multiplication.

0.0042 ;

0.0043 ; 8.HF2 and HF0 is used to synchronize DMA transfer to the PC.

0.0044 ;

0.0045 ; 9.The Direct channel is sampled into X and the Quadrature

0.0046 ; channel into Y memory.

0.0047 ;

0.0048 ; PROBLEMS :

0.0049 ; 1.Sampling into anything but the lower 256 positions produces

0.0050 ; a "clipped" waveform. The fast INT and external memory have

0.0051 ; been tested. They are not at fault. The problem prevent us

0.0052 ; from running say sampling and DMA transfers concurrently as

0.0053 ; the "loading" area then contains a corrupted signal.

0.0054 ;

0.0055 ; 2.The FFT algorithm is sensitive to where the data is placed

0.0056 ; in X and Y memory. In addition the FFT argument appears to

0.0057 ; dislike a high memory placement.

```

0.0058 ;
0.0059 ;*****
0.0060
0.0061      opt      fc,mu,s,w,mex ;ASM COMMANDS
0.0062
0.0063 Port      equ      $8000      ;A/D WRITE PORT
0.0064 ADC       equ      $8001      ;A/D READ PORT
0.0065
0.0066 HRX       equ      $FFE8      ;PERIPHERAL REGISTERS
0.0067 HTX       equ      $FFE9
0.0068 HSR       equ      $FFE9
0.0069 HCR       equ      $FFE8
0.0070 PBC       equ      $FFE0
0.0071 SCR       equ      $FFF0
0.0072 SCCR      equ      $FFF2      ;SCI CLOCK CONTROL REGISTER
0.0073 IPR       equ      $FFFF
0.0074 CT0       equ      $9000      ;8254 TIMER
0.0075 CT1       equ      $9001
0.0076 CT2       equ      $9002
0.0077 CREG      equ      $9003
0.0078 Delay     equ      30         ;MUX
0.0079
0.0080 A_Addr     equ      $600        ;FFT ARGUMENT ADDRESS
0.0081 Ca_Addr    equ      $800        ;CARRIER ADDRESS
0.0082 HamA       equ      $1000      ;HAMMING WINDOW
0.0083
0.0084 T0         set     128         ;COUNTER 0 DIVIDE BY (FOR Fs)
0.0085 Fs         set     40e3        ;SAMPLING FREQUENCY (USED IN INTARG)
0.0086 InNum      set     256         ;NUMBER OF SAMPLES RECIEVED
0.0087 Wsp        set     $2000      ;ADDITIONAL WORKSPACE
0.0088 N          set     128         ;SAMPLE NUMBER
0.0089 N_FFT      set     512        ;NUMBER OF FFT DONE
0.0090 Frate      set     $2258      ;FRAMERATE (SEE 11-62) ($2258=15ms)
0.0091           ;(10Hz => 40kHz OR 80Hz => 320kHz)
0.0092
0.0093 ;LOADING AREA
0.0094 WkAddr     set     $0          ;SEE PROBLEM 1 AND RawToPC
0.0095 DMAIn1     set     (WkAddr+InNum-1) ;TEST SIGNAL RECEIVED FROM PC
0.0096 DMAIn2     set     (Wsp+InNum-1) ;TEST SIGNAL RECEIVED FROM PC
0.0097
0.0098 ;DISPLAY AREA (NB!! TRANSFER REVERSED)
0.0099 DsAddr1     set     (WkAddr+$3F)
0.0100 DsAddr2     set     (WkAddr+$40+$3F)
0.0101 RawDisp     set     (WkAddr+N+$7F)
0.0102
0.0103 ;SHIFT INPUT VALUES DOWN BY ...
0.0104 ;REP #0 NOT ALLOWED (IF 0 DESIRED REMOVE ASR IN IrqA)
0.0105 ; ANOTHER OPTION IS THE USE OF
0.0106 ; @CVI(@LOG(N)/@LOG(2)) AS THE MAX VALUE
0.0107 ; RECIEVED FROM A FFT IS A SUMMATION OF N
0.0108 ; NUMBERS AND WILL THUS NOT EXCEED 1.
0.0109
0.0110 In_SHFT     set     3
0.0111
0.0112           org      p:$0000
0.0113           jmp      BEGIN
0.0114           org      p:$08
0.0115           jsr      IntA      ;ATTACHED TO COUNTER 0 (SAMPLING)

```

```

0.0116      org      p:$0A
0.0117      jsr      IntB          ;ATTACHED TO COUNTER 1 (WINDOW TAKEN)
0.0118      org      p:$001C      ;SCI TIMER INT
0.0119      jsr      FrameR      ;FRAMERATE CONTROL (SCI TIMER 11-61)
0.0120      org      p:$20        ;THE FOLLOWING 2 ARE FAST INT ROUTINES
0.0121      movep   x:HRX,x:(r3)- ;RECIEVE VALUES (SEE dmarx)
0.0122      org      p:$22
0.0123      movep   x:(r3)-,x:HTX ;WRITE TO HOST (NOTE DIRECTION OF Rn!!)
0.0124
0.0125      org      p:$0080
0.0126
0.0127
0.0128 ;***** INTERRUPT ROUTINES *****
0.0129 ;ISSUED Fs TIMES PER SECOND (EXECUTION TIME APPROX 12us)
0.0130  IntA  move   #$0A,a1      ;SET UP CH2
0.0131      move   a1,p:Port ;
0.0132      rep   #Delay          ;ALLOW MUX TO SETTLE
0.0133      nop                    ;(APPROX 2.5us DELAY BETWEEN CHANNELS)
0.0134      move   p:ADC,a
0.0135      rep   #In_SHFT      ;SHIFT DOWN !! (SEE LIMITING)
0.0136      asr   a
0.0137      move   a,y:(r0)      ;READ CH0 INTO Y:MEM
0.0138      move   #$08,a1      ;SET UP CH0
0.0139      move   a1,p:Port ;
0.0140      rep   #Delay          ;ALLOW MUX TO SETTLE
0.0141      nop
0.0142      move   p:ADC,a
0.0143      rep   #In_SHFT
0.0144      asr   a
0.0145      move   a,x:(r0)+    ;READ CH2 INTO X:MEM
0.0146      rti
0.0147
0.0148 ;ISSUED EVERY (N*T)sec i.e AFTER EVERY TIME WINDOW CAPTURED
0.0149  IntB          ;IPL RAISED (INT A DISABLED) 8-19
0.0150      movec  ssl,x1        ;DESIRED INT STATE AFTER RTI
0.0151      move   #>$FCFF,a1    ;INT IPL0 NOT ALLOWED
0.0152      move   #>$100,x0
0.0153      and   x1,a
0.0154      or    x0,a
0.0155      move   a1,x1
0.0156      movec  x1,ssl
0.0157      nop
0.0158
0.0159 ; USE THE FOLLOWING IF FrameR INT NOT USED
0.0160 ;      jclr   #3,x:HSR,_HNR ;HF0 (HOST NOT READY TO RECIEVE DMA)
0.0161 ;      jsr   RawToPC      ;TRANSFER RAW DATA
0.0162 ;      jsr   WORK          ;PROCEED WITH CALCULATIONS
0.0163 ;      jsr   ToPC         ;TRANSFER DATA
0.0164 ; HNR      jsr   CNTSetup   ;USE THE FOLLOWING IF FrameR INT
0.0165 ;      jsr   DisCard      ;NOT USED
0.0166 ;      move  #WkAddr,r0
0.0167 ;      rti
0.0168
0.0169  FrameR          ;IPL RAISED TO IPL3
0.0170      jsr   CNTSetup   ;RESET EXTERNAL TIMERS
0.0171      jsr   DisCard      ;INITIAL VALUES FROM ADC
0.0172      move  #WkAddr,r0
0.0173

```

```

0.0174      andi    #\$FC,mr          ;UNMASK INTERRUPTS
0.0175  _busy    jclr    #8,SR_busy    ;WAIT FOR COMPLETION OF SAMPLING
0.0176      jclr    #3,x:HSR,_HNR    ;HF0 (HOST NOT READY TO RECIEVE DMA)
0.0177      jsr    RawToPC          ;TRANSFER RAW DATA
0.0178      jsr    WORK            ;PROCEED WITH CALCULATIONS
0.0179      jsr    ToPC           ;TRANSFER DATA
0.0180  _HNR
0.0181 ;      move    r5,p:Port        ;TOGGLE PORT TO SEE THAT INT WORKS
0.0182      ;NOTE PIN 11 ON IC574
0.0183      rti                    ;SR POPPED (ONLY IPL2 ALLOWED)
0.0184
0.0185 ;***** MACROS *****
0.0186  Lsp2Wsp MACRO LAddr,WAddr
0.0187
0.0188 ;LOAD DATA INTO WORKING AREA
0.0189 ;
0.0190 ;NB!! NOTE :
0.0191 ; DATA TRANSFER TO PC AND SAMPLING HAPPENS CONCURRENTLY.
0.0192 ; BE THEREFORE CAREFUL AS TO WHICH REGISTERS ARE USED.
0.0193
0.0194      move    #LAddr,r2
0.0195      move    #WAddr,r4
0.0196      move    r2,r5
0.0197      move    r4,r1
0.0198
0.0199      do      #N,_xym
0.0200      move    x:(r2)+,x0 y:(r5)+,y0
0.0201      move    x0,x:(r1)+ y0,y:(r4)+
0.0202  _xym
0.0203      ENDM
0.0204
0.0205
0.0206 ;*****
0.0207  BitRev MACRO I_addr,T_addr,Num
0.0208
0.0209 ;MACRO TO BIT REVERSE FFT OUTPUT
0.0210 ;
0.0211 ;where :
0.0212 ;I_addr = input data pointer
0.0213 ;T_addr = temporary buffer to hold "unscrambled" data
0.0214 ;Num = Number of values to be RETURNED from
0.0215 ; temporary buffer
0.0216
0.0217      move    #0,m0
0.0218      move    m0,m4
0.0219      move    #T_addr,r1
0.0220      move    r1,r5
0.0221      move    #N_FFT/2,n0
0.0222      move    n0,n4
0.0223      move    #I_addr,r0
0.0224      move    r0,r4
0.0225
0.0226 ;BIT REVERSE DATA
0.0227      do      #N_FFT,_l1
0.0228      move    x:(r0)+n0,x0      y:(r4)+n4,y0
0.0229      move    x0,x:(r1)+      y0,y:(r5)+
0.0230  _l1
0.0231

```

```

0.0232 ;RESET LINEAR ADDRESS MODE
0.0233     move    #$FFFF,m0
0.0234     move    m0,m4
0.0235     move    #I_addr,r0
0.0236     move    r0,r4
0.0237     move    #T_addr,r1
0.0238     move    r1,r5
0.0239
0.0240 ;PUT DATA VALUES BACK
0.0241     do      #Num_ l2
0.0242     move    x:(r1)+,x0          y:(r5)+,y0
0.0243     move    x0,x:(r0)+ y0,y:(r4)+
0.0244     _l2
0.0245
0.0246     ENDM
0.0247
0.0248 ;*****
0.0249     sincos macro points,coef
0.0250
0.0251 ;MACRO TO ESTABLISH ARGUMENTS FOR FFT
0.0252 ;
0.0253 ;where :
0.0254 ;points = FFT number that will be done.
0.0255 ;coeff = base address of the arguments
0.0256 ;negative cosine value in X:MEM
0.0257 ;negative sine value in Y:MEM
0.0258
0.0259
0.0260     pi      set    3.141592654
0.0261     freq    set    2.0*pi/@cvf(points)
0.0262
0.0263     org     x:coef
0.0264     count   set    0
0.0265     dup     points/2
0.0265     dc     -@cos(@cvf(count)*freq)
0.0266     count   set    count+1
0.0267     endm
0.0268
0.0269     org     y:coef
0.0270     count   set    0
0.0271     dup     points/2
0.0272     dc     -@sin(@cvf(count)*freq)
0.0273     count   set    count+1
0.0274     endm
0.0275     org     p:
0.0276     endm
0.0277
0.0278 ;*****
0.0279     Carrier macro addr
0.0280
0.0281 ;MACRO TO ESTABLISH CARRIER WAVES FOR FREQUENCY
0.0282 ;DOMAIN PROCESSING.
0.0283 ;
0.0284 ;where :
0.0285 ;addr = base address of the arguments
0.0286 ;
0.0287 ; cos(wt) in X:MEM (f = 10khz (Fs\4))
0.0288 ; sin(wt) in Y:MEM

```

```

0.0289
0.0290   freqe   set    5e3
0.0291   pi     set    3.141592654
0.0292   omega  set    (2.0*pi*freqe/Fs)
0.0293
0.0294           org    x:addr
0.0295   count  set    0
0.0296           dup    N
0.0297           dc    0.9999999*@cos(@cvf(count)*omega)
0.0298   count  set    count+1
0.0299           endm
0.0300
0.0301           org    y:addr
0.0302   count  set    0
0.0303           dup    N
0.0304           dc    0.9999999*@sin(@cvf(count)*omega)
0.0305   count  set    count+1
0.0306           endm
0.0307           org    p:
0.0308           endm
0.0309
0.0310 ;*****
0.0311   HAMMING macro points,addr
0.0312
0.0313 ;MACRO TO DETERMINE WINDOW WEIGHTS
0.0314 ;
0.0315 ;w(n) = 0.54-0.46cos(2*pi*n/N)
0.0316 ;where :
0.0317 ;points = N (Sample number taken)
0.0318 ;addr = base address of the weights
0.0319
0.0320   pi     set    3.141592654
0.0321   freq  set    2.0*pi/@cvf(points)
0.0322
0.0323           org    x:addr
0.0324   count  set    0
0.0325           dup    points
0.0326           dc    0.9999999*(0.54-0.46*@cos(@cvf(count)*freq))
0.0327   count  set    count+1
0.0328           endm
0.0329
0.0330           org    p:
0.0331           endm
0.0332
0.0333 ;*****
0.0334   MulHAM MACRO Mem,Addr
0.0335
0.0336 ;MULTIPLY TIME WINDOW WITH HAMMING WINDOW
0.0337 ;A window length of N is assumed.
0.0338 ;
0.0339 ;where :
0.0340 ;Mem = X or Y memory
0.0341 ;Addr = address pointer
0.0342 ;Fraction * Integer
0.0343
0.0344           move   #Addr,r0
0.0345           move   #Hama,r1
0.0346

```

```

0.0347      do      #N,_win
0.0348      move    \Mem:(r0),y0
0.0349      move    x:(r1)+,y1
0.0350      mpyr   y0,y1,b
0.0351      move    b,\Mem:(r0)+
0.0352      _win
0.0353      ENDM
0.0354
0.0355 ;*****
0.0356      HAM_XY MACRO Addr
0.0357
0.0358 ;MULTIPLY TIME WINDOW WITH HAMMING WINDOW
0.0359 ;BOTH X AND Y MEM ARE MULTIPLIED.
0.0360
0.0361 ;A window length of N is assumed.
0.0362 ;where :
0.0363 ;Addr = address pointer
0.0364 ;Fraction * Integer
0.0365
0.0366      move    #Addr,r0
0.0367      move    #HamA,r1
0.0368      move    r0,r4
0.0369      move    r0,r2
0.0370      move    r0,r5
0.0371
0.0372      move    x:(r1)+,x0
0.0373      move    x:(r0)+,x1 y:(r4)+,y1
0.0374      do      #N,_win
0.0375      mpyr   x0,x1,a          x:(r0)+,x1
0.0376      mpyr   x0,y1,b          x:(r1)+,x0 y:(r4)+,y1
0.0377      move    a,x:(r2)+ b,y:(r5)+
0.0378      _win
0.0379      ENDM
0.0380
0.0381 ;*****
0.0382      MulCar MACRO
0.0383
0.0384 ;MULTIPLY TIME WINDOW WITH CARRIER WAVES
0.0385 ;A WINDOW LENGTH OF N IS ASSUMED.
0.0386 ;
0.0387 ;DIRECT CHANNEL IN X MEM * SINE WAVE
0.0388 ;QUADRATURE CHANNEL IN Y MEM * COS WAVE
0.0389 ;
0.0390      move    #WkAddr,r0
0.0391      move    #2,n0
0.0392      move    #Ca_Add,r4
0.0393      move    r0,r5
0.0394
0.0395      move    x:(r0),x0 y:(r4),y0
0.0396      move    x:(r4)+,x1 y:(r0)+,y1
0.0397      do      #N,_cwin
0.0398      mpyr   x0,y0,a          x:(r0),x0 y:(r4),y0
0.0399      mpyr   y1,x1,b          x:(r4)+,x1 y:(r0)+,y1
0.0400      move    a,x:(r0)+n0    b,y:(r5)+
0.0401      _cwin
0.0402      ENDM
0.0403
0.0404 ;*****

```

```

0.0405  Add_DQ MACRO
0.0406
0.0407 ;MULTIPLY TIME WINDOW WITH CARRIER WAVES
0.0408 ;A WINDOW LENGTH OF N IS ASSUMED.
0.0409 ;
0.0410 ;DIRECT CHANNEL IN X MEM * SINE WAVE
0.0411 ;QUADRATURE CHANNEL IN Y MEM * COS WAVE
0.0412 ;Y MEM ZEROED AFTER USE.
0.0413
0.0414      move    #WkAddr,r0
0.0415      move    r0,r4
0.0415      clr     b                #WkAddr,r1
0.0416
0.0417      move    x:(r0),x0
0.0418      move    y:(r4)+,a
0.0419      do     #N,_awin
0.0420      add     x0,a                x:(r4),x0  b,y:(r1)+
0.0421      move    a,x:(r0)+  y:(r4)+,a
0.0422
0.0423      _awin
0.0424      ENDM
0.0425
0.0426 ;*****
0.0427  Mirr_ZP MACRO Raddr
0.0428
0.0429 ;MIRROR AND ZEROPAD EXTRAPOLATED Rxx
0.0430 ;IT IS ASSUMED THAT N_FFT/2 POINTS EXIST.
0.0431 ;
0.0432 ;where :
0.0433 ;Raddr = Rxx address
0.0434
0.0435      move    #Raddr,r0
0.0436      clr     a                #Raddr-1,r4
0.0437      move    a,y:(r0)+
0.0438
0.0439      do     #(N_FFT/2-1),_mirror
0.0440      move    x:(r0),b                a,y:(r4)
0.0441      move    b,x:(r4)-                a,y:(r0)+
0.0442      _mirror
0.0443      move    b,x:(r4)                ;COMPENSATE FOR 2N-1 POINT Rxx
0.0444      move    a,y:(r4)                ;THROUGH DUPLICATING LAST POINT
0.0445
0.0446      ENDM
0.0447
0.0448 ;*****
0.0449  ForRev  MACRO Daddr,Qaddr,P_N
0.0450
0.0451 ;MACRO TO CALCULATE FORWARD AND REVERSE FLOW COMPONENTS
0.0452 ;
0.0453 ;where :
0.0454 ;Daddr = Direct channel address
0.0455 ;Qaddr = Quadrature channel address
0.0456 ;P_N = 1 for doing positive AND negative frequencies
0.0457 ;      0 for doing only positive.
0.0458 ;
0.0459 ;THE INPUT TO THE MACRO IS THE FREQUENCY SPECTRUM OF BOTH
0.0460 ;CHANNELS IN TERMS OF THEIR REAL AND IMAGINARY COMPONENTS
0.0461 ;AS CALCULATED PREVIOUSLY.

```

```

0.0462 ;
0.0463 ;PHASE DOMAIN PROCESSING IS USED AS FOLLOWS :
0.0464 ;FORWARD FLOW = DIRECT CHANNEL + QUADRATURE CHANNEL PHASE
0.0465 ;           SHIFTED BY PI/2
0.0466 ;REVERSE FLOW = QUADRATURE CHANNEL + DIRECT CHANNEL PHASE
0.0467 ;           SHIFTED BY PI/2
0.0468 ;
0.0469 ;USE IS MADE OF THE FOLLOWING MATHEMATICAL ENTITY :
0.0470 ;POSITIVE FREQUENCIES :
0.0471 ; (Re + jIm)j = (-Im + jRe)
0.0472 ;           = Phase advance by PI/2
0.0473 ;
0.0474 ;NEGATIVE FREQUENCIES :
0.0475 ; (Re + jIm)(-j) = (-Im + jRe)
0.0476 ;           = Phase delay by PI/2
0.0477 ;
0.0478 ;
0.0479         move    #Daddr,r0
0.0480         move    #Qaddr,r4
0.0481
0.0482         do      #N_FFT/2,_posloop
0.0483
0.0484 ;FORWARD FLOW
0.0485         move    x:(r0),a    y:(r4),y0
0.0486         sub     y0,a        x:(r4),b    y:(r0),y0
0.0487         add     y0,b        a,y0
0.0488         move    b,x0
0.0489
0.0490 ;REVERSE FLOW
0.0491         move    x:(r4),a    y:(r0),y1
0.0492         sub     y1,a        x:(r0),b    y:(r4),y1
0.0493         add     y1,b        x0,x:(r0)  a,y:(r4)
0.0494         move    b,x:(r4)+  y0,y:(r0)+
0.0495         _posloop
0.0496
0.0497         clr    b    #>P_N,x0
0.0498         cmp    x0,b
0.0499         jeq    _negloop
0.0500
0.0501         do      #N_FFT/2,_negloop
0.0502
0.0503 ;FORWARD FLOW
0.0504         move    x:(r0),a    y:(r4),y0
0.0505         add     y0,a        x:(r4),x0  y:(r0),b
0.0506         sub     x0,b        a,y0
0.0507         move    b,x0
0.0508
0.0509 ;REVERSE FLOW
0.0510         move    x:(r4),a    y:(r0),y1
0.0511         add     y1,a        x:(r0),x1  y:(r4),b
0.0512         sub     x1,b        x0,x:(r0)  a,y:(r4)
0.0513         move    y0,y:(r0)+  b,x:(r4)+
0.0514         _negloop
0.0515
0.0516         ENDM
0.0517
0.0518 ;*****
0.0519     AveSub MACRO     Addr

```

```

0.0520
0.0521 ;FIR FILTER. THIS LEAVES Rxx(O)=VARIANCE OF SEQUENCE.
0.0522 ;DETERMINE THE AVERAGE VALUE PER WINDOW AND SUBTRACT
0.0523 ;THAT VALUE FROM EACH SAMPLE FOR BOTH X AND Y MEM.
0.0524 ;
0.0525 ;where :
0.0526 ;Addr = starting address
0.0527
0.0528   SRT      set      @CVI(@LOG(N)/@LOG(2))
0.0529
0.0530           move     #Addr,r4
0.0531           clr      a              #Addr,r0
0.0532           clr      b
0.0533           move     x:(r0)+,x0 y:(r4)+,y0
0.0534
0.0535           do       #N,_ave
0.0536           add     x0,a              x:(r0)+,x0
0.0536           add     y0,b              y:(r4)+,y0
0.0537   _ave
0.0538           rep     #SRT
0.0539           asr     a
0.0540           rep     #SRT
0.0541           asr     b
0.0542
0.0543           move     #Addr,r0
0.0544           move     #Addr,r4
0.0545           move     a,x0
0.0546           move     b,y0
0.0547
0.0548           do       #N,_sub
0.0549           move     x:(r0),a      y:(r4),b
0.0550           sub     x0,a
0.0551           sub     y0,b
0.0552           move     a,x:(r0)+    b,y:(r4)+
0.0553   _sub
0.0554
0.0555           ENDM
0.0556
0.0557 ;*****
0.0558   ZeroPad MACRO addr,numz
0.0559
0.0560 ;MACRO TO INTERPOLATE FFT FREQUENCIES
0.0561 ;
0.0562 ;where :
0.0563 ;addr = from where do we start zeropadding
0.0564 ;numz = how many zeros do we add
0.0565 ;
0.0566 ;NB! ZERO "do" LOOPS NOT ALLOWED.
0.0567
0.0568           clr      b              #>numz,x0
0.0569           cmp     x0,b              #addr,r4
0.0570           jeq     _padloop
0.0571           clr      a              #addr,r0
0.0572           do       #numz,_padloop
0.0573           move     a,x:(r0)+      b,y:(r4)+
0.0574   _padloop
0.0575
0.0576           ENDM

```

```

0.0577
0.0578 ;*****
0.0579   ExtrRxx MACRO C_Addr,R_Addr
0.0580
0.0581 ;MACRO TO EXTRAPOLATE AUTOCORR TO N_FFT/2 POINTS.
0.0582 ;
0.0583 ;   p
0.0584 ;Rxx(n)=-aa(k)*Rxx(n-k) for |n|>p
0.0585 ;   k=1
0.0586 ;
0.0587 ;p = order of AR filter.
0.0588 ;a = LPC coeff of AR filter.
0.0589 ;LPC IN Y MEM
0.0590 ;Rxx IN X MEM (SEE AUTOCOR)
0.0591 ;
0.0592 ;where :
0.0593 ;C_Addr = LPC coeff address.
0.0594 ;R_Addr = Rxx address.
0.0595
0.0596         move    #(R_Addr+N_Aut-1),r4
0.0597         move    #(R_Addr+N_Aut),r1
0.0598         move    #(C_Addr+1),r0
0.0599         move    #(N_AR+2),n4
0.0600
0.0601         do      #N_Pred,_Outl
0.0602         clr     a
0.0603         move    x:(r4)-,x0 y:(r0)+,y0
0.0604         do      #N_AR,_Inl
0.0605         mac    -x0,y0,a  x:(r4)-,x0          y:(r0)+,y0
0.0606         _Inl
0.0607         move    a,x:(r1)+          y:(r4)+n4,y1      ;Y MEM DUMMY READ
0.0608         move    #(C_Addr+1),r0
0.0609         _Outl
0.0610
0.0611         ENDM
0.0612
0.0613 ;*****
0.0614   sqrt3 MACRO Prec
0.0615
0.0616 ; INTEGER SQRT
0.0617 ; Full (Prec) bit precision square root routine using
0.0618 ; a successive approximation technique.
0.0619 ;
0.0620 ; y = double precision (48 bit) positive input number
0.0621 ; b = 24 bit output root
0.0622 ;
0.0623 ; a = temporary storage
0.0624 ; x0 = guess
0.0625 ; x1 = bit being tested
0.0626 ; y1:y0 = input number
0.0627
0.0628         clr     b          #<$40,x0          ;init root and guess
0.0629         move    x0,x1          ;init bit to test
0.0630
0.0631         do      #Prec,_endl          ;START OF LOOP
0.0632         mpy    -x0,x0,a          ;square and negate the guess
0.0633         asr     a
0.0634         add     y,a          ; compare to double precision input

```

```

0.0635      tge      x0,b          ; update root if input >= guess
0.0636      tfr      x1,a          ; get bit to test
0.0637      asr      a              ; shift to next bit to test
0.0638      add      b,a          a,x1 ; form new guess
0.0639      move     a,x0          ; save new guess
0.0640      _endl
0.0641      ENDM
0.0642
0.0643 ;*****
0.0644      Magn  MACRO Addr,Nump
0.0645
0.0646 ;MACRO TO TAKE ABS OF FFT
0.0647 ;
0.0648 ;ABS = SQRT(SQR(REAL)+SQR(IMAG))
0.0649 ;
0.0650 ;where :
0.0651 ;Addr = pointer to buffer space
0.0652 ;Nump = number of points to do
0.0653
0.0654      move     #Addr,r0
0.0655      move     r0,r4
0.0656      move     #Addr,r1
0.0657
0.0658      do      #Nump,_abs
0.0659      move     x:(r0)+,x0      y:(r4)+,y0
0.0660      mpy     x0,x0,a
0.0661      mac     y0,y0,a
0.0662
0.0662      asr      a              ;INT * (Result in a0)
0.0663      move     a0,y0
0.0664      move     a1,y1
0.0665      sqrt3   13              ;TAKE INT SQRT
0.0666
0.0667      move     b,x:(r1)+
0.0668      _abs
0.0669      ENDM
0.0670
0.0671 ;*****
0.0672      ChrShft MACRO IAddr,Num
0.0673
0.0674 ;PREPARE FOR CHAR (8 BIT) TRANSFER.
0.0675
0.0676      move     #IAddr,r0
0.0677
0.0678      do      #Num,_chr
0.0679      move     x:(r0),a
0.0680      rep     #16
0.0681      asr      a
0.0682      move     a,x:(r0)+
0.0683      _chr
0.0684      ENDM
0.0685
0.0686 ;*****
0.0687      PerG   MACRO Addr,Nump
0.0688
0.0689 ;MACRO TO TAKE ABS^2 OF FFT
0.0690 ;
0.0691 ;ABS = SQR(REAL)+SQR(IMAG)

```

```

0.0692 ;
0.0693 ;where :
0.0694 ;Addr = pointer to buffer space
0.0695 ;Nump = number of points to do
0.0696
0.0697     move    #Addr,r0
0.0698     move    r0,r4
0.0699     move    r0,r1
0.0700
0.0701     move    x:(r0)+,x0      y:(r4)+,y0
0.0702     do      #Nump,_per
0.0703     mpy    x0,x0,a          x:(r0)+,x0
0.0704     mac   y0,y0,a          y:(r4)+,y0
0.0705
0.0706     move    a,x:(r1)+
0.0707     _per
0.0708     ENDM
0.0709
0.0710 ;*****
0.0711     AutoCor MACRO NumC,Ms1,InA,Ms2,OutA
0.0712
0.0713 ;MACRO TO ESTIMATE BIASED AUTOCORRELATION FUNCTION
0.0714 ;
0.0715 ;           n=N-m-1
0.0716 ;C(m) =    1/N ∑ x(n)*x(n+m)  m=0..NumC
0.0717 ;           n=0
0.0718 ;
0.0719 ;NumC = number of coeff to be calculated
0.0720 ;MsX = memory space being used
0.0721 ;InA = vector address of input array
0.0722 ;OutA = vector address of autocorr coeff
0.0723 ;
0.0724 ;INSERT THE FOLLOWING IF INT MULTIPLICATION
0.0725 ;IS DESIRED.
0.0726 ;     rep    #(SHTCNT+1)          ;DIVIDE BY N + INT *
0.0727 ;     asr    a
0.0728 ;     move   a0,\Ms2:(r1)+        ;TRANSFER OUT
0.0729
0.0730     SHTCNT SET    @CVI(@LOG(N)/@LOG(2))
0.0731
0.0732     move    #OutA,r1
0.0733     move    #N,r2
0.0734     move    #0,r4                ;r4=m
0.0735
0.0736     do      #NumC,_out1
0.0737     clr    a                      #InA,r0
0.0738     move   r4,n0
0.0739
0.0740     do      r2,_in1                ;r2=N-m-1
0.0741     move   \Ms1:(r0),x0            ;x(n)
0.0742     move   \Ms1:(r0+n0),y0        ;x(n+m)
0.0743     macr  x0,y0,a (r0)+
0.0744     _in1
0.0745     move   a,\Ms2:(r1)+
0.0746     move   x:(r2),x1              y:(r4)+,y1      ;dummy read
0.0747     _out1
0.0748     ENDM
0.0749

```

```

0.0750 ;*****
0.0751  InterP MACRO  INum,addr
0.0752
0.0753 ;TO INTERPOLATE BETWEEN FREQUENCIES. THE OPERATION IS
0.0754 ;EQUIVALENT TO ZEROPADDING WHEN DOING AN FFT.
0.0755 ;
0.0756 ;TWO SUCCESSIVE NUMBERS ARE COMPARED AND A LINEAR INTERPOLATION IS
0.0757 ;MADE. THE FINAL SEGMENT DECAYS TO ZERO.
0.0758 ;where :
0.0759 ;INum = initial number count
0.0760 ;FNum = final number count
0.0761
0.0762  IncN    set      8
0.0763  IncD    set      @CVI(@LOG(IncN)/@LOG(2))
0.0764
0.0765          clr      a                #WkAddr,r0
0.0766          move    #addr,r1
0.0767          move    a,x:(WkAddr+INum)
0.0768
0.0769          move    x:(r0)+,x0
0.0770          move    x:(r0),a
0.0771
0.0772          do      #INum,_betw
0.0773          sub     x0,a                x0,x:(r1)+
0.0774          rep     #IncD
0.0775          asr     a                x0,b
0.0776          move    x:(r0)+,x0 a,y1
0.0777
0.0778          add     y1,b
0.0778          do      #IncN-1,_ins
0.0779          add     y1,b                b,x:(r1)+
0.0780  _ins
0.0781          move    x:(r0),a
0.0782  _betw
0.0783
0.0784 ;          do      #IncN,_last
0.0785 ;          add     y1,b                b,x:(r1)+
0.0786 ;_last
0.0787          ENDM
0.0788
0.0789 ;*****
0.0790  BartL MACRO    Num
0.0791
0.0792 ;SMOOTHING OF SPECTRUM THROUGH AVERAGING
0.0793 ;where Num = number of points for each spectrum
0.0794
0.0795          move    #WkAddr,r0
0.0796          move    #WkAddr+M,r4
0.0797          move    #M,n2
0.0798
0.0799          do      #Num,_add
0.0800          move    r4,r2
0.0801          move    x:(r0),a
0.0802          move    x:(r2)+n2,x0
0.0803
0.0804          do      #BartN-1,_smooth
0.0805          add     x0,a                x:(r2)+n2,x0
0.0806  _smooth

```

```

0.0807      move    a,x:(r0)+      y:(r4)+,y0 ;dummy read
0.0808      _add
0.0809
0.0810      ENDM
0.0811
0.0812 ;*****
0.0813      Spect    MACRO
0.0814
0.0815 ;CALCULATE SPECTRA OF THE TIME SIGNAL. BARTLETT SMOOTHING
0.0816 ;OF THE PERIODOGRAM FOLLOWS. NOTE THAT THE VARIANCE OF THE FINAL
0.0817 ;SPECTRUM APPROACHES ZERO AS THE SAMPLE NUMBER INCREASES
0.0818 ;
0.0819 ;USE THE FOLLOWING FOR BARTLETT SPECTRAL ESTIMATION
0.0820 ;BartN      set      4          ;BARTLETT BLOCKS
0.0821 ;M          set      128        ;SAMPLES PER BLOCK
0.0822 ;
0.0823 ;          Spect          ;CALCULATE SPECTRA
0.0824 ;          BartL      N_FFT/4      ;DO BARTLETT SMOOTHING
0.0825 ;          InterP     N_FFT/4,2*N_FFT ;INTERPOLATE FREQ POINTS (32->256)
0.0826 ;
0.0827 ;          ffr2c      N_FFT,WkAddr,A_Addr ;DO COMPLEX FFT
0.0828 ;          BitRev     WkAddr,Wsp,N_FFT/4 ;BITREV OUTPUT OF FFT
0.0829 ;          PerG       WkAddr,N_FFT/4    ;DETERMINE MAGNITUDE
0.0830 ;
0.0831 ;          ffr2c      N_FFT,WkAddr+M,A_Addr
0.0832 ;          BitRev     WkAddr+M,Wsp,N_FFT/4
0.0833 ;          PerG       WkAddr+M,N_FFT/4
0.0834 ;
0.0835 ;          ffr2c      N_FFT,WkAddr+2*M,A_Addr
0.0836 ;          BitRev     WkAddr+2*M,Wsp,N_FFT/4
0.0837 ;          PerG       WkAddr+2*M,N_FFT/4
0.0838 ;
0.0839 ;          ffr2c      N_FFT,WkAddr+3*M,A_Addr
0.0840 ;          BitRev     WkAddr+3*M,Wsp,N_FFT/4
0.0841 ;          PerG       WkAddr+3*M,N_FFT/4
0.0842 ;
0.0843      ENDM
0.0844
0.0845 ;*****
0.0846      Leroux    MACRO nk,Raddr,Kaddr
0.0847
0.0848 ;USE THE FOLLOWING FOR AR SPECTRAL ESTIMATION
0.0849 ; N_AR      set      15          ;NEED ONE MORE Rxx THAN ORDER REQUIRED
0.0850 ; N_Aut      set      64          ;IN DURBIN ALGORITHM.
0.0851 ; N_Pred      set      (N_FFT/4-N_Aut) ;Rxx POINTS PREDICTED.
0.0852 ; N_Zero      set      N_FFT/4      ;NUMBER OF POINTS TO ZEROPAD
0.0853 ;
0.0854 ;          AutoCor    N_Aut,x,WkAddr,x,WkAddr+N
0.0855 ;          Leroux     N_AR,WkAddr+N,WkAddr ;REFLECTION COEFF
0.0856 ;          Durbin     N_AR,WkAddr+N,WkAddr ;AR COEFF
0.0857 ;
0.0858 ;          ExtrRxx    WkAddr,WkAddr+N      ;EXTRAPOLATE Rxx
0.0859 ;          ZeroPad    WkAddr+N+N_FFT/4,N_Zero ;ZEROPAD Rxx
0.0860 ;          Mirr_ZP    (WkAddr+N)          ;MIRROR AND ZERO Y MEM
0.0861 ;          ffr2c      N_FFT,WkAddr,A_Addr ;DO COMPLEX FFT
0.0862 ;          BitRev     WkAddr,Wsp,N_FFT/4 ;BITREV OUTPUT OF FFT
0.0863 ;          Magn       WkAddr,N_FFT/4
0.0864 ;

```

```

0.0865 ; Leroux-Gueguen Solution for PARCOR (LPC) Coefficients
0.0866 ;
0.0867 ;   Implements the Leroux-Gueguen algorithm.
0.0868 ;
0.0869 ;X MEMORY VALUES
0.0870 fk      set      Kaddr      ;LPC reflection coefficients
0.0871 r       set      Raddr
0.0872
0.0873          org      y:Kaddr
0.0874 bim1    ds      nk+1
0.0875 bi      ds      nk
0.0876 ;
0.0877          org      p:
0.0878
0.0879 ; Start of LPC analysis
0.0880 ; The following registers correspond to the FORTRAN variables:
0.0881 ;
0.0882 ; r0 - r[i] = irptr   r4 - bim1[j] = jbm1
0.0883 ; r1 - fk[i] = ifk   r5 - bim1[i] = ibim1
0.0884 ; r2 - fk[j] = jfk   r6 - bi[i] = ibi
0.0885 ; r3 - bi[j] = jbi   r7 - loop counter = im1
0.0886 ;
0.0887 ; Approximate analysis times:
0.0888 ;
0.0889 ; 8th order: 46.2 uS
0.0890 ; 10th order: 61.6 uS
0.0891 ; 16th order: 117.2 uS
0.0892 ;
0.0893 ; Initialization
0.0894 ;
0.0895     move   #r,r0           ;r(0) points to r[0]
0.0896     move   #fk,r1        ;r(1) points to fk[1]
0.0897     move   #bim1,r5      ;r(5) points to bim1[1]
0.0898     move   #bi+1,r6      ;r(6) points to bi[2]
0.0899 ;
0.0899 ; Begin Leroux - Gueguen Algorithm
0.0900 ;
0.0901     move   x:(r0)+,x0     ;get r[0]
0.0902     move   x:(r0)+,a     ;get r[1], r(0) points to r[2]
0.0903     abs    a             a,b ;get abs(r[1]), copy r[1] to b
0.0904     eor    x0,b         b,y0 ;N = sign bit, copy r[1] to y0
0.0905     and    #$fe,ccr      ;clear quotient sign bit
0.0906     rep    #24           ;set up for 24-bit quotient
0.0907     div   x0,a           ;get abs(r[1])/r[0]
0.0908     jmi   L1            ;check sign bit N
0.0909     neg   a             ;negate quotient if needed
0.0910 L1  move   a0,a         ;move -r[1]/r[0] to a
0.0911     move   a,x:(r1)+   a,y1 ;copy -r[1]/r[0] to fk[1] and y1
0.0912     tfr   x0,b         y0,y:(r5)+ ;bim1[1] = r[1], copy r[0] to b
0.0913     macr  y1,y0,b     #1,r7 ;r[0] + fk[1] * r[1]; loop count = 1
0.0914     move   b,y:(r5)    ;bim1[2] = r[0] + (fk[1] * r[1])
0.0915 ;
0.0916 ; outer do loop
0.0917 ;
0.0918     do    #nk-1,L4
0.0919     move   #fk,r2        ;r(2) points to fk[1]
0.0920     move   #bi,r3        ;r(3) points to bi[1]
0.0921     move   #bim1,r4     ;r(4) points to bim1[1]

```

```

0.0922      move      x:(r0)+,b  y:(r5),y1      ;yi = r[i], get bim1[i]
0.0923      move      b,x1      b,y:(r3)      ;copy yi to x1, bi[1] = yi
0.0924 ;
0.0925 ; inner do loop
0.0926 ;
0.0927      do        r7,L2                        ;begin inner do loop
0.0928      move      x:(r2)+,x0      y:(r4),a      ;get fk[j] and bim1[j]
0.0929      macr      x1,x0,a      a,x1 y:(r3)+,y0      ;bim1[j]+(fk[j]*yi)
0.0930      macr      x1,x0,b      a,y:(r3)      ;save bi[j+1], yi=yi+(fk[j]*bim1[j])
0.0931      move      b,x1      y0,y:(r4)+      ;copy yi to x1, save bim1[j]
0.0932 ;
0.0933 ; end of inner do loop (note: b = x1 = yi)
0.0934 ;
0.0935      L2 abs      b      b,a      ;get abs(yi), copy yi to a
0.0936      eor      y1,a      (r7)+      ;N = sign bit, inc. outer loop counter
0.0937      and      #$fe,ccr      ;clear quotient sign bit
0.0938      rep      #24      ;set up for 24-bit quotient
0.0939      div      y1,b      ;get abs(yi)/bim1[i]
0.0940      jmi      L3      ;check sign bit N
0.0941      neg      b      ;negate quotient if needed
0.0942      L3 move      b0,x0      ;move -yi/bim1[i] to x0
0.0943      tfr      y1,b      x0,x:(r1)+ y:(r6)+,y0      ;get bim1[i],save fk[i],get bi[i]
0.0944      macr      x1,x0,b      y0,y:(r5)+      ;bim1[i]+(fk[i]*yi), bim1[i] = bi[i]
0.0945      move      b,y:(r5)      ;bim1[i+1] = bim1[i]*(fk[i]*yi)
0.0946 ;
0.0947 ; end of outer do loop
0.0948 ; end of analysis
0.0949      L4
0.0950
0.0951      ENDM
0.0952
0.0953 ;*****
0.0954      Durbin MACRO nk,Raddr,Kaddr
0.0955
0.0956 ;MACRO TO ESTIMATE AR COEFFICIENTS GIVEN REFLECTION COEFF
0.0957 ;
0.0958 ;nk      = number of AR reflection coeff
0.0959 ;Raddr   = autocorrelation address
0.0960 ;Kaddr   = AR coefficient address
0.0961
0.0962      acoef      set      Kaddr      ;LPC filter coeff (nk+1 values)
0.0963      refl      set      Kaddr      ;Reflection coeff (nk values)
0.0964      anew      set      (Kaddr+nk+1) ;updated LPC filter coeff (nk values)
0.0965      error      set      (anew+nk);error
0.0966      alpha     set      (error+1) ;alpha
0.0967
0.0968      move      #refl,r1      ;r1 points to k[0]
0.0969      move      #Raddr,r0      ;r0 points to r[0]
0.0970      move      #anew,r3      ;r3 points to anew[0]
0.0971      move      #acoef,r4      ;r4 points to a[0]
0.0972      move      #error,r6      ;r6 points to error
0.0973
0.0974      move      #$7FFFFFFF,a
0.0975      move      a,y:(r4)+      ;define a[0]=1.0
0.0976      move      a,y:(r3)      ;anew[0]=1.0
0.0977
0.0978      move      #2,r2      ;r2=i (SEE BELOW)
0.0979      move      x:(r1)+,a      ;k[0]

```

```

0.0980      move    x:(r0)+,x0 a,y:(r4)+ ;r[0] in x0, a[1]=k[0]
0.0981
0.0982      clr     b          a,y0          ;k[0] in y0
0.0983      mpy    y0,y0,a      x0,b          ;r[0] in b
0.0984      move    a,y0          ;k[0]^2 in y0
0.0985      mac    -y0,x0,b      #2,n2          ;alpha in b
0.0986      move    b,x1          ;alpha in x1
0.0987
0.0988      move    #refl,n1          ;n1 points to k[0]
0.0989      move    #Raddr,n0        ;n0 points to r[0]
0.0990      move    #acoef,n5
0.0991
0.0992      do     #nk-1,_L1          ;OUTER LOOP
0.0993      move    r2,r0
0.0994      clr     a          #acoef,r4 ;r4 points to a[0]
0.0995      move    (r0)+n0
0.0996
0.0997      do     r2,_L2          ;DO i TIMES
0.0998      move    x:(r0)-,x0      y:(r4)+,y0
0.0999      mac    x0,y0,a          ;error in a
0.1000  _L2
0.1001      move    (r2)-          ;r2=i-1
0.1002      move    r2,r1
0.1003      move    a,y:(r6)+      ;error out
0.1004
0.1005      move    x:(r1+n1),y0      ;k[i-1]
0.1006      mpy    y0,y0,a      x1,b          ;y0=k[i-1]
0.1007      move    a,y1
0.1008      mac    -y1,x1,b      y0,y:(r4)      ;a[i-1]=k[i-1]
0.1009      move    b,x1          b,y:(r6)-      ;alpha in x1
0.1010
0.1011      move    #(anew+1),r3
0.1012      move    #(acoef+1),r4      ;r4=a[1]
0.1013      move    r2,r5          ;r5=i-1
0.1014
0.1015      do     r2,_L3          ;DO i-1 TIMES
0.1016      move    y:(r4)+,b
0.1017      move    y:(r5+n5),x0      ;a[i-j+1]
0.1018      mac    x0,y0,b (r5)-
0.1019      move    b,y:(r3)+
0.1020
0.1020  _L3
0.1021      do     r2,_L4          ;DO i-1 TIMES
0.1022      move    y:- (r3),b
0.1023      move    b,y:- (r4)
0.1024  _L4
0.1025      move    (r2)+n2          ;INC i
0.1026  _L1
0.1027
0.1028      ENDM
0.1029
0.1030 ;*****
0.1031 ffr2c MACRO points,data,coef
0.1032
0.1033 ;NB! IF COEF IS TO LARGE A OVERFLOW OCCURS WHEN THE OFFSET
0.1034 ;n6 IS ADDED !!!
0.1035 ;
0.1036 ;MACRO TO AID IN CALCULATION OF FREQUENCY CONTENT

```

```

0.1037 ;OF A SIGNAL. THE SAME DATA BUFFER IS USED TO HOLD
0.1038 ;INPUT AND OUTPUT DATA
0.1039 ;
0.1040 ; Radix 2 Decimation in Time In-Place Fast Fourier Transform Routine
0.1041 ; Complex input and output data
0.1042 ; Real data in X memory
0.1043 ; Imaginary data in Y memory
0.1044 ; Normally ordered input data
0.1045 ; Bit reversed output data
0.1046 ; Coefficient lookup table
0.1047 ; -Cosine values in X memory
0.1048 ; -Sine values in Y memory
0.1049 ;
0.1050 ; Macro Call - ffr2c points,data,coef
0.1051 ; points number of points (16-32768, power of 2)
0.1052 ; data start of data buffer
0.1053 ; coef start of sine/cosine table
0.1054 ;
0.1055 ; Alters Data ALU Registers
0.1056 ; x1 x0 y1 y0
0.1057 ; a2 a1 a0 a
0.1058 ; b2 b1 b0 b
0.1059 ;
0.1060 ; Alters Address Registers
0.1061 ; r0 n0 m0
0.1062 ; r1 n1 m1
0.1063 ; n2
0.1064 ; r4 n4 m4
0.1065 ; r5 n5 m5
0.1066 ; r6 n6 m6
0.1067 ;
0.1068 ; Alters Program Control Registers
0.1069 ; pc sr
0.1070 ;
0.1071 ; Uses 6 locations on System Stack
0.1072 ;
0.1073 move #data,r0 ;initialize input pointer
0.1074 move #points/4,n0 ;initialize butterflies per group
0.1075 move n0,n4 ;initialize pointer offsets
0.1076 move n0,n6 ;initialize coefficient offset
0.1077 move #points-1,m0 ;initialize address modifiers
0.1078 move m0,m1 ;for modulo(points) addressing
0.1079 move m0,m4
0.1080 move m0,m5
0.1081 ;
0.1082 ; Do first and second Radix 2 FFT passes
0.1083 move x:(r0)+n0,x0
0.1084 tfr x0,a x:(r0)+n0,y1
0.1085
0.1086 do n0,_twopass
0.1087 tfr y1,b x:(r0)+n0,y0
0.1088 add y0,a x:(r0),x1 ;ar+cr
0.1089 add x1,b r0,r4 ;br+dr
0.1090 add a,b (r0)+n0 ;ar'=(ar+cr)+(br+dr)
0.1091 subl b,a b,x:(r0)+n0 ;br'=(ar+cr)-(br+dr)
0.1092 tfr x0,a a,x0 y:(r0),b
0.1093 sub y0,a y:(r4)+n4,y0 ;ar-cr
0.1094 sub y0,b x0,x:(r0) ;bi-di

```

```

0.1095      add      a,b      y:(r0)+n0,x0      ;cr'=(ar-cr)+(bi-di)
0.1096      subl     b,a      b,x:(r0)                ;dr'=(ar-cr)-(bi-di)
0.1097      tfr      x0,a      a,x0                    y:(r4),b
0.1098      add      y0,a      y:(r0)+n0,y0            ;bi+di
0.1099      add      y0,b      x0,x:(r0)+n0            ;ai+ci
0.1100      add      b,a      y:(r0)+,x0              ;ai'=(ai+ci)+(bi+di)
0.1101      subl     a,b      a,y:(r4)+n4              ;bi'=(ai+ci)-(bi+di)
0.1102      tfr      x0,a      b,y:(r4)+n4
0.1103      sub      y0,a      x1,b                    ;ai-ci
0.1104      sub      y1,b      x:(r0)+n0,x0            ;dr-br
0.1105      add      a,b      x:(r0)+n0,y1            ;ci'=(ai-ci)+(dr-br)
0.1106      subl     b,a      b,y:(r4)+n4              ;di'=(ai-ci)-(dr-br)
0.1107      tfr      x0,a      a,y:(r4)+
0.1108      _twopass
0.1109 ;
0.1110 ; Do all FFT passes but first, second and last pass
0.1111      move     #points/8,n0                        ;initialize butterflies per group
0.1112      move     #4,n2                                ;initialize groups per pass
0.1113      move     #0,m6                                ;initialize coefficient address modifier
0.1114                                             ;for reverse carry (bit reversed) addressing
0.1115
0.1116      do      #@cvi(@log(points)/@log(2)-2.5),_end_pass
0.1117      move     #data,r0                            ;initialize A input pointer
0.1118      move     r0,r4                                ;initialize A output pointer
0.1119      lua      (r0)+n0,r1                          ;initialize B input pointer
0.1120      move     #coef,r6                            ;initialize C input pointer
0.1121      lua      (r1)-,r5                            ;initialize B output pointer
0.1122      move     n0,n1                              ;initialize pointer offsets
0.1123      move     n0,n4
0.1124      move     n0,n5
0.1125
0.1126      do      n2,_end_grp
0.1127      move     x:(r1),x1      y:(r6),y0          ;lookup -sine value
0.1128      move     x:(r5),a      y:(r0),b
0.1129      move     x:(r6)+n6,x0                        ;lookup -cosine value
0.1130
0.1131      do      n0,_end_bfy
0.1132                                             ;Radix 2 DIT butterfly kernel with constant
0.1133      mac      x1,y0,b      y:(r1)+,y1            ;twiddle factor
0.1134      macr     -x0,y1,b      a,x:(r5)+          y:(r0),a
0.1135      subl     b,a      x:(r0),b      b,y:(r4)
0.1136      mac      -x1,x0,b      x:(r0)+,a      a,y:(r5)
0.1137      macr     -y1,y0,b      x:(r1),x1
0.1138      subl     b,a      b,x:(r4)+          y:(r0),b
0.1139      _end_bfy
0.1140      move     a,x:(r5)+n5      y:(r1)+n1,y1      ;dummy load of x1 and y1
0.1141      move     x:(r0)+n0,x1      y:(r4)+n4,y1
0.1141      _end_grp
0.1142      move     n0,b1
0.1143      lsr      b      n2,a1      ;divide butterflies per group by two
0.1144      lsl      a      b1,n0      ;multiply groups per pass by two
0.1145      move     a1,n2
0.1146      _end_pass
0.1147
0.1148 ; Do last FFT pass
0.1149      move     n1,n0      ;correct pointer offset for last pass
0.1150      move     #data,r0      ;initialize A input pointer
0.1151      move     r0,r4      ;initialize A output pointer

```

```

0.1152      lua      (r0)+,r1                ;initialize B input pointer
0.1153      move    #coef,r6                ;initialize C input pointer
0.1154      lua      (r1)-n1,r5              ;initialize B output pointer
0.1155      move    x:(r1),x1                y:(r6),y0
0.1156      move    x:(r5),a                 y:(r0),b
0.1157
0.1158      do      n2,_lastpass              ;Radix 2 DIT butterfly kernel with one
0.1159                                     ;butterfly per group
0.1160      mac     x1,y0,b                    x:(r6)+n6,x0    y:(r1)+n1,y1
0.1161      macr   -x0,y1,b                   a,x:(r5)+n5    y:(r0),a
0.1162      subl  b,a                          x:(r0),b       b,y:(r4)
0.1163      mac     -x1,x0,b                  x:(r0)+n0,a    a,y:(r5)
0.1164      macr   -y1,y0,b                  x:(r1),x1      y:(r6),y0
0.1165      subl  b,a                          b,x:(r4)+n4    y:(r0),b
0.1166      _lastpass
0.1167      move    a,x:(r5)+n5
0.1168
0.1169      move    #$FFFF,m0                ;DEFAULT LINEAR ARITHMETIC SELECTED
0.1170      move    m0,m1
0.1171      move    m0,m4
0.1172      move    m0,m5
0.1173      move    m0,m6
0.1174
0.1175      ENDM
0.1176
0.1177 ;***** PROGRAM START *****
0.1178 BEGIN  ori      #$3,mr                ;ONLY INT WITH IPL3 ALLOWED
0.1179      bclr   #2,omr                    ;DISABLE ROM TABLES
0.1180      movep  #1,x:PBC                    ;CONFIGURE PORT B AS HOST PORT
0.1181                                     ;(AS OPPOSED TO GENERAL-PURPOSE I/O)
0.1182      bset   #2,x:HCR                    ;ENABLE VECTORED HOST INTERRUPTS
0.1183      bset   #1,x:HCR                    ;ENABLE XMT INT
0.1184      bset   #0,x:HCR                    ;ENABLE RCV INT
0.1185
0.1186      movep  #$C809,x:IPR                ;GIVE IRQA IPL0, IRQB IPL0, HI IPL1
0.1187                                     ;AND SCI TIMER IPL2(SEE ALSO 8.17)
0.1188      sincos N_FFT,A_Addr              ;INITLIAZE THE VARIOUS LOOKUP TABLES
0.1189      Carrier Ca_Add
0.1190      HAMMING N,Hama
0.1191
0.1192 ;...AND ACTION...
0.1193      movep  #$2000,x:SCR                ;ENABLE TIMER INT
0.1194      movep  #Frate,x:SCCR              ;INT RATE = Fosc[20.48MHz]/(512*Frate)
0.1195      andi   #$FC,mr
0.1196      ori    #$2,mr                    ;ONLY INT WITH IPL2,3 ALLOWED
0.1197
0.1198 ; USE THE FOLLOWING IF FrameR INT NOT USED
0.1199 ;      jsr    CNTSetup                    ;SETUP OF IRQA AND IRQB
0.1200 ;      jsr    DisCard                    ;START UP A/D CONVERTER
0.1201 ;      move   #WkAddr,r0                ;DATA ADDRESS
0.1202 ;      andi   #$FC,mr                    ;UNMASK INTERRUPTS
0.1203
0.1204 ;      andi   #$FC,mr
0.1205 ;      ori    #$1,mr                    ;INT WITH IPL1,2,3 ALLOWED
0.1206 ;      jsr    FromPC                    ;USE WITH To_DSP ON PC
0.1207
0.1208
0.1209 ;## MAIN PROGRAM LOOP ##

```

```

0.1210  MAIN_LOOP
0.1211 ;      jsr      ToPC
0.1212      jmp      MAIN_LOOP
0.1213
0.1214 ;##### SUBROUTINES #####
0.1215 ;SETUP OF PROGRAMABLE INTERVAL TIMER
0.1216  CNTSetup
0.1217
0.1218 ;SET WAIT STATES TO SLOW SPEED FOR 8254-2 DEVICE
0.1219      movep   #$2222,x:$FFFE      ;(x,y,p,io)
0.1220
0.1221 ;TIMING SETUP IRQA
0.1222      move    #%110100,r0          ;COUNTER 0 IN MODE 2
0.1223      move    #CREG,r1             ;CONTROL REG ADDR
0.1224      jsr     wrtimer
0.1225      clr     a                    #>$FF,x0
0.1226      move    #>T0,a1             ;INT PERIOD:5.12MHz/128=40kHz
0.1227      and     x0,a                 #CT0,r1
0.1228      move    a1,r0
0.1229      jsr     wrtimer              ;LSB TO COUNTER
0.1230      clr     a                    #>$FF00,x0
0.1231      move    #>T0,a1
0.1232      and     x0,a
0.1233      rep     #8
0.1234      asr     a
0.1235      move    a1,r0
0.1236      jsr     wrtimer              ;MSB TO COUNTER
0.1237
0.1238 ;TIMING SETUP IRQB
0.1239      move    #%1110100,r0        ;COUNTER 1 IN MODE 2
0.1240      move    #CREG,r1             ;CONTROL REG ADDR
0.1241      jsr     wrtimer
0.1242      clr     a                    #>$FF,x0
0.1243      move    #>N,a1              ;INT PERIOD:40kHz/N
0.1244      and     x0,a                 #CT1,r1
0.1245      move    a1,r0
0.1246      jsr     wrtimer              ;LSB TO COUNTER
0.1247      clr     a                    #>$FF00,x0
0.1248      move    #>N,a1
0.1249      and     x0,a
0.1250      rep     #8
0.1251      asr     a
0.1252      move    a1,r0
0.1253      jsr     wrtimer              ;MSB TO COUNTER
0.1254
0.1255 ;SET WAIT STATES FOR ADDED SPEED
0.1256      movep   #$0010,x:$FFFE      ;(x,y,p,io)
0.1257      rts
0.1258
0.1259 ;#####
0.1260 ;WRITING TO 8254 TIMER
0.1260      wrtimer  move    r0,p:(r1)
0.1261      rep     #4
0.1262      nop
0.1263      rts
0.1264
0.1265 ;#####
0.1266 ;DISCARD INITIAL VALUES FROM THE ADC

```

```

0.1267 DisCard move    #S08,a1
0.1268         move    a1,p:Port    ;SET UP CH0
0.1269         rep     #Delay        ;ALLOW MUX TO SETTLE
0.1270         nop     ;(APPROX 2.5us DELAY BETWEEN CHANNELS)
0.1271         move    p:ADC,a1     ;READ CH0 AND DISCARD VALUE
0.1272         rts
0.1273
0.1274 ;#####
0.1275 RawToPC jclr    #3,x:HSR,_NReI ;HF0 (HOST NOT READY)
0.1276
0.1277 ;TRANSFER OF RAW DATA. (INTEGERS ARE TRANSFERED)
0.1278 ;FIRST X AND THEN Y MEM IS SHIFTED FROM WkAddr TO Wk_Addr+N
0.1279 ;IN X MEM AND THEN SHIFTED DOWN ACCORDING TO THE VALUE OF
0.1280 ;In_SHFT. THE DESIRED POSITION IS THE LOWEST
0.1281
0.1282         move    #WkAddr,r2
0.1283         move    #WkAddr+N,r4
0.1284
0.1285         do     #N,_xm
0.1286         move    x:(r2)+,a
0.1287         rep     #12-In_SHFT
0.1288         asr     a
0.1289         move    a,x:(r4)+
0.1290 _xm
0.1291         move    #RawDisp,r3
0.1292         jsr     dmatx           ;TRANSFER DATA
0.1293
0.1294         move    #WkAddr,r2
0.1295         move    #WkAddr+N,r4
0.1296
0.1297         do     #N,_ym
0.1298         move    y:(r2)+,a
0.1299         rep     #12-In_SHFT
0.1300         asr     a
0.1301         move    a,x:(r4)+
0.1302 _ym
0.1303         move    #RawDisp,r3
0.1304         jsr     dmatx           ;TRANSFER DATA
0.1305
0.1306 _NReI
0.1307         rts
0.1308
0.1309 ;#####
0.1310 ToPC jclr    #3,x:HSR,_NReI ;HF0 (HOST NOT READY)
0.1311         move    #DsAddr1,r3
0.1312         jsr     dmatx           ;TRANSFER DATA
0.1313
0.1314         move    #DsAddr2,r3
0.1315         jsr     dmatx           ;TRANSFER DATA
0.1316 _NRe
0.1317         rts
0.1318
0.1319 ;#####
0.1320 ;DMA TRANSFER TO PC
0.1321 dmatx bset    #3,x:HCR           ;HF2 (HOST MAY REQUEST DMA)
0.1322 _bf jclr    #7,x:HSR,_tf         ;WAIT FOR DMA PROCESS TO BE INITIATED
0.1323         jmp     _bg             ;DMA PROCESS INITIATED
0.1324 _tf jset    #3,x:HSR,_bf         ;HOST DO WANT DMA

```

```

0.1325  _bg      jset    #7,x:HSR_bg      ;WAIT FOR COMPLETION OF DMA
0.1326      bclr    #3,x:HCR      ;HF2 (NO DMA TRANSFER)
0.1327      rts
0.1328
0.1329 ;#####
0.1330  FromPC  move    #DMAIn1,r3
0.1331      jsr     dmarx
0.1332      move    #DMAIn2,r3
0.1333      jsr     dmarx
0.1334
0.1335      move    #Wsp,r0
0.1336      move    #WkAddr,r1
0.1337      do     #InNum,_mv
0.1338      move    x:(r0)+,a
0.1339      move    a,y:(r1)+
0.1340  _mv
0.1341
0.1342      jsr     WORK
0.1343      rts
0.1344
0.1345 ;#####
0.1346 ;DMA TRANSFER FROM PC
0.1347  dmarx
0.1348  _bs      jclr    #7,x:HSR_bs      ;WAIT FOR DMA PROCESS TO BE INITIATED
0.1349  _bt      jset    #7,x:HSR_bt      ;WAIT FOR COMPLETION OF DMA
0.1350      rts
0.1351
0.1352 ;#####
0.1353  ENDIMEM
0.1354      org     p:$E000
0.1355
0.1356
0.1357 ;##### DO NUMBERCRUNCHING #####
0.1358
0.1359  WORK
0.1360      HAM_XY  WkAddr      ;MPY WITH HAMMING WINDOW
0.1361      MulCar
0.1362      Add_DQ
0.1363
0.1364      ZeroPad  WkAddr+N,(N_FFT-N)      ;ZEROPAD
0.1365      fft2c   N_FFT,WkAddr,A_Addr      ;DO COMPLEX FFT
0.1366      BitRev  WkAddr,Wsp,N_FFT/4      ;BIT REVERSE OUTPUT
0.1367      PerG    WkAddr,N_FFT/4          ;PERIODOFRAM
0.1368      ChrShft WkAddr,N_FFT/4          ;PREPARE FOR TRANSFER
0.1369
0.1370      rts
0.1371  END
0.1372 ;***** PROGRAM END *****

```

PC SOURCE CODE LISTING (1.0000-3.0067)

```

1.0000 /* *****
1.0001 /*
1.0002 /*          CW DOPPLER ULTRASOUND
1.0003 /*
1.0004 /* *****
1.0005
1.0006 // Copyright(C) L.SMITH
1.0006 // Department of Biomedical Engineering
1.0007 // University of Cape Town
1.0008 // Observatory
1.0009 // Cape Town
1.0010 // New South Africa
1.0011 // (021) 471250 x 235
1.0012
1.0013 // MISC FUNCTIONS FOR WA_MAIN.C
1.0014
1.0015 #include "interface.h"
1.0016 #include "wa_main.h"
1.0017
1.0017 //***** GLOBAL VARIABLES *****
1.0018 extern unsigned long int StartT,ElapsT;
1.0019 extern int far TW,TH;
1.0020 extern int MaxX,MaxY,clip,Ret_Main;
1.0021
1.0022 //***** STRUCTURE DECLARATIONS *****
1.0023
1.0024 struct MenuWindow Menu[MNum]=
1.0025 {          //MAIN MENU (0)
1.0026     { " Capture Save Import MeanF LookW Ex Quit",
1.0027     0,0,639,28,14,6,2,
1.0028     "CSIMLEQ",
1.0028     {&Go,&Menu_Man,&Get_FileN,&MeanF,&LookW,&Menu_Man,&Quit},
1.0029     {NULL,3,1,NULL,NULL,1,NULL},
1.0030     NULL
1.0031     },
1.0032
1.0033     //SUB MENU (1)
1.0034     { "Sub\nMenu\nExample",
1.0035     10,40,90,100,14,6,1,
1.0036     "SME",
1.0037     {&Menu_Man,&Menu_Man,&Menu_Man},
1.0038     {2,2,2},
1.0039     NULL
1.0039     },
1.0040
1.0041     //SUB_SUB MENU (2)
1.0042     { "Another\nLevel\nDown",
1.0043     60,120,140,180,14,6,1,
1.0044     "ALD",
1.0045     {&Test,&Test,&Test},
1.0046     {NULL,NULL,NULL},
1.0047     NULL
1.0048     },
1.0049
1.0050     //SAVE SUB MENU (3)
1.0050     { "Overwrite\nNew",
1.0051     67,40,157,85,14,6,1,
1.0052     "ON",

```

```

1.0053      {&WriteFile,&Get_FileN},
1.0054      {NULL,0},
1.0055      NULL
1.0056      }
1.0057
1.0058 };
1.0059
1.0060 struct DisplayWindow DisW[DWNuM]=
1.0061 {
1.0062     {1,"SPECTROGRAM","Time (sec)","Frequency (kHz)",
1.0063     14,0,15,NULL,
1.0064     75,76,560,418, // viewing area 2x256x400
1.0065     {NULL,NULL,NULL,NULL},
1.0066     {NULL,NULL,NULL,NULL}, //XOFF (graph wraparound)
1.0067     NULL
1.0068     },
1.0069     {1,"DSP56001 Output 1","", "",
1.0070     14,0,15,6,
1.0071     5,329,319,479,
1.0072     {NULL,NULL,NULL,NULL},
1.0073     {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0074     NULL
1.0075     },
1.0076     {1,"DSP56001 Output 2","", "",
1.0077     14,0,15,6,
1.0078     322,329,636,479,
1.0079     {NULL,NULL,NULL,NULL},
1.0080     {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0081     NULL
1.0082     },
1.0083     {1,"X-Memory","Time","Amplitude",
1.0084     14,0,15,6,
1.0085     5,64,191,214,
1.0086     {NULL,NULL,NULL,NULL},
1.0087     {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0088     NULL
1.0089     },
1.0090
1.0091     {1,"Y-Memory","Time","Amplitude",
1.0092     14,0,15,6,
1.0093     5,280,191,430,
1.0094     {NULL,NULL,NULL,NULL},
1.0095     {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0096     NULL
1.0097     },
1.0098     {1,"Spectrum-F","Frequency","Amplitude",
1.0099     14,0,15,6,
1.0100     205,64,391,214,
1.0101     {NULL,NULL,NULL,NULL},
1.0102     {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0103     NULL
1.0104     },
1.0105     {1,"Spectrum-R","Frequency","Amplitude",

```

```

1.0106         14,0,15,6,
1.0107         205,280,391,430,
1.0108         {NULL,NULL,NULL,NULL},
1.0109         {NULL,NULL,NULL,NULL}, //ymax,xoff,width
1.0110         NULL
1.0111     }
1.0112
1.0113 };
1.0114
1.0115 #####
1.0115 void Init_Graph(void)
1.0116 {
1.0117 #define SVGA             134 // Code given by BGI toolkit
1.0118 #define SVGA640x480v5F  2 // Mode 5Fh 640x480x256
1.0119
1.0120 int GraphDriver,GraphMode,ErrorCode;
1.0121
1.0122 GraphDriver = SVGA;
1.0123 GraphMode = SVGA640x480v5F;
1.0124 installuserdriver("PARADISE",NULL);
1.0125 initgraph(&GraphDriver,&GraphMode,"");
1.0126 ErrorCode = graphresult();
1.0126 if (ErrorCode != grOk)
1.0127 {
1.0128     sound(1000);
1.0129     delay(500);
1.0130     nosound();
1.0131     exit(20);
1.0132 }
1.0133
1.0134 return;
1.0135 }
1.0136
1.0136 #####
1.0137 void Test(void)
1.0138 // DEBUGGING
1.0139 {
1.0140     Ret_Main = 1;
1.0141     return;
1.0142 }
1.0143
1.0144 #####
1.0145 void Listprint(int l,int t,int fcol,char *list,char *PossKey)
1.0146 // PRINTS THE GIVEN OPTIONS AND HIGHLIGHTS THE HOTKEY.
1.0147 {
1.0147     char str[200];
1.0148     int xl,xloc,yloc;
1.0149     char *strptr,*charptr,*keyptr,shortstr[2];
1.0150
1.0151     xl = l+TW;
1.0152     xloc = xl;
1.0153     yloc = t+(int)(TH/2);
1.0154
1.0155     strcpy(str,list);
1.0156     setcolor(fcol);
1.0157
1.0158     charptr = str;
1.0158     keyptr = PossKey;

```

```

1.0159  strptr = strtok(str, "\n");
1.0160
1.0161  do
1.0162  {
1.0163      do
1.0164      {
1.0165          shortstr[0] = *charptr;
1.0166          shortstr[1] = '\0';
1.0167
1.0168          if (*keyptr == shortstr[0])
1.0169          {
1.0169              setcolor(15);
1.0170              keyptr++;
1.0171          }
1.0172
1.0173          outtextxy(xloc,yloc,shortstr);
1.0174          setcolor(fcol);
1.0175
1.0176          xloc += TW;
1.0177          charptr++;
1.0178      }while(*charptr != '\0');
1.0179
1.0180      xloc = xl-TW; //ALLOW FOR EXTRA NULL CHAR INSERTED
1.0180      yloc += TH+2;
1.0181      strptr = strtok(NULL, "\n");
1.0182
1.0183  }while(strptr != NULL);
1.0184
1.0185  return;
1.0186 }
1.0187
1.0188 //#####
1.0189 /*
1.0190 TYPE OF OPERATIONS THAT MAY BE PERFORMED ON THE VARIOUS WINDOWS :
1.0191
1.0191  (1) OPEN WITH SINGLE BORDER
1.0192  (2) OPEN WITH DOUBLE BORDER
1.0193  (3) "SOLID" BORDER AROUND WINDOW
1.0194  (4) "BROKEN" BORDER AROUND WINDOW
1.0195  (5) CLEARING THE WINDOW
1.0196  (6) RECTANGLE OF GIVEN COLOUR
1.0197
1.0198 */
1.0199
1.0200 void GrWin(int l,int t,int r,int b,int fcol,int bcol,int type_op)
1.0201 {
1.0202     setcolor(fcol);
1.0202     switch (type_op)
1.0203     {
1.0204         case 1 : {
1.0205             setfillstyle(SOLID_FILL,bcol);
1.0206             bar(l,t,r,b);
1.0207             rectangle(l,t,r,b);
1.0208         }break;
1.0209
1.0210         case 2 : {
1.0211             setfillstyle(SOLID_FILL,bcol);
1.0212             bar(l,t,r,b);

```

```

1.0212             rectangle(l,t,r,b);
1.0213             rectangle(l+3,t+3,r-3,b-3);
1.0214             }break;
1.0215
1.0216             case 3 : {
1.0217                 setlinestyle(0,0,NORM_WIDTH);
1.0218                 rectangle(l,t,r,b);
1.0219             }break;
1.0220
1.0221             case 4 : {
1.0222                 setcolor(bcol);
1.0223                 rectangle(l,t,r,b);
1.0223                 setcolor(fcol);
1.0224                 setlinestyle(3,0,NORM_WIDTH);
1.0225                 rectangle(l,t,r,b);
1.0226             }break;
1.0227
1.0228             case 5 : {
1.0229                 setfillstyle(SOLID_FILL,bcol);
1.0230                 bar(l+1,t+1,r-1,b-1);
1.0231             }break;
1.0232
1.0233             case 6 : {
1.0234                 setcolor(fcol);
1.0234                 rectangle(l,t,r,b);
1.0235             }break;
1.0236
1.0237         }
1.0238         setlinestyle(0,0,NORM_WIDTH);
1.0239         return;
1.0240 }
1.0241
1.0242 //#####
1.0243 char far *GrSave(int l,int t,int r,int b)
1.0244 // SAVE THE BACKGROUND TO ENABLE RESTORATION OF THE
1.0245 // ORIGINAL IMAGE.
1.0245 {
1.0246     unsigned size;
1.0247     char far *Bitmapptr;
1.0248
1.0249     if ((size = imagesize(l,t,r,b)) <= 0)
1.0250     {
1.0251         sound(500);
1.0252         delay(500);
1.0253         nosound();
1.0254         Bitmapptr=NULL;
1.0255         return(Bitmapptr);
1.0256     }
1.0256     if ((Bitmapptr = (char far *)farmalloc(size)) == (char far *)NULL)
1.0257     {
1.0258         sound(500);
1.0259         delay(500);
1.0260         nosound();
1.0261         return(Bitmapptr);
1.0262     }
1.0263     getimage(l,t,r,b,Bitmapptr);
1.0264     return(Bitmapptr);
1.0265 }

```

```

1.0266
1.0267 #####
1.0267 void GrRestore(int L,int t,char far *Bitmapptr)
1.0268 // RESTORE THE BACKGROUND IMAGE.
1.0269 {
1.0270     putimage(L,t,Bitmapptr,COPY_PUT);
1.0271     farfree(Bitmapptr);
1.0272     return;
1.0273 }
1.0274
1.0275 #####
1.0276 void DefaultWin(void)
1.0277 // OPENS DEFAULT DISPLAY WINDOWS.
1.0277 // NOTE THAT IMAGESIZE() WILL NOT ALLOW US TO SAVE WINDOWS
1.0278 // LARGER THAN 64k
1.0279 {
1.0280 #define D DisW
1.0281     GrWin(D[0].L,D[0].T,D[0].R,D[0].B,D[0].fcol,D[0].bcol,1);
1.0282     Spec_Prep(0);
1.0283
1.0284     /* WINDOWS USED FOR DEBUGGING
1.0285     GrWin(D[1].L,D[1].T,D[1].R,D[1].B,D[1].fcol,D[1].bcol,1);
1.0286     XY_Prep(1);
1.0287
1.0288     GrWin(D[2].L,D[2].T,D[2].R,D[2].B,D[2].fcol,D[2].bcol,1);
1.0288     XY_Prep(2);
1.0289     */
1.0290
1.0291     return;
1.0292 #undef D DisW
1.0293 }
1.0294
1.0295 #####
1.0296 void Menu_Man(int Num)
1.0297 // MENU MANAGER. THE VARIOUS MENU WINDOWS ARE CALLED FROM
1.0298 // HERE IN A RECURSIVE MANNER.
1.0299 {
1.0299 #define Mn Menu[Num]
1.0300     char Option;
1.0301     int OptionNr;
1.0302
1.0303     if ((Mn.Mapptr = GrSave(Mn.L,Mn.T,Mn.R,Mn.B)) != NULL)
1.0304     {
1.0305         GrWin(Mn.L,Mn.T,Mn.R,Mn.B,Mn.fcol,Mn.bcol,Mn.type_op);
1.0306         Listprint(Mn.L,Mn.T,Mn.fcol,Mn.MenuList,Mn.KeyList);
1.0307         if (Num == 0)DefaultWin();
1.0308         while (((Option = toupper(getch())) != 27) ||
1.0309             ((Option == 27) && (Num == 0)))
1.0310         {
1.0311             if (Option == 0x0)
1.0312                 Option = getch();
1.0312             OptionNr = strcspn(Mn.KeyList,&Option);
1.0313             if (OptionNr < strlen(Mn.KeyList))
1.0314             {
1.0315                 (*Mn.ProcPointer[OptionNr])(Mn.ArgList[OptionNr]);
1.0316                 if (Num == 0) Ret_Main = 0;
1.0317                 if (Ret_Main == 1)break;
1.0318             }

```

```

1.0319     }
1.0320     if (Num != 0)GrRestore(Mn.L,Mn.T,Mn.Mapptr);
1.0321 }
1.0321
1.0322 return;
1.0323 #undef Mn
1.0324 }
1.0325
1.0326 #####
1.0327 void BackWin(int col)
1.0328 // BACKGROUND DISPLAY.
1.0329 {
1.0330     setfillstyle(HATCH_FILL,col);
1.0331     bar(0,36,MaxX,MaxY);
1.0332     setcolor(col);
1.0332     rectangle(0,36,MaxX,MaxY);
1.0333     setfillstyle(SOLID_FILL,col);
1.0334     return;
1.0335 }
1.0336
1.0337 #####
1.0338 void Spec_Prep(int n)
1.0339 // PREPARE THE SPECTROGRAM DISPLAY WINDOW. AXIS,TITLE AND
1.0340 // THE COLORBAR ARE DISPLAYED.
1.0341 {
1.0342 #define D DisW
1.0342
1.0343     int dist,*intptr,ylabpos,xlabpos,titlepos;
1.0344     int xst,xdiv,yst,ydiv,i,j,m;
1.0345     char val[3],str[80];
1.0346
1.0347     intptr = D[n].Act;
1.0348     dist = 3*TH;
1.0349
1.0350     xlabpos = (int)((D[n].R+D[n].L-textwidth(D[n].xlabel))/2);
1.0351     titlepos = (int)((D[n].R+D[n].L-textwidth(D[n].title))/2);
1.0352     ylabpos = (int)((D[n].B+D[n].T+textwidth(D[n].ylabel))/2);
1.0353
1.0353     setcolor(D[n].fcol);
1.0354     rectangle(D[n].L+dist,D[n].T+dist,D[n].R-dist,D[n].B-dist);
1.0355
1.0356     setcolor(D[n].Icol);
1.0357     outtextxy(xlabpos,D[n].B-TH-1,D[n].xlabel);
1.0358     outtextxy(titlepos,D[n].T+TH,D[n].title);
1.0359     settxtstyle(DEFAULT_FONT,VERT_DIR,0);
1.0360     outtextxy(D[n].L+TH+1,ylabpos,D[n].ylabel);
1.0361
1.0362     *intptr = D[n].L+dist+1; intptr++; // SET ACTIVE VIEWING COORD.
1.0363     *intptr = D[n].T+dist+1; intptr++;
1.0364     *intptr = D[n].R-dist-1; intptr++;
1.0364     *intptr = D[n].B-dist-1;
1.0365
1.0366     rectangle(D[n].Act[2]+10,D[n].Act[1]+47,D[n].R-10,D[n].Act[3]-48);
1.0367     yst = D[n].Act[3]-49;
1.0368     for(i = PalOff;i < PalOff+ColLevels; i++)
1.0369     {
1.0370         for(m = 0;m < 10; m++)
1.0371         {

```

```

1.0372         for(j = D[n].Act[2]+11;j < D[n].R-11;j++)
1.0373         {
1.0374             putpixel(j,yst,i);
1.0375         }
1.0375         yst--;
1.0376     }
1.0377 }
1.0378
1.0379 setcolor(D[n].fcol);
1.0380 settextstyle(DEFAULT_FONT,HORIZ_DIR,0);
1.0381
1.0382 D[n].MisC[0] = D[n].Act[0];
1.0383 line(D[n].Act[0],(D[n].Act[1]+D[n].Act[3])/2,
1.0384 D[n].Act[2],(D[n].Act[1]+D[n].Act[3])/2);
1.0385
1.0386 setcolor(D[n].lcol);
1.0386 yst = (int)((D[n].Act[1]+D[n].Act[3])/2-5);
1.0387 ydiv = (int)((D[n].Act[3]-D[n].Act[1])/10);
1.0388 for(i = 0;i <= 5; i++)
1.0389 {
1.0390     outtextxy(D[n].Act[0]-18,yst,itoa(i,val,10));
1.0391     yst -= ydiv;
1.0392 }
1.0393
1.0394 yst = (int)((D[n].Act[1]+D[n].Act[3])/2-5);
1.0395 for(i = 1;i <= 5; i++)
1.0396 {
1.0397     yst += ydiv;
1.0397     outtextxy(D[n].Act[0]-26,yst,itoa(-i,val,10));
1.0398 }
1.0399
1.0400 xst = D[n].Act[0]-3;
1.0401 xdiv = (int)((D[n].Act[2]-D[n].Act[0])/6);
1.0402 for(i = 0;i <= 30;i += 5)
1.0403 {
1.0404     outtextxy(xst,D[n].Act[3]+12,gcvt((double)(i)/10,4,str));
1.0405     xst += xdiv;
1.0406 }
1.0407
1.0407 return;
1.0408 #undef D DisW
1.0409 }
1.0410
1.0411 #####
1.0412 void XY_Prep(int n)
1.0413 // PREPARES X-Y DISPLAY WINDOWS.
1.0414 {
1.0415 #define D DisW
1.0416     int dist,*intptr,ylabpos,xlabpos,titlepos;
1.0417
1.0418     intptr = D[n].Act;
1.0418     dist = 2*TH;
1.0419
1.0420     xlabpos = (int)((D[n].R+D[n].L-textwidth(D[n].xlabel))/2);
1.0421     titlepos = (int)((D[n].R+D[n].L-textwidth(D[n].title))/2);
1.0422     ylabpos = (int)((D[n].B+D[n].T+textwidth(D[n].ylabel))/2);
1.0423
1.0424     setcolor(D[n].fcol);

```

```

1.0425   rectangle(D[n].L+dist,D[n].T+dist,D[n].R-dist,D[n].B-dist);
1.0426
1.0427   setcolor(D[n].Icol);
1.0428   outtextxy(xlabpos,D[n].B-TH-1,D[n].xlabel);
1.0429   outtextxy(titlepos,D[n].T+(TH/2),D[n].title);
1.0429   settextstyle(DEFAULT_FONT,VERT_DIR,0);
1.0430   outtextxy(D[n].L+TH+1,ylabpos,D[n].ylabel);
1.0431
1.0432   *intptr = D[n].L+dist+1; intptr++; // SET ACTIVE VIEWING COORD.
1.0433   *intptr = D[n].T+dist+1; intptr++;
1.0434   *intptr = D[n].R-dist-1; intptr++;
1.0435   *intptr = D[n].B-dist-1;
1.0436
1.0437   setcolor(D[n].fcol);
1.0438   settextstyle(DEFAULT_FONT,HORIZ_DIR,0);
1.0439
1.0440   D[n].MisC[0] = (int)(D[n].Act[3]-D[n].Act[1])/2);
1.0440   D[n].MisC[1] = D[n].Act[0];
1.0441
1.0442   if((n == 3) | (n == 4))
1.0443       D[n].MisC[2] = (int)(D[n].Act[2]-D[n].Act[0])/NRaw);
1.0444   if((n == 5) | (n == 6))
1.0445       D[n].MisC[2] = (int)(D[n].Act[2]-D[n].Act[0])/NR);
1.0446
1.0447   return;
1.0448 #undef D DisW
1.0449 }
1.0450
1.0451 #####
1.0451 unsigned long int TimeElapsed(void)
1.0452 // TIME CALCULATION USED DURING DEBUGGGING PROCESS.
1.0453 {
1.0454   unsigned long int t,*tix;
1.0455
1.0456   tix = (unsigned long int *)0x46C;
1.0457   t = (*tix)*55;
1.0458
1.0459   return(t-StartT);
1.0460 }
1.0461
1.0462 #####
1.0462 void TimerStart(void)
1.0463 // TIME CALCULATION USED DURING DEBUGGGING PROCESS.
1.0464 {
1.0465   unsigned long int *tix;
1.0466
1.0467   tix = (unsigned long int *)0x46C;
1.0468   StartT = (*tix)*55;
1.0469   return;
1.0470 }
1.0471
1.0472 #####
1.0472 void Quit(void)
1.0473 {
1.0474   fcloseall();
1.0475   cleardevice();
1.0476   closegraph();
1.0477   exit(0);

```

1.0478 }
1.0479
1.0480 //#####

```

2.0000 //*****
2.0001 //          ACQUISITION AND ANALYSIS OF
2.0002 //          LOWER LIMB FLOW WAVEFORMS
2.0003 //*****
2.0003
2.0004 // Copyright(C) L.SMITH
2.0005 // Department of Biomedical Engineering
2.0005 // University of Cape Town
2.0006 // Observatory
2.0006 // Cape Town
2.0007 // New South Africa
2.0008 // (021) 471250 x 235
2.0008
2.0009 /* PLEASE NOTE THE FOLLOWING :
2.0010
2.0010 1.FONT=DEFAULT_FONT
2.0011 2.USERCHARSIZE = 8X14 FOR MENUS
2.0011 3.SETUSERCHARSIZE DOES NOT WORK WITH PARADISE DRIVER !!
2.0012 4.VGA RESOLUTION : 640X480X256 COLOURS IS USED (MODE $5F)
2.0013 5.THE REVERSE FREQUENCY ARRAY IS REVERSED : DUE TO TYPE OF
2.0013 DEMODULATION USED ON DSP BOARD.
2.0014
2.0014 */
2.0015
2.0016 #include "intface.h"
2.0016 #include "wa_main.h"
2.0017 #define      MaxRow 200
2.0017
2.0018 //***** CONSTANTS *****
2.0019 //CARDADDRESS
2.0019 const word CardAddr = 0x2b0;
2.0020
2.0021 //2^23 : 24 BIT TRANSFERS FROM DSP56001
2.0021 const unsigned long MAXVALUE = 8388608;
2.0022
2.0022 typedef union _addr
2.0023 {
2.0024     dword l;
2.0024     struct
2.0025     {
2.0025         word lo;
2.0026         word hi;
2.0027     }w;
2.0027 }_addrtype;
2.0028
2.0028 //***** GLOBAL VARIABLES *****
2.0029 int      far TW,TH,srow,frow;
2.0030 int      MaxX,MaxY,clip,Ret_Main;
2.0030 int      Shft,ChkISR,HF0,HF2,row;
2.0031 int      ICReg = 0;
2.0032 char    FileName[10] = "";
2.0032 int      Bvel_flag = 0;
2.0033 unsigned long int StartT,ElapsT,maxval;
2.0033
2.0034 int      BVel[MaxRow];
2.0035 char    ForFreq[MaxRow][NR];
2.0035 char    RevFreq[MaxRow][NR];
2.0036 char    *FFreq[MaxRow];

```

```

2.0036 char      *RFreq[MaxRow];
2.0037 int       far XRaw[MaxRow][NRaw];
2.0038 int       far YRaw[MaxRow][NRaw];
2.0038 int       far *Xmem[MaxRow];
2.0039 int       far *Ymem[MaxRow];
2.0039
2.0040 //long      huge ForFreq[NR]; // USEFUL FOR DEBUGGING (24 BIT TRANSFER)
2.0041 //long      huge RevFreq[NR];
2.0041
2.0042 //***** STRUCTURES *****
2.0043 struct palette
2.0043 {
2.0044     int color[ColLevels][3];
2.0044 }pal;
2.0045
2.0046 extern struct MenuWindow Menu[MNum];
2.0046 extern struct DisplayWindow DisW[DWNum];
2.0047
2.0047 //##### MAIN PROGRAM START #####
2.0048 void main(void)
2.0049 {
2.0049     int ChrMax = 128;
2.0050
2.0050     Init_Graph();
2.0051     cleardevice();
2.0052     InitPalette();
2.0052
2.0053     // DEFAULT VALUES USED
2.0053     TW = 8;
2.0054     TH = 14;
2.0055     MaxX = 639;
2.0055     MaxY = 479;
2.0056     maxval = MAXVALUE;
2.0057
2.0057     // INITIALIZE POINTERS
2.0058     for(row=0;row<MaxRow;row++)
2.0058     {
2.0059         FFreq[row] = &ForFreq[row][0];
2.0060         RFreq[row] = &RevFreq[row][0];
2.0060         Xmem[row] = &XRaw[row][0];
2.0061         Ymem[row] = &YRaw[row][0];
2.0061     }
2.0062     row=0;
2.0063
2.0063     // USE THE FOLLOWING WHEN 24 BIT TRANSFERS ARE DONE
2.0064     // Shft = (int)(log(((float)(maxval))/(ColLevels))/log(2));
2.0064     Shft = (int)(log(((float)(ChrMax))/(ColLevels))/log(2));
2.0065
2.0066     // INITIALIZE DSP BOARD
2.0066     do
2.0067     {
2.0068         Reset56(CardAddr,0);
2.0068         ChkISR = inportb(CardAddr+2);
2.0069         if(ChkISR != 6)
2.0069         {
2.0070             sound(500);
2.0071             delay(100);
2.0071             nosound();

```

```

2.0072     }
2.0072     }while(ChkISR != 6);
2.0073     Load56(CardAddr,"dsp_dop");
2.0074     Go56(CardAddr);
2.0074
2.0075     // STARTUP OF USER INTERFACE
2.0075     BackWin(3);
2.0076     Menu_Man(0);
2.0077     return;
2.0077 }
2.0078
2.0079 ##### FUNCTIONS USED #####
2.0079 void InitPalette(void)
2.0080
2.0080 // SET UP PALETTE FOR USE BY SPECTROGRAM
2.0081 // INTENSITIES RANGE FROM
2.0082 // LOW BLUE=>GREEN=>YELLOW=>RED (HIGH)
2.0082 {
2.0083     unsigned int x,k,m,i;
2.0083
2.0084     x = (int)((MaxColInt-MinColInt)/(ColLevels>>2));
2.0085     k = x;
2.0085     m = ColLevels/4;
2.0086
2.0086     for(i=0;i<ColLevels;i++)
2.0087     {
2.0088         pal.color[i][0] = 0;
2.0088         pal.color[i][1] = 0;
2.0089         pal.color[i][2] = 0;
2.0089     }
2.0090
2.0091     for(i=0;i<ColLevels>>2;i++)
2.0091     {
2.0092         pal.color[i][2] = MinColInt+k;
2.0093         pal.color[i+m][1] = MinColInt+k;
2.0093         pal.color[i+2*m][0] = MinColInt+k;
2.0094         pal.color[i+2*m][1] = MinColInt+k;
2.0094         pal.color[i+3*m][0] = MinColInt+k;
2.0095         k += x;
2.0096     }
2.0096
2.0097     for(i = 0; i < ColLevels; i++)
2.0097     {
2.0098         setrgbpalette(PalOff+i,pal.color[i][0],pal.color[i][1],
2.0099         pal.color[i][2]);
2.0099     }
2.0100     return;
2.0100 }
2.0101
2.0102 #####
2.0102 void SOUND(void)
2.0103 // SOUND A TONE. USED FOR BEGUGGING.
2.0104 {
2.0104     sound(500);
2.0105     delay(100);
2.0105     nosound();
2.0106     return;
2.0107 }

```

```

2.0107
2.0108 #####
2.0108 void ValWindow(void)
2.0109
2.0110 // DISPLAY WINDOW OF MAXIMUM VALUE RECIEVED. USED FOR BEGUGGING.
2.0110 {
2.0111     char value[20];
2.0111
2.0112     GrWin(540,35,638,57,15,0,1);
2.0113     outtextxy(550,40,ultoa(maxval,value,10));
2.0113     return;
2.0114 }
2.0115
2.0115 #####
2.0116 void ScaleVal(void)
2.0116
2.0117 // DISPLAY CURRENT MAXIMUM VALUE. USED FOR BEGUGGING.
2.0118 {
2.0118     char kbchar,value[20];
2.0119     long chval = 1000000;
2.0119     int test = 1,l = 540,t = 35,r = 638,b = 57;
2.0120
2.0121     GrWin(l,t,r,b,15,0,4);
2.0121     outtextxy(550,40,ultoa(maxval,value,10));
2.0122     while(!kbhit()){
2.0122
2.0123     while(test)
2.0124     {
2.0124         if (kbhit())
2.0125         {
2.0126             kbchar = getch();
2.0126             if(kbchar == 0xD)
2.0127                 test = 0;
2.0127             else
2.0128             {
2.0129                 switch(kbchar)
2.0129                 {
2.0130                 case 0x49 :{
2.0130
2.0131                     GrWin(l,t,r,b,15,0,5);
2.0131                     if(maxval<(MAXVALUE-chval)) maxval += chval;
2.0132                     outtextxy(l+10,t+5,ultoa(maxval,value,10));
2.0132                     break;
2.0133                 }
2.0133
2.0134                 case 0x51 :{
2.0134
2.0135                     GrWin(l,t,r,b,15,0,5);
2.0135                     if(maxval>chval)
2.0136                         maxval -= chval;
2.0136                     else
2.0137                     {
2.0138                         if(maxval>100000)
2.0138                             maxval -= 100000;
2.0139                         else
2.0140                             if(maxval>10000)
2.0140                                 maxval -= 10000;
2.0141                     }
2.0141                     outtextxy(l+10,t+5,ultoa(maxval,value,10));
2.0142                     break;

```

```

2.0143         }
2.0143
2.0144         case 0x47 :{
2.0144             GrWin(1,t,r,b,15,0,5);
2.0145             maxval = MAXVALUE;
2.0146             outtextxy(1+10,t+5,ultoa(maxval,value,10));
2.0146             break;
2.0147         }
2.0147
2.0148         case 0x4F :{
2.0149             GrWin(1,t,r,b,15,0,5);
2.0149             maxval = 2048;
2.0150             outtextxy(1+10,t+5,ultoa(maxval,value,10));
2.0151             break;
2.0151         }
2.0152     }
2.0152 }
2.0153 }
2.0154 }
2.0154 }
2.0155
2.0155 Shft = (int)(log(((float)(maxval))/(ColLevels))/log(2));
2.0156 GrWin(1,t,r,b,15,0,3);
2.0157 return;
2.0157 }
2.0158
2.0158 #####
2.0159 void Get_FileN(int type_op)
2.0160
2.0160 // GET FILE NAME TO BE IN/EXPORTED.
2.0161 // CHARACTERS BETWEEN VALUES 47 AND 91 ARE ALLOWED.
2.0162 // ENTER WILL ACCEPT OPTION AND ESC WILL RETURN.
2.0162 {
2.0163     char far *Map, fname[2], *ptr;
2.0163     int l = 122, t = 45, r = 212, b = 90, x = l+4, y = t+5;
2.0164
2.0165     if ((Map = GrSave(1,t,r,b)) != NULL)
2.0165         GrWin(1,t,r,b,14,6,1);
2.0166     else return;
2.0166
2.0167     setcolor(14);
2.0168     gprintf(&x,&y, "Filename ?");
2.0168     setcolor(15);
2.0169
2.0169     sprintf(fileName, "");
2.0170     fname[1] = '\0';
2.0171
2.0171     while(((fname[0] = getch()) != 0x0D) && (strlen(fileName) < 8))
2.0172     {
2.0173         if((toupper(fname[0]) > 47) && (toupper(fname[0]) < 91))
2.0173         {
2.0174             outtextxy(x+5,y+5,fname);
2.0174             x += TW;
2.0175             strcat(fileName,fname);
2.0176         }
2.0176
2.0177         if(fname[0] == 27)
2.0177         {

```

```

2.0178             GrRestore(1,t,Map);
2.0179             return;
2.0179         }
2.0180
2.0180         if(fname[0] == 0x08)
2.0181         {
2.0182             if(strlen(fileName) > 0)
2.0182             {
2.0183                 setcolor(6);
2.0183                 x -= TW;
2.0184                 fname[0] = 219;
2.0185                 outtextxy(x+5,y+5,fname);
2.0185                 setcolor(15);
2.0186                 ptr = strchr(fileName,'\0');
2.0187                 ptr--;
2.0187                 *ptr = '\0';
2.0188             }
2.0188         }
2.0189     }
2.0190
2.0190     switch(type_op)
2.0191     {
2.0191         case 0 :WriteFile();break;
2.0192         case 1 :ReadFile();break;
2.0193     }
2.0193
2.0194     GrRestore(1,t,Map);
2.0194     return;
2.0195 }
2.0196
2.0196 //#####
2.0197 void WriteFile(void)
2.0198
2.0198 // WRITE THE OUPUT FILES. ASCII VALUES ARE SAVED TO ENABLE
2.0199 // INPUT TO MATLAB. THE SELECTED FREQUENCY VALUES ARE STORED
2.0199 // IN .SPC EXTENSION AND THE RAW DATA IN A .RAW FILE.
2.0200 // NB! REMEMBER THAT DUE TO THE TYPE OF DEMODULATION USED THE
2.0201 // REVERSE FREQUENCIES ARE IN REVERSED ORDER.
2.0201 {
2.0202     FILE      *stream;
2.0202     int       i,j;
2.0203     char      Path_Name[80] = "",path[50] = "",fext[4] = "",mext[4] = "";
2.0204
2.0204     sprintf(path,"c:\\bc\\files\\patdata\\");
2.0205     sprintf(fext,".spc");
2.0205     strcat(Path_Name,path);
2.0206     strcat(Path_Name,FileName);
2.0207     strcat(Path_Name,fext);
2.0207
2.0208     if((stream = fopen(Path_Name,"wt")) == NULL)
2.0209     {
2.0209         sound(100);
2.0210         delay(500);
2.0210         nosound();
2.0211         return;
2.0212     }
2.0212
2.0213     for(i = 0;i < MaxRow; i++)

```

```

2.0213 {
2.0214     for(j = 0;j < NR; j++)
2.0215         fprintf(stream,"%d ",FFreq[i][j]);
2.0215     fprintf(stream,"\n");
2.0216 }
2.0216
2.0217 fprintf(stream,"\n");
2.0218 for(i = 0; i < MaxRow; i++)
2.0218 {
2.0219     for(j = 0; j<NR; j++)
2.0220         fprintf(stream,"%d ",RFreq[i][j]);
2.0220     fprintf(stream,"\n");
2.0221 }
2.0221 fclose(stream);
2.0222
2.0223 sprintf(Path_Name,"");
2.0223 sprintf(mext,".raw");
2.0224 strcat(Path_Name,path);
2.0224 strcat(Path_Name,FileName);
2.0225 strcat(Path_Name,mext);
2.0226
2.0226 if((stream = fopen(Path_Name,"wt")) == NULL)
2.0227 {
2.0227     sound(100);
2.0228     delay(500);
2.0229     nosound();
2.0229     return;
2.0230 }
2.0230
2.0231 for(i = 0;i < MaxRow; i++)
2.0232 {
2.0232     for(j = 0;j < NRaw; j++)
2.0233         fprintf(stream,"%d ",Xmem[i][j]);
2.0234     fprintf(stream,"\n");
2.0234 }
2.0235
2.0235 fprintf(stream,"\n");
2.0236 for(i = 0;i < MaxRow; i++)
2.0237 {
2.0237     for(j = 0;j < NRaw; j++)
2.0238         fprintf(stream,"%d ",Ymem[i][j]);
2.0238     fprintf(stream,"\n");
2.0239 }
2.0240
2.0240 fclose(stream);
2.0241 SOUND();
2.0241
2.0242 Ret_Main = 1;
2.0243 return;
2.0243 }
2.0244
2.0245 #####
2.0245 void ReadFile(void)
2.0246
2.0246 // READ THE SELECTED INPUT FILES. THE FUNCTIONS ARE THE
2.0247 // REVERSE OF THOSE USED IN WriteFile().
2.0248 {
2.0248     FILE     *stream;

```

```

2.0249 int      ij_raw_available = 1;
2.0249 char    Path_Name[80] = "",path[50] = "",fext[4] = "",mext[4] = "",
2.0250
2.0251 sprintf(path,"c:\\bc\\files\\patdata\\");
2.0251 sprintf(fext,".spc");
2.0252 strcat(Path_Name,path);
2.0252 strcat(Path_Name,FileName);
2.0253 strcat(Path_Name,fext);
2.0254
2.0254 if((stream=fopen(Path_Name,"rt"))==NULL)
2.0255 {
2.0256     sound(100);
2.0256     delay(500);
2.0257     nosound();
2.0257     return;
2.0258 }
2.0259
2.0259 for(i = 0;i < MaxRow; i++)
2.0260 {
2.0260     for(j = 0;j < NR; j++)
2.0261         fscanf(stream,"%d",&FFreq[i][j]);
2.0262     fscanf(stream,"n");
2.0262 }
2.0263
2.0263 fscanf(stream,"n");
2.0264 for(i = 0;i < MaxRow; i++)
2.0265 {
2.0265     for(j = 0;j < NR; j++)
2.0266         fscanf(stream,"%d",&RFreq[i][j]);
2.0267     fscanf(stream,"n");
2.0267 }
2.0268 fclose(stream);
2.0268
2.0269 sprintf(Path_Name,"");
2.0270 sprintf(mext,".raw");
2.0270 strcat(Path_Name,path);
2.0271 strcat(Path_Name,FileName);
2.0271 strcat(Path_Name,mext);
2.0272
2.0273 if((stream = fopen(Path_Name,"rt")) == NULL)
2.0273 {
2.0274     sound(100);
2.0274     delay(500);
2.0275     nosound();
2.0276     for(i = 0;i < MaxRow; i++)
2.0276     {
2.0277         for(j = 0;j < NRaw; j++)
2.0277         {
2.0278             Xmem[i][j] = 0;
2.0279             Ymem[i][j] = 0;
2.0279         }
2.0280     }
2.0281     raw_available = 0;
2.0281 }
2.0282
2.0282 if(raw_available == 1)
2.0283 {
2.0284     for(i=0;i<MaxRow;i++)

```

```

2.0284     {
2.0285         for(j = 0; j < NRow; j++)
2.0285             fscanf(stream, "%d ", &Xmem[i][j]);
2.0286         fscanf(stream, "\n");
2.0287     }
2.0287
2.0288     fscanf(stream, "\n");
2.0288     for(i = 0; i < MaxRow; i++)
2.0289     {
2.0290         for(j = 0; j < NRow; j++)
2.0290             fscanf(stream, "%d ", &Ymem[i][j]);
2.0291         fscanf(stream, "\n");
2.0292     }
2.0292 }
2.0293
2.0293 fclose(stream);
2.0294 row = 0;
2.0295 Bvel_flag = 1;
2.0295 ReDraw();
2.0296 return;
2.0296 }
2.0297
2.0298 //#####
2.0298 void ReDraw(void)
2.0299
2.0299 // FOLLOWING USER SELECTION THE CIRCULAR ARRAYS ARE
2.0300 // LINEARIZED TO ALLOW DATA DISPLAY FROM LEFT TO RIGHT.
2.0301 {
2.0301     int     i;
2.0302     int     far *Rptr[MaxRow];
2.0303     char    *Tptr[MaxRow];
2.0303
2.0304     if(row != 0)
2.0304     {
2.0305         for(i = 0; i < MaxRow; i++)
2.0306         {
2.0306             (row == (MaxRow-1)) ? (row = 0) : (row++);
2.0307             Tptr[i] = FFreq[row];
2.0307             Rptr[i] = Xmem[row];
2.0308         }
2.0309
2.0309         for(i = 0; i < MaxRow; i++)
2.0310         {
2.0310             FFreq[i] = Tptr[i];
2.0311             Xmem[i] = Rptr[i];
2.0312         }
2.0312
2.0313         for(i = 0; i < MaxRow; i++)
2.0313         {
2.0314             (row == (MaxRow-1)) ? (row = 0) : (row++);
2.0315             Tptr[i] = RFreq[row];
2.0315             Rptr[i] = Ymem[row];
2.0316         }
2.0317
2.0317         for(i = 0; i < MaxRow; i++)
2.0318         {
2.0318             RFreq[i] = Tptr[i];
2.0319             Ymem[i] = Rptr[i];

```

```

2.0320     }
2.0320 }
2.0321
2.0321 DisW[0].MisC[0]=DisW[0].Act[0];
2.0322 for(row=0;row<MaxRow;row++)
2.0323     Spectrogram(row);
2.0323
2.0324 row = 0;
2.0324 SOUND();
2.0325 return;
2.0326 }
2.0326
2.0327 #####
2.0328 void Spectrogram(int row)
2.0328
2.0329 // DISPLAY THE SPECTROGRAM IN NEAR REAL TIME ON THE USER SCREEN.
2.0329 // NOTE DIRECTION OF REVERSE DISPLAY. THIS IS DUE TO THE
2.0330 // FREQUENCY DOMAIN PROCESS USED DURING DEMODULATION.
2.0331 // DUE TO GRAPHIC SPEED LIMITATIONS THE ENTIRE SPECTRUM WINDOW
2.0331 // CAN NOT BE SCROLLED IN THE DESIRED TIME FRAME AND THEREFORE
2.0332 // A CIRCULAR SCROLLING TECHNIQUE IS USED.
2.0332 {
2.0333     int     ij = 0,x = DisW[0].MisC[0];
2.0334     int     y1 = DisW[0].Act[1],y2 = DisW[0].Act[3];
2.0334     char    *TF = FFreq[row];
2.0335     char    *TR = RFreq[row];
2.0335
2.0336     for(i = 0;i < (NR << 1);i += 2)
2.0337     {
2.0337         putpixel(x,y1+i,(TF[NR-j-1]>>Shft)+PalOff);
2.0338         putpixel(x,y2-i,(TR[j]>>Shft)+PalOff);
2.0339         j++;
2.0339     }
2.0340
2.0340 (x == (DisW[0].Act[0] + VLine_DWn-2)) ? (DisW[0].MisC[0] =
2.0341 DisW[0].Act[0]) : (DisW[0].MisC[0] += 2);
2.0342 return;
2.0342 }
2.0343
2.0343 #####
2.0344 void signal_view(long huge *signal,int n,int nr)
2.0345
2.0345 // DISPLAY DATA (32 BIT) RECEIVED FROM DSP BOARD IN A SELECTED
2.0346 // WINDOW. USED DURING DEBUGGING.
2.0346 {
2.0347     int     i,xoff;
2.0348     long    yval[NRaw];
2.0348
2.0349     setviewport(DisW[n].Act[0],DisW[n].Act[1],DisW[n].Act[2],
2.0350 DisW[n].Act[3],clip);
2.0350     clearviewport();
2.0351     setviewport(0,0,MaxX,MaxY,clip);
2.0351     xoff = DisW[n].MisC[1];
2.0352
2.0353     for(i = 0;i < nr; i++)
2.0353     {
2.0354         yval[i] = DisW[n].Act[1]+(long)(-(((float)(signal[i])/maxval)*
2.0354 DisW[n].MisC[0])-DisW[n].MisC[0]));

```

```

2.0355 }
2.0356
2.0356 setcolor(DisW[n].Dcol);
2.0357 moveto(xoff,yval[0]);
2.0357 xoff += DisW[n].MisC[2];
2.0358
2.0359 for (i = 1;i < nr; i++)
2.0359 {
2.0360     lineto(xoff,yval[i]);
2.0360     xoff+=DisW[n].MisC[2];
2.0361 }
2.0362 return;
2.0362 }
2.0363
2.0364 #####
2.0364 void I_signal_view(int *signal,int n,int nr)
2.0365
2.0365 // DISPLAY DATA (16 BIT) RECEIVED FROM DSP BOARD IN A SELECTED
2.0366 // WINDOW. USED DURING DEBUGGING.
2.0367 {
2.0367     int i,xoff;
2.0368     int yval[NRaw],A_Dmax = 2048;
2.0368
2.0369     setviewport(DisW[n].Act[0],DisW[n].Act[1],DisW[n].Act[2],
2.0370     DisW[n].Act[3],clip);
2.0370     clearviewport();
2.0371     setviewport(0,0,MaxX,MaxY,clip);
2.0371     xoff = DisW[n].MisC[1];
2.0372
2.0373     for (i = 0;i < nr; i++)
2.0373     {
2.0374         yval[i]=DisW[n].Act[1]+(int)(-(((float)(signal[i])/A_Dmax)*
2.0375         DisW[n].MisC[0])-DisW[n].MisC[0]));
2.0375     }
2.0376
2.0376     setcolor(DisW[n].Dcol);
2.0377     moveto(xoff,yval[0]);
2.0378     xoff += DisW[n].MisC[2];
2.0378
2.0379     for (i = 1;i < nr; i++)
2.0379     {
2.0380         lineto(xoff,yval[i]);
2.0381         xoff += DisW[n].MisC[2];
2.0381     }
2.0382     return;
2.0382 }
2.0383
2.0384 #####
2.0384 void C_signal_view(char *signal,int n,int nr)
2.0385
2.0385 // DISPLAY DATA (8 BIT) RECEIVED FROM DSP BOARD IN A SELECTED
2.0386 // WINDOW. USED DURING DEBUGGING.
2.0387 {
2.0387     int i,xoff;
2.0388     int yval[NRaw],Cmax = 128;
2.0389
2.0389     setviewport(DisW[n].Act[0],DisW[n].Act[1],DisW[n].Act[2],
2.0390     DisW[n].Act[3],clip);

```

```

2.0390 clearviewport();
2.0391 setviewport(0,0,MaxX,MaxY,clip);
2.0392 xoff = DisW[n].MisC[1];
2.0392
2.0393 for (i = 0; i < nr; i++)
2.0393 {
2.0394     yval[i]=DisW[n].Act[1]+(int)(-(((float)(signal[i])/Cmax)*
2.0395     DisW[n].MisC[0])-DisW[n].MisC[0]));
2.0395 }
2.0396
2.0397 setcolor(DisW[n].Dcol);
2.0397 moveto(xoff,yval[0]);
2.0398 xoff += DisW[n].MisC[2];
2.0398 for (i = 1; i < nr; i++)
2.0399 {
2.0400     lineto(xoff,yval[i]);
2.0400     xoff += DisW[n].MisC[2];
2.0401 }
2.0401 return;
2.0402 }
2.0403
2.0403 #####
2.0404 void DMA_Done(word Timeout)
2.0404
2.0405 // POLL DMA CONTROLLER TO ASSERTAIN DMA STATUS. THIS FUNCTION
2.0406 // WAS TAKEN FROM THE DSP DRIVER SOFTWARE AND OPTIMISED TO
2.0406 // GAIN A SPEED ADVANTAGE.
2.0407 {
2.0407     word pollstatus,timecount = 0;
2.0408     byte TCMask = 1 << DMA_Chan,DMask = DMA_Chan+4,State;
2.0409
2.0409     State = inportb(CardAddr);
2.0410     do
2.0411     {
2.0411         pollstatus = inportb(8);
2.0412     }while (((pollstatus & TCMask) == 0) && (++timecount <= Timeout));
2.0412     outportb(0xA,DMask);
2.0413     outportb(CardAddr,State&0x18);    //DISABLE DMA MODE
2.0414     if (timecount > Timeout)
2.0414     {
2.0415         printf("%d ",row);
2.0415         sound(100);
2.0416         delay(1);
2.0417         nosound();
2.0417     }
2.0418     return;
2.0418 }
2.0419
2.0420 #####
2.0420 int InTPoll56(word Timeout)
2.0421
2.0422 // POLL DMA CONTROLLER TO ASSERTAIN DMA STATUS. THIS FUNCTION
2.0422 // WAS TAKEN FROM THE DSP DRIVER SOFTWARE AND OPTIMISED TO
2.0423 // GAIN A SPEED ADVANTAGE.
2.0423 {
2.0424 #define DMAstatus 8
2.0425 #define DMAmaskS 0xA
2.0425

```

```

2.0426 word pollstatus,timecount = 0;
2.0426 byte TCMask = 1<<DMA_Chan,DMask = DMA_Chan+4;
2.0427
2.0428 do
2.0428 {
2.0429     pollstatus=inportb(DMAstatus);
2.0429 }while (((pollstatus & TCMask) == 0) && (++timecount <= Timeout));
2.0430 outportb(DMAmaskS,DMask);
2.0431 return !(timecount<Timeout);
2.0431 }
2.0432
2.0433 #####
2.0433 void InTDMA(byte icr,word DMApageno,word DMAoffset,word count,byte ICRmask)
2.0434
2.0434 // SETUP OF DMA CONTROLLER. THIS FUNCTION WAS TAKEN FROM THE
2.0435 // DSP DRIVER SOFTWARE AND OPTIMISED TO GAIN A SPEED ADVANTAGE.
2.0436 // (SEE IBM TECHNICAL REFERENCE MANUAL FOR DETAILS).
2.0436 {
2.0437     word PageOffs = 3,DMAcount,DMAaddr;
2.0437     byte ICR = icr,State;
2.0438
2.0439     State = inportb(CardAddr);
2.0439     outportb(0x80+PageOffs,DMApageno);
2.0440     outportb(0xB,DMA_Chan+4+(DMAdown<<5));
2.0440     outportb(0xA,DMA_Chan+4);
2.0441     outportb(0xC,0xFF);
2.0442     count--;
2.0442     DMAaddr = DMA_Chan<<1;
2.0443     DMAcount = DMAaddr+1;
2.0444     outportb(DMAaddr,DMAoffset&0xFF);
2.0444     outportb(DMAaddr,DMAoffset>>8);
2.0445     outportb(DMAcount,count&0xFF);
2.0445     outportb(DMAcount,count>>8);
2.0446     outportb(0xA,DMA_Chan);
2.0447     inportb(8);
2.0447     ICR |= ICRmask|0x80;
2.0448     ICR |= State&0x18; // PRESERVE FLAGS IN DSP
2.0448     if(ICRmask) outportb(CardAddr,ICR);
2.0449 }
2.0450
2.0450 #####
2.0451 void DMAIn56(byte icr,word segmentpart,word offsetpart,word count)
2.0451
2.0452 // SETUP OF DMA CONTROLLER. THIS FUNCTION WAS TAKEN FROM THE
2.0453 // DSP DRIVER SOFTWARE AND OPTIMISED TO GAIN A SPEED ADVANTAGE.
2.0453 // (SEE IBM TECHNICAL REFERENCE MANUAL FOR DETAILS).
2.0454 {
2.0454     _addrtype addr;
2.0455     word DMAoffset,DMApageno,DMAcount;
2.0456     byte ICRmask = 1;
2.0456
2.0457     addr.l = (dword)offsetpart+((dword)segmentpart<<4);
2.0458     DMApageno = addr.w.hi;
2.0458     DMAoffset = addr.w.lo;
2.0459
2.0459     if((DMAoffset+1)<count)
2.0460     {
2.0461         DMAcount = DMAoffset+1;

```

```

2.0461      InTDMA(icr,DMApageno,DMAoffset,DMAcount,ICRmask);
2.0462      ICRmask = 0;
2.0462      addr.l -= (dword) DMAcount;
2.0463      count -= DMAcount;
2.0464      DMApageno = addr.w.hi;
2.0464      DMAoffset = addr.w.lo;
2.0465      if(InTPoll56(10000))
2.0465      {
2.0466          printf("%d ",row);
2.0467          sound(100);
2.0467          delay(1);
2.0468          nosound();
2.0469      }
2.0469  }
2.0470  InTDMA(icr,DMApageno,DMAoffset,count,ICRmask);
2.0470 }
2.0471
2.0472 //#####
2.0472 void DMAfromDSP(long huge *Freq)
2.0473
2.0473 // NOTE :
2.0474 // 1.INTEL USES LITTLE ENDIAN ORDER. (ADDRESS OF WORD = LSB)
2.0475 //  DSP SENDS MSB FIRST.
2.0475 // 2.24 BIT TRANSFERS ARE DONE. AN ADDITIONAL BYTE IS INSERTED
2.0476 //  TO TRANSFORM THE VALUE INTO A LONG.
2.0476 //  INVESTIGATE MSB FOR SIGN :
2.0477 //  IF NEG INSERT $FF FOR 2's COMPLIMENT
2.0478 //  IF POS INSERT $00.
2.0478 // 3.TRANSFER PROCESS IS REVERSED !!!!!
2.0479 // => COMPENSATED FOR ON DSP CARD
2.0480 // 4.READ ISR (TIMING OF DMA) TO TEST IF HF2 IS SET
2.0480 // 5.HF0 CAN NOT BE PLACED HERE AS THE DSP IS TO QUICK BETWEEN
2.0481 //  SUCCESSIVE DMA TRANSFERS. (TIME BETWEEN CLEAR AND NEXT SET)
2.0481 // 6.VALUE WRITTEN TO ICR DEPENDS ON THE DESIRED BYTES PER TRANSFERED WORD
2.0482 //  ICR = 0x60      ;SINGLE BYTE ONLY
2.0483 //  ICR = 0x40      ;INTEGER TRANSFER
2.0483 //  ICR = 0x20      ;24 BIT WORDS
2.0484
2.0484 {
2.0485     long huge *ptr = &Freq[NR-1];
2.0486     byte *b1 = (byte *)&Freq[NR/4],*b2 = (byte *)&Freq[0];
2.0486     int i;
2.0487
2.0487     do HF2 = inportb(CardAddr+2);
2.0488     while((HF2&8) != 8);
2.0489
2.0489     DMAIn56(0x20,FP_SEG(ptr),(FP_OFF(ptr)+3),3*NR);
2.0490     DMA_Done(10000);
2.0491
2.0491     for(i=0;i<NR;i++)
2.0492     {
2.0492         *b2++=*b1++;
2.0493         *b2++=*b1++;
2.0494         *b2++=*b1;
2.0494         if((*b1&128)==128)
2.0495             *b2=0xFF;
2.0495         else
2.0496             *b2=0;

```

```

2.0497         b2++;b1++;
2.0497     }
2.0498     return;
2.0498 }
2.0499
2.0500 #####
2.0500 void C_DMAfromDSP(char *Freq)
2.0501
2.0501 // RECIEVE CHAR (8 BIT) VALUE FROM DSP.
2.0502 {
2.0503     char *ptr = &Freq[NR-1];
2.0503
2.0504     do HF2 = inportb(CardAddr+2);
2.0505     while((HF2&8) != 8);
2.0505
2.0506     DMAIn56(0x60,FP_SEG(ptr),FP_OFF(ptr),NR);
2.0506     DMA_Done(10000);
2.0507     return;
2.0508 }
2.0508
2.0509 #####
2.0509 void I_DMAfromDSP(int *Signal)
2.0510
2.0511 // RECIEVE INT (16 BIT) VALUE FROM DSP.
2.0511 {
2.0512     int *ptr = &Signal[NRaw-1];
2.0512
2.0513     do HF2 = inportb(CardAddr+2);
2.0514     while((HF2&8) != 8);
2.0514
2.0515     DMAIn56(0x40,FP_SEG(ptr),FP_OFF(ptr)+1,NRaw<<1);
2.0516     DMA_Done(10000);
2.0516     return;
2.0517 }
2.0517
2.0518 #####
2.0519 void DMAtoDSP(long Ch[])
2.0519
2.0520 // TRANSFER 24 BIT VALUES TO DSP. THE LONG ARRAY IS FIRST MADE
2.0520 // INTO A 3 BYTE ARRAY THROUGH REMOVING THE FOURTH BYTE. THE
2.0521 // MAXIMUM VALUE TRANSFERRED IS THEREFORE 2^23.
2.0522 {
2.0522     byte *b1 = (byte *)&Ch[0];*b2 = (byte *)&Ch[0];
2.0523     int i;
2.0523
2.0524     for(i = 0;i < NS; i++)
2.0525     {
2.0525         *b1++ = *b2++;
2.0526         *b1++ = *b2++;
2.0527         *b1++ = *b2++;
2.0527         b2++;
2.0528     }
2.0528     b1--;
2.0529
2.0530     DMAWrite56(CardAddr,FP_SEG(b1),FP_OFF(b1),(3*NS),DMA_Chan,3,DMAdown);
2.0530     DMAPoll56(CardAddr,1,1000);
2.0531     return;
2.0531 }

```

```

2.0532
2.0533 //#####
2.0533 void To_DSP(void)
2.0534
2.0534 // TRANSFER TWO ARRAYS TO THE DSP.
2.0535 // NB! THE MAXIMUM VALUE ALLOWED IS 2^23. (SEE DMAtoDSP())
2.0536 //
2.0536 // EX : TO SYNTHESIZE TWO QUADRATURE SIGNALS
2.0537 // Ch1 = Bfsin(Wft)+Brsin(Wrt)
2.0537 // Ch2 = Bfsin(Wft-pi/2)+Brsin(Wrt+pi/2)
2.0538 {
2.0539     int i;
2.0539     long Ch1[NS],Ch2[NS];
2.0540
2.0541     for(i = 0; i < NS; i++)
2.0541     {
2.0542         Ch1[i] = (long)(1024e2*cos(2*3.1416*4000*i/Fs))
2.0542             +(long)(512e2*cos(2*3.1416*1000*i/Fs));
2.0543
2.0544         Ch2[i] = (long)(1024e2*cos(2*3.1416*4000*i/Fs-1.5708))
2.0544             +(long)(512e2*cos(2*3.1416*1000*i/Fs+1.5708));
2.0545     }
2.0545
2.0546     DMAtoDSP(Ch1);
2.0547     DMAtoDSP(Ch2);
2.0547     return;
2.0548 }
2.0548
2.0549 //#####
2.0550 void Time(void)
2.0550
2.0551 // CALCULATE TIME DURATION OF OPERATIONS. USED IN DEBUGGING.
2.0552 // TO USE THIS OPTION INSERT ANOTHER OPTION IN THE MAIN MENU
2.0552 // WHICH POINTS TO THIS FUNCTION.
2.0553 {
2.0553     int i;
2.0554
2.0555     TimerStart();
2.0555     outportb(CardAddr+ICReg,inportb(CardAddr+ICReg)/8);
2.0556     for(i = 1; i <= 200; i++)
2.0556     {
2.0557         I_DMAfromDSP(Xmem[row]);
2.0558         I_DMAfromDSP(Ymem[row]);
2.0558         C_DMAfromDSP(RFreq[row]);
2.0559         C_DMAfromDSP(FFreq[row]);
2.0559         Spectrogram(row);
2.0560         (row == (MaxRow-1)) ? (row=0) : (row++);
2.0561     }
2.0561     outportb(CardAddr+ICReg,inportb(CardAddr+ICReg)&0xF7);
2.0562
2.0563     ElapsT = TimeElapsed();
2.0563     printf("%ld",ElapsT);
2.0564     getch();
2.0564     return;
2.0565 }
2.0566
2.0566 //#####
2.0567 void MeanF(void)

```

```

2.0567 // IN CALCULATING THE MEAN FREQUENCY WE SEARCH FOR THAT FREQUENCY SUCH
2.0568 // THAT THE ENERGY CONTENT TO THE LEFT IS EQUAL TO THE ENERGY CONTENT TO
2.0569 // THE RIGHT.
2.0569 {
2.0570 #define D DisW
2.0570
2.0571 int i,j,xoff,yoff,MaxFreq=5000;
2.0572 float FBin = MaxFreq/NR,EF,ER,Esum;
2.0572 char *TF,*TR;
2.0573 int yval[MaxRow];
2.0574
2.0575 if(Bvel_flag != 1) return;
2.0575
2.0576 for(row = 0;row < MaxRow; row++)
2.0577 {
2.0577     EF = 0;
2.0578     ER = 0;
2.0578     TF = FFreq[row];
2.0579     TR = RFreq[row];
2.0580     for(j = 0;j < NR; j++)
2.0580     {
2.0581         EF += TF[j];
2.0581         ER += TR[j];
2.0582     }
2.0583     Esum = 0;
2.0583     j = 0;
2.0584     do
2.0584     {
2.0585         Esum += TF[j];
2.0586         j++;
2.0586     }while(Esum < (EF/2));
2.0587     BVel[row] = (int)((EF*FBin*j)/(EF+ER));
2.0588
2.0588     Esum = 0;
2.0589     j = 0;
2.0589     do
2.0590     {
2.0591         Esum += TR[NR-1-j];
2.0591         j++; // NEG FREQ ARRAY REVERSED!!
2.0592     }while(Esum < (ER/2));
2.0592     BVel[row] -= (int)((ER*FBin*j)/(EF+ER));
2.0593 }
2.0594
2.0594 yoff = (int)((DisW[0].Act[3]+DisW[0].Act[1])/2);
2.0595 xoff = DisW[0].Act[0];
2.0595
2.0596 for (i=0;i<MaxRow;i++)
2.0597     yval[i]=yoff-(int)((float)(BVel[i])/MaxFreq)*(DisW[0].Act[3]-yoff);
2.0597
2.0598 setcolor(15);
2.0599 moveto(xoff,yval[0]);
2.0599 xoff += 2;
2.0600 for (i = 1;i < MaxRow; i++)
2.0600 {
2.0601     lineto(xoff,yval[i]);
2.0602     xoff += 2;
2.0602 }

```

```

2.0603
2.0603 while(!kbhit()){
2.0604
2.0605 // REDRAW THE SPECTROGRAM TO OVERWRITE THE VELOCITY DISPLAY.
2.0605 GrWin(D[0].Act[0],D[0].Act[1],D[0].Act[2],D[0].Act[3],D[0].fcol,D[0].bcol,1);
2.0606 Spec_Prep(0);
2.0606 DisW[0].MisC[0]=DisW[0].Act[0];
2.0607 for(row=0;row<MaxRow;row++)
2.0608     Spectrogram(row);
2.0608 row = 0;
2.0609
2.0610 #undef D DisW
2.0610 return;
2.0611 }
2.0611
2.0612 #####
2.0613 void WaveS(void)
2.0613
2.0614 // SAVE THE VELOCITY WAVEFORM AFTER CALCULATION. A .VEL
2.0614 // FILE EXTENSION IS USED.
2.0615 {
2.0616     int i;
2.0616     FILE *stream;
2.0617     char Path_Name[80] = "",path[50] = "",fext[4] = "",
2.0617
2.0618     sprintf(Path_Name,"");
2.0619     sprintf(path,"c:\\bc\\files\\patdata\\");
2.0619     sprintf(fext,".vel");
2.0620     strcat(Path_Name,path);
2.0621     strcat(Path_Name,FileName);
2.0621     strcat(Path_Name,fext);
2.0622
2.0622     if((stream=fopen(Path_Name,"wt"))==NULL)
2.0623     {
2.0624         sound(100);
2.0624         delay(500);
2.0625         nosound();
2.0625         return;
2.0626     }
2.0627
2.0627     for(i=0;i<MaxRow;i++)
2.0628         fprintf(stream,"%d\n",BVel[i]);
2.0628
2.0629
2.0630     fclose(stream);
2.0630     SOUND();
2.0631     return;
2.0631 }
2.0632
2.0633 #####
2.0633 int gprintf(int *xloc,int *yloc,char *fmt, ... )
2.0634
2.0635 // OUTPUT OF TEXT IN GRAPHICS MODE.
2.0635 {
2.0636     va_list argptr;
2.0636     char str[140];
2.0637     int cnt;
2.0638

```

```

2.0638 va_start(argptr,fmt);
2.0639 cnt = vsprintf(str,fmt,argptr);
2.0639 outtextxy(*xloc,*yloc,str);
2.0640 *yloc += textheight("H")+2;
2.0641 va_end(argptr);
2.0641
2.0642 return(cnt);
2.0642 }
2.0643
2.0644 #####
2.0644 int Boundary_Select(int start)
2.0645
2.0646 // SELECT A VELOCITY SEGMENT FOR ANALYSIS.
2.0646 // EXAMPLE OF USAGE :
2.0647 // srow=Boundary_Select(10);
2.0647 // frow=Boundary_Select(srow+10);
2.0648
2.0649 {
2.0649 int i,x = DisW[0].MisC[0],xoff = start<<1,Rnum = start;
2.0650 int y1 = DisW[0].Act[1],y2 = DisW[0].Act[3];
2.0650 char ch;
2.0651
2.0652 do
2.0652 {
2.0653     for(i = 0;i < NR<<1; i += 2)
2.0653         putpixel(xoff+x+1,y1+i,15);putpixel(xoff+x+1,y2-i,15);
2.0654     ch = getch();
2.0655
2.0655     switch(ch)
2.0656     {
2.0657         case 0x4B:{
2.0657             for(i = 0;i < NR<<1;i += 2)
2.0658                 putpixel(xoff+x+1,y1+i,0);putpixel(xoff+x+1,y2-i,0);
2.0658             Rnum--;
2.0659             xoff -= 2;
2.0660             };break;
2.0660
2.0661         case 0x4D:{
2.0661             for(i = 0;i < NR<<1;i += 2)
2.0662                 putpixel(xoff+x+1,y1+i,0);putpixel(xoff+x+1,y2-i,0);
2.0663             Rnum++;
2.0663             xoff += 2;
2.0664             };break;
2.0664     }
2.0665 }while (ch != 0xD);
2.0666 return(Rnum);
2.0666 }
2.0667
2.0668 #####
2.0668 void LookW(void)
2.0669
2.0669 // INSPECT THE CAPTURED TIME WAVEFORM. ONE USE IS TO ENSURE
2.0670 // THAT THE AFFORE MENTIONED WAS CAPTURED WITHOUT CLIPPING.
2.0671 // THE OPTION EXIST TO STEP THROUGH (PAGE UP/DOWN) THE ARRAYS
2.0671 // AND SO VIEW ALL EXISTING DATA.
2.0672 {
2.0672 #define D DisW
2.0673     int test = 1,r = row,i;

```

```

2.0674 char kbchar;
2.0674
2.0675 {
2.0676     if ((D[i].Mapptr = GrSave(D[i].L,D[i].T,D[i].R,D[i].B)) != NULL)
2.0677     {
2.0677         GrWin(D[i].L,D[i].T,D[i].R,D[i].B,D[i].fcol,D[i].bcol,1);
2.0678         XY_Prep(i);
2.0678     }else return;
2.0679 }
2.0680
2.0680 I_signal_view(Xmem[r],3,NRow);
2.0681 I_signal_view(Ymem[r],4,NRow);
2.0682 C_signal_view(FFreq[r],5,NR);
2.0682 C_signal_view(RFreq[r],6,NR);
2.0683
2.0683 while(!kbhit()){}
2.0684 while(test)
2.0685 {
2.0685
2.0686     if (kbhit())
2.0686     {
2.0687         kbchar=getch();
2.0688         switch(kbchar)
2.0688         {
2.0689
2.0689             case 0x49 :{
2.0690                 (r==(MaxRow-1))?r=0:(r++);
2.0691                 I_signal_view(Xmem[r],3,NRow);
2.0691                 I_signal_view(Ymem[r],4,NRow);
2.0692                 C_signal_view(FFreq[r],5,NR);
2.0693                 C_signal_view(RFreq[r],6,NR);
2.0693             }break;
2.0694
2.0694             case 0x51 :{
2.0695                 (r==0)?r=MaxRow-1:(r--);
2.0696                 I_signal_view(Xmem[r],3,NRow);
2.0696                 I_signal_view(Ymem[r],4,NRow);
2.0697                 C_signal_view(FFreq[r],5,NR);
2.0697                 C_signal_view(RFreq[r],6,NR);
2.0698             }break;
2.0699
2.0699             case 0x1B : test=0;
2.0700             break;
2.0700         }
2.0701     }
2.0702 }
2.0703
2.0704 for(i = 3; i <= 6; i++)
2.0704     GrRestore(D[i].L,D[i].T,D[i].Mapptr);
2.0705
2.0705 #undef D DisW
2.0706 return;
2.0707 }
2.0707
2.0708 #####
2.0708 void Go(void)
2.0709

```

```

2.0710 // READ DATA AS CAPTURED BY DSP56001. FOR DEBUGGING PURPOSES
2.0710 // EXAMPLES ARE SHOWN OF WHERE TO PLACE RELEVANT FUNCTIONS.
2.0711 // THE DMA-DISPLAY SEQUENCE IS REPEATED UNTIL INTERRUPTED BY
2.0711 // THE USER BY EITHER THE ESC (EXIT) OR SPACEBAR (INSPECTION)
2.0712 // KEYSTROKES. NOTE THE PLACEMENT OF THE HOST2DSP FLAG WHICH
2.0713 // IS USED TO COORDINATE THE DMA TRANSFERS.
2.0713 {
2.0714     char choice;
2.0714     int test = 1,i;
2.0715
2.0716     //ValWindow();
2.0716     //To_DSP();
2.0717
2.0718     TimerStart();
2.0718
2.0719     // SET HF0 IN ICR
2.0719     outportb(CardAddr+ICReg,inportb(CardAddr+ICReg)&8);
2.0720     while(test)
2.0721     {
2.0721         I_DMAfromDSP(Xmem[row]);
2.0722         I_DMAfromDSP(Ymem[row]);
2.0722         C_DMAfromDSP(RFreq[row]);
2.0723         C_DMAfromDSP(FFreq[row]);
2.0724         Spectrogram(row);
2.0724
2.0725         // I_signal_view(Xmem[row],1,NRow); // NOTE TIMING OF DSP INT !!
2.0725         // I_signal_view(Ymem[row],2,NRow);
2.0726
2.0727         if (kbhit())
2.0727         {
2.0728             // RESET HF0 IN ICR
2.0729             outportb(CardAddr+ICReg,inportb(CardAddr+ICReg)&0xF7);
2.0729             choice=toupper(getch());
2.0730
2.0730             switch(choice)
2.0731             {
2.0732                 case 0x20 : ReDraw();
2.0732                     Bvel_flag = 1;
2.0733                     return;
2.0733                 case 0x1B : test = 0;
2.0734             }
2.0734             break;
2.0735         }
2.0735     }
2.0736     row = ++row % MaxRow;
2.0736 }
2.0737     return;
2.0738 }
2.0738 //#####

```

```

3.0000 //*****
3.0001 //      ACQUISITION AND ANALYSIS OF
3.0002 //      LOWER LIMB FLOW WAVEFORMS
3.0003 //*****
3.0003
3.0004 // Copyright(C) L.SMITH
3.0005 // Department of Biomedical Engineering
3.0005 // University of Cape Town
3.0006 // Observatory
3.0006 // Cape Town
3.0007 // New South Africa
3.0008 // (021) 471250 x 235
3.0008
3.0009 #define DWNum 7           // NUMBER OF DISPLAY WINDOWS
3.0010 #define MNum 4          // NUMBER OF SUBMENUS CALLED FROM MAIN MENU
3.0010
3.0011 #define DMA_Chan 1
3.0011 #define Fs 40e3         // SAMPLING FREQUENCY
3.0012 #define NR 64          // NUMBER OF VALUES RECEIVED FROM DSP56001
3.0013 #define NRaw 128       // NUMBER OF RAW DATA VALUES RECEIVED FROM DSP56001
3.0013 #define NS 256         // NUMBER SEND TO DSP
3.0014 #define Val_Win 200    // NUMBER OF SPECTRAL LINES PER DISPLAY WINDOW
3.0014 #define VLine_DWn 400 // NUMBER OF VERTICAL LINES PER DISPLAY WINDOW
3.0015
3.0016 #define ColLevels 16    // SELECT AS A EVEN NUMBER !!
3.0016 #define MaxCollnt 63   // MAXIMUM COLOUR INTENSITY
3.0017 #define MinCollnt 23   // MINIMUM COLOUR INTENSITY
3.0017 #define PalOff 16     // START OF GRAY SCALES
3.0018
3.0019 #include <dos.h>
3.0019 #include <ctype.h>
3.0020 #include <stdio.h>
3.0021 #include <stdlib.h>
3.0021 #include <conio.h>
3.0022 #include <graphics.h>
3.0022 #include <math.h>
3.0023 #include <mem.h>
3.0024 #include <stdarg.h>
3.0024 #include <alloc.h>
3.0025 #include <string.h>
3.0025
3.0026 //***** STRUCTURES *****
3.0027 struct MenuWindow{
3.0027   char    *MenuList;
3.0028   int     L,T,R,B;
3.0028   int     fcol,bcol;           // FORE- AND BACKGROUND
3.0029   int     type_op;           // HOW OPENED
3.0030   char    *KeyList;         // HOTKEYS
3.0030   void    (*ProcPointer[15])(); // POINTERS TO FUNCTIONS CALLED
3.0031   int     ArgList[15];
3.0032   char    far *Mapptr;      // BACKGROUND SAVE
3.0032 };
3.0033
3.0033 struct DisplayWindow{
3.0034   int     status;           // 0:CLOSED,1:ACTIVE ,2:DE-ACTIVATED
3.0035   char    title[50];
3.0035   char    xlabel[50];
3.0036   char    ylabel[50];

```

```

3.0036 int      fool,bcol,Icol,Dcol;          // ...,PAINTCOLOR, LETTERCOLOR
3.0037 int      L,T,R,B;
3.0038 int      Act[4];
3.0038 int      MisC[4];
3.0039 char     far *Mapptr;
3.0039 };
3.0040
3.0041 //***** PROTOTYPE DECLARATIONS *****
3.0041 void GrRestore(int l,int t,char far *Bitmapptr);
3.0042 char far *GrSave(int l,int t,int r,int b);
3.0043 void InitPalette(void);
3.0043 void To_DSP();
3.0044 void ValWindow(void);
3.0044 void ScaleVal(void);
3.0045 void ReadFile(void);
3.0046 void Get_FileN(int type_op);
3.0046 void WriteFile(void);
3.0047 void Spectrogram(int row);
3.0047 void DMAtoDSP(long Ch[]);
3.0048 void I_DMAfromDSP(int *Signal);
3.0049 void C_DMAfromDSP(char *Freq);
3.0049 void DMAfromDSP(long huge *Freq);
3.0050 void DMAIn56(byte icr,word segmentpart,word offsetpart,word count);
3.0050 void InTDMA(byte icr,word DMApageno,word DMAoffset,word count,byte ICRmask);
3.0051 int InTPoll56(word Timeout);
3.0052 void DMA_Done(word Timeout);
3.0052 void C_signal_view(char *signal,int n,int nr);
3.0053 void I_signal_view(int *signal,int n,int nr);
3.0053 void signal_view(long huge *signal,int n,int nr);
3.0054 int Boundary_Select(int start);
3.0055 void ReDraw(void);
3.0055 void WaveS(void);
3.0056 int gprintf(int *xloc,int *yloc,char *fmt,...);
3.0057 void Go(void);
3.0057 void MeanF(void);
3.0058 void LookW(void);
3.0058 void GrWin(int l,int t,int r,int b,int fool,int bcol,int type_op);
3.0059 void Init_Graph(void);
3.0060 void Test(void);
3.0060 void Listprint(int l,int t,int fcol,char *list,char *PossKey);
3.0061 void XY_Prep(int n);
3.0061 void Spec_Prep(int n);
3.0062 void DefaultWin(void);
3.0063 void Menu_Man(int Num);
3.0063 unsigned long int TimeElapsed(void);
3.0064 void TimerStart(void);
3.0064 void Time(void);
3.0065 void BackWin(int col);
3.0066 void Quit(void);
3.0066
3.0067 //*****

```

MATLAB SOURCE CODE LISTING (4.0000-4.0807)

```

4.0000 %/* *****
4.0001 %/* *
4.0002 %/* *           CW DOPPLER ULTRASOUND
4.0003 %/* *
4.0004 %/* *****
4.0005
4.0006 % Waveform analysis (See respective m files for help regarding parameters)
4.0007 % NB !! 386 MATLAB required.
4.0008 %
4.0009 % Usage :
4.0010 % **           1 : load c:\bc\files\patdata\name.raw
4.0011 % **           2 : [X,Y] = lrd(name);
4.0012 %           3 : [F,R] = arspec(X,Y,15,1);
4.0013 %           4 : B = bvel(F,R,1);
4.0014 %           5 : Bfil = vfil(B,10,5,0);
4.0015 %           6 : Bseg = ss3(Bfil);
4.0016 %                   (or ..ss,..ss1,..ss1)
4.0017 %           7 : global Data
4.0018 % ^           8 : var = disfit(Bseg);
4.0019 %           9 : disnel(Bseg,var,'name');
4.0020 %           10 : If so desired save parameters using "save"
4.0021 %
4.0022 % **           If so desired the periodogram data may be analyzed as an
4.0023 %           alternative. Replace 1-3 with :
4.0024 %           1 : load c:\bc\files\patdata\name.spc
4.0025 %           2 : [F,R] = lsd(name);
4.0026 % ^           For Laplacian analysis use
4.0027 %                   par = lapfit(Bseg);
4.0028 %                   To calculate and view the result use
4.0029 %                   par = dislap(Bseg);
4.0030
4.0031 %*****
4.0032 % Calculate biased autocorrelation of a given function
4.0033 % R = acorr(fun,k,norm)
4.0034 % k = number of correlation points to do
4.0035 % if norm = 1 R is normalized so that R(1) = 1;
4.0036
4.0037 function [R] = acorr(fun,k,norm)
4.0038 N = max(size(fun));
4.0039 m = 1;
4.0040 for i = 1:k
4.0041     R(m) = (sum(fun(1:N-m+1).*fun(m:N)))/N;
4.0042     m = m+1;
4.0043 end
4.0044 if norm == 1
4.0045     R = R./max(R);
4.0046 end
4.0047
4.0048 %*****
4.0049 % Perform AR spectral estimation on given raw data.
4.0050 % [F,R] = arspec(xmem,ymem,nk,gr)
4.0051 %           xmem = raw data sampled into X memory
4.0052 %           ymem = raw data sampled into Y memory
4.0053 %           nk = AR filter order
4.0054 %           gr = 0 no graph
4.0055 %           = 1 display graph
4.0056 % The following is done :
4.0057 %           1.           Frequency domain processing

```

```

4.0058 %      2.      Autocorrelation (Rx)
4.0059 %      3.      Extrapolation of Rx
4.0060 %      4.      Spectral calculation
4.0061
4.0062 function [F,R] = arspec(xmem,ymem,nk,gr)
4.0063 [N,M] = size(xmem);
4.0064 fs = 40e3; t = 0:N-1; fc = 5000; N_FFT = 512;
4.0065 csin = sin(2*pi*t*fc/fs);
4.0066 ccos = cos(2*pi*t*fc/fs);
4.0067
4.0068 for I = 1:M
4.0069     clc
4.0070     I
4.0071     xc = csin.*xmem(:,I);
4.0072     yc = ccos.*ymem(:,I);
4.0073     Ssum = xc+yc;
4.0074     Rx = acorr(Ssum,nk+1,0);
4.0075     [k,a,error,alpha] = durbin(Rx,nk);
4.0076
4.0077     m = 0;
4.0078     for n = nk+2:N_FFT/2
4.0079         Rx(n) = -sum(a(2:nk+1).*Rx(m+nk+1:-1:2+m));
4.0080         m = m+1;
4.0081     end
4.0082     Rx = [Rx(N_FFT/2:-1:2) Rx];
4.0083     Fsum = abs(fft(Rx,N_FFT));
4.0084     R(:,I) = Fsum(((fc/fs)*N_FFT):-1:1);
4.0085     F(:,I) = Fsum(((fc/fs)*N_FFT)+1:((fc/fs)*2*N_FFT));
4.0086 end
4.0087 if (gr == 1)
4.0088     f3d(F,R,max(size(F)));
4.0089 end
4.0090
4.0091 %*****
4.0092 % Calculate mean blood velocity given forward and reverse flow.
4.0093 % B = bvel(F,R,gr)
4.0094 % F = Forward Frequencies
4.0095 % R = Reverse Frequencies
4.0096 % gr = 0 no graph
4.0097 %   = 1 display graph
4.0098
4.0099 function [B] = bvel(F,R,gr)
4.0100 clg;
4.0101 FBin = 5000/64;
4.0102 freqs = (FBin:FBin:5000);
4.0103 for i = 1:max(size(F))
4.0104     f = sum((freqs.*F(:,i)));
4.0105     r = sum((freqs.*R(:,i)));
4.0106     B(i) = f/(sum(F(:,i))+sum(R(:,i)))-r/(sum(F(:,i))+sum(R(:,i)));
4.0107 end
4.0108 if (gr == 1)
4.0109     plot(B),grid,title('Mean Frequency'),xlabel('Frames (200 = 3sec)'),...
4.0110     ylabel('Frequency (Hz)');
4.0111 end
4.0112
4.0113 %*****
4.0114 % Fitting a function to a set of data using the Nelder-Mead
4.0115 % simplex algorithm for minimizing a function of several

```

```

4.0116 % variables.
4.0117 % NB! FIRST USE : global Data
4.0118 % var = a1,a2,a3,b1,b2 (SEE DIFFERENCE EQU)
4.0119 % var = disfit(bvel)
4.0120
4.0121 function [var] = disfit(bvel)
4.0122 Data = [];
4.0123 Data(1,:) = 1:max(size(bvel));      % Time increments
4.0124 Data(2,:) = bvel;
4.0125 Data = Data';
4.0126 % var = a1,a2,a3,b1,b2 (SEE DIFFERENCE EQU)
4.0127 % var = [-2.8 2.7 -0.9 0.2 -0.2]';
4.0128 var = [lap2dis(bvel)]';
4.0129 var = fmins('disfun',var,1e-5);
4.0130
4.0131 %*****
4.0132 % DISFUN is used by DISFIT. Disfun(var) returns the error
4.0133 % between the data and the values computed by the current
4.0134 % function of var. DISFIT assumes the following function :
4.0135 %
4.0136 %  $y(n) = -a1*y(n-1) - a2*y(n-2) - a3*y(n-3) +$ 
4.0137 %  $b1*u(n-1) + b2*u(n-2)$ 
4.0138 %
4.0139 % where  $u(n)$  = unit impulse function
4.0140 % i.e.  $u(1) = 1$ 
4.0141 %  $= 0 \quad n > 1$ 
4.0142
4.0143 function f = disfun(var)
4.0144 a1 = var(1);a2 = var(2);a3 = var(3);b1 = var(4);b2 = var(5);
4.0145 t = Data(:,1);
4.0146 y = Data(:,2);
4.0147
4.0148 yp(1) = 0;
4.0149 yp(2) = b1;
4.0150 yp(3) = -a1*yp(2)+b2;
4.0151 for n = 4:max(size(y))
4.0152   yp(n) = -a1*yp(n-1)-a2*yp(n-2)-a3*yp(n-3);
4.0153 end
4.0154 yp = yp(:);
4.0155 f = sum((yp-y).^2);
4.0156
4.0157 % Statements to plot progress of fitting:
4.0158 %c1g;
4.0159 %plot(t,y,t,yp)
4.0160 %xt = max(t)/4.5;
4.0161 %yt = max(y)/1.5;
4.0162 %text(xt,1.1*yt,['variable = ' num2str(var(1)) ' ' num2str(var(2)) ...
4.0163 % ' num2str(var(3)) ' ' num2str(var(4)) ' ' num2str(var(5))]);
4.0164 %text(xt,1.0*yt,['err norm = ' num2str(f)])
4.0165
4.0166 %*****
4.0167 % Curve fitting procedure to calculate model parameters of velocity
4.0168 % segment. Laplacian fitting done in the frequency domain. A display
4.0169 % of the parameters calculated is given.
4.0170 % par = displ(segment);
4.0171
4.0172 function [par] = displ(bseg)
4.0173

```

```

4.0174 % points = no of FFT points, fft2 is the two dim. (real+imag) fft
4.0175 echo off
4.0176 clg;
4.0177 points = 512;
4.0178 fs = 66;
4.0179 Bpoints = max(size(bseg));
4.0180 T = (1:Bpoints)/fs;
4.0181 A = 0;B = 0;G = 0;LTD = 0;
4.0182
4.0183 FREQ = fft2(bseg,1,points)/fs;
4.0184 MAG = (sqrt(real(FREQ).*real(FREQ)+imag(FREQ).*imag(FREQ)));
4.0185 MAG = MAG/MAG(1);
4.0186
4.0187 % Note that the the index of the FREQ and MAG arrays must be multiplied by;
4.0188 % 2*pi * (index=fft bin no)*fs/points to get radian freq;
4.0189 % No is the maximum fft bin no of interest, this corresponds to
4.0190 % a frequency of 2*pi*No*fs/points;
4.0191 No = 50;
4.0192 w1 = 0:1:No-1;
4.0193 % Note that the actual frequency (radians) is 2*pi*(fft bin no)*fs/points;
4.0194 w = 2*pi*w1*fs/points;
4.0195 UNIT = ones(1:No);
4.0196
4.0197 % Perform weighted least squares fit in the frequency domain
4.0198 % No of iterations
4.0199 for ITERATION = 0:5
4.0200     if ITERATION == 0
4.0201         Ek = ones(1:No);
4.0202     else
4.0203         for j = 1:No
4.0204             i = 2*pi*j*fs/points;
4.0205             W3 = [i^3 i^2 i 1];
4.0206             Ek(j) = 1/(P*W3)^2;
4.0207         end;
4.0208     end;
4.0209
4.0210 L0 = UNIT*Ek';
4.0211 L2 = (Ek.*w)*w';
4.0212
4.0213 T1 = (w.*Ek)*(imag(FREQ(1:No)));
4.0214 T3 = (w.*w.*w.*Ek)*(imag(FREQ(1:No)));
4.0215
4.0216 S0 = (UNIT.*Ek)*(real(FREQ(1:No)));
4.0217 S2 = (w.*w.*Ek)*(real(FREQ(1:No)));
4.0218 S4 = (w.*w.*w.*w.*Ek)*(real(FREQ(1:No)));
4.0219
4.0220 U2 = (w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0221 U4 = (w.*w.*w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0222 U6 = (w.*w.*w.*w.*w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0223
4.0224 M = [L0 0 T1 S2 -T3;0 L2 -S2 T3 S4; T1 -S2 U2 0 -U4;S2 T3 0 U4 0;T3 -S4 U4 0 -U6];
4.0225 C = [S0 T1 0 U2 0]';
4.0226 SOL = inv(M)*C;
4.0227
4.0228 P = [SOL(5) SOL(4) SOL(3) 1];
4.0229 poles = roots(P);
4.0230 Z = [SOL(2) SOL(1)];
4.0231 zeros = roots(Z);

```

```

4.0232
4.0233 Error = 0;
4.0234 for j = 0:No-1
4.0235     i = j*2*pi*fs/points;
4.0236     W1 = [i 1];
4.0237     W3 = [i^3 i^2 i 1];
4.0238     Error = Error+(MAG(j+1)-(Z*W1)/(P*W3))^2;
4.0239 end;
4.0240 Error = sqrt(Error)/No;
4.0241
4.0242 Best = imag(poles(1));
4.0243 if Best == 0,
4.0244     Gest = -poles(1);
4.0245     Aest = -real(poles(2));
4.0246     Best = imag(poles(2));
4.0247 else
4.0248     Aest = -real(poles(1));
4.0249     Gest = -poles(3);
4.0250 end;
4.0251
4.0252 end;
4.0253
4.0254 H = (Gest*Aest*Aest+Gest*Best*Best)/((Gest-Aest)*(Gest-Aest)+Best*Best);
4.0255 FLOWest = H*( exp(-Gest.*T) + exp(-Aest.*T).*(((Gest-Aest)/Best).*sin(Best.*T)-cos(Best.*T) ));
4.0256 FREQest = fft2(FLOWest,1,points)/fs;
4.0257 MAGest = (sqrt(real(FREQest).*real(FREQest)+imag(FREQest).*imag(FREQest)));
4.0258
4.0259 OUT = [Aest;Best;Gest];
4.0260 LTDest = cos(atan(Best/Aest));
4.0261 LTEest = sqrt(Aest^2+Best^2);
4.0262 LTRest = Gest;
4.0263 OUT2 = [LTDest;LTEest;LTRest];
4.0264 MSE = sum((FLOWest/max(FLOWest)-bseg/max(bseg)).^2);
4.0265 par = OUT2;
4.0266
4.0267 subplot(221);
4.0268 plot(T,bseg/max(bseg),T,FLOWest/max(FLOWest)),grid;
4.0269 title('Impulse response');
4.0270 subplot(222);
4.0271 plot(w,MAG(1:No)/max(MAG),w,MAGest(1:No)/max(MAGest)),grid;
4.0272 title('Frequency transform');
4.0273
4.0274 subplot(224),title('S-Domain'),s_zpplot([[0 0 0]' poles'],'c5'),grid;
4.0275
4.0276 xe = 0.1;xr = 0.2;xd = 0.3;xp = 0.4;
4.0277 text(xe,0.45,'LTE','sc');
4.0278 text(xr,0.45,'LTD','sc');
4.0279 text(xd,0.45,'LTR','sc');
4.0280 text(xp,0.45,['PI ' '( num2str(abs((max(bseg)-min(bseg))/mean(bseg))) )'],'sc');
4.0281
4.0282 text(0.0,0.415,'Laplace','sc');
4.0283 text(xe,0.415,num2str(LTEest),'sc');
4.0284 text(xr,0.415,num2str(LTDest),'sc');
4.0285 text(xd,0.415,num2str(LTRest),'sc');
4.0286 text(xp,0.415,num2str(abs((max(FLOWest)-min(FLOWest))/mean(FLOWest))),'sc');
4.0287
4.0288 text(0.0,0.375,'S-Dom','sc');
4.0289 text(xe,0.375,'Zeros','sc');

```

```

4.0290 text(xe,0.34,num_str(zeros(1)),'sc');
4.0291
4.0292 text(xd,0.375,'Poles','sc');
4.0293 text(xd,0.34,num_str(ones(1)),'sc');
4.0294 text(xd,0.315,num_str(ones(2)),'sc');
4.0295 text(xd,0.29,num_str(ones(3)),'sc');
4.0296
4.0297 text(0.0,0.25,'MSE','sc');
4.0298 text(xe,0.25,num_str(MSE),'sc');
4.0299
4.0300 %echo on
4.0301 % The waveform parameter estimates are:
4.0302 % A B G (Real,imaginary for complex poles, and real pole position resp.)
4.0303 %OUT
4.0304 % LTD,LTE,LTR are
4.0305 %OUT2
4.0306 %Error
4.0307 %echo off
4.0308
4.0309 %*****
4.0310 % Display variables analyzed.
4.0311 % varc(Bseg,var,string)
4.0312 % var = a1,a2,a3,b1,b2 (SEE DIFFERENCE EQU)
4.0313
4.0314 function disnel(s,var,str)
4.0315 clg;fs = 66;
4.0316 u = [1 zeros(1,max(size(s)-1))]; % system input
4.0317
4.0318 % var = a1,a2,a3,b1,b2 (SEE DIFFERENCE EQU)
4.0319 a = [1 var(1:3)'];
4.0320 b = [0 var(4:5)'];
4.0321 T = filter(b,a,u);
4.0322
4.0323 [Zz,Pz,k] = tf2zp(b,a);
4.0324 Ps = fs.*log(Pz);
4.0325 Zs = fs.*log(Zz);
4.0326 Ps = Ps(:); % ensure column vector
4.0327 if imag(Ps(1)) == -imag(Ps(2))
4.0328     sys_2 = poly(Ps(1:2));
4.0329     LTR = -real(Ps(3));
4.0330 else
4.0331     if imag(Ps(2)) == -imag(Ps(3))
4.0332         sys_2 = poly(Ps(2:3));
4.0333         LTR = -real(Ps(1));
4.0334     else
4.0335         str = 'error';
4.0336     end
4.0337 end
4.0338 LTE = sqrt(sys_2(3));
4.0339 LTD = sys_2(2)/(2*LTE);
4.0340 if LTD > 1
4.0341     LTD = 1;
4.0342 end
4.0343 PI = abs((max(T)-min(T))/mean(T));
4.0344 MSE = disfun(var);
4.0345
4.0346 subplot(221),title('Impulse Response'),plot([s' T']),grid;
4.0347 subplot(222),title('Z-Domain'),z_zpplot([0.*Zz Pz],'c5'),grid;

```

```

4.0348 subplot(224),title('S-Domain'),s_zpplot([0.*Zs Ps],'c5'),grid;
4.0349
4.0350 xe = 0.1;xr = 0.2;xd = 0.3;xp = 0.4;
4.0351 text(xe,0.45,'LTE','sc');
4.0352 text(xr,0.45,'LTD','sc');
4.0353 text(xd,0.45,'LTR','sc');
4.0354 text(xp,0.45,['PI ' 'C ...
4.0355     num2str(abs((max(s)-min(s))/mean(s)))'],'sc');
4.0356
4.0357 text(0.0,0.415,'Nelder','sc');
4.0358 text(xe,0.415,num2str(LTE),'sc');
4.0359 text(xr,0.415,num2str(LTD),'sc');
4.0360 text(xd,0.415,num2str(LTR),'sc');
4.0361 text(xp,0.415,num2str(PI),'sc');
4.0362
4.0363 text(0.0,0.375,'Z-Dom','sc');
4.0364 text(xe,0.375,'Zeros','sc');
4.0365 text(xe,0.34,num_str(Zz(1)),'sc');
4.0366 text(xe,0.315,num_str(Zz(2)),'sc');
4.0367 text(xe,0.29,num_str(Zz(3)),'sc');
4.0368
4.0369 text(xd,0.375,'Poles','sc');
4.0370 text(xd,0.34,num_str(Pz(1)),'sc');
4.0371 text(xd,0.315,num_str(Pz(2)),'sc');
4.0372 text(xd,0.29,num_str(Pz(3)),'sc');
4.0373
4.0374 text(0.0,0.25,'S-Dom','sc');
4.0375 text(xe,0.25,'Zeros','sc');
4.0376 text(xe,0.215,num_str(Zs(1)),'sc');
4.0377 text(xe,0.19,num_str(Zs(2)),'sc');
4.0378 text(xe,0.165,num_str(Zs(3)),'sc');
4.0379
4.0380 text(xd,0.25,'Poles','sc');
4.0381 text(xd,0.215,num_str(Ps(1)),'sc');
4.0382 text(xd,0.19,num_str(Ps(2)),'sc');
4.0383 text(xd,0.165,num_str(Ps(3)),'sc');
4.0384
4.0385 text(0.0,0.125,'MSE','sc');
4.0386 text(xe,0.125,num_str(MSE),'sc');
4.0387
4.0388 text(0.0,0.085,'LD','sc');
4.0389 text(xe,0.085,str,'sc');
4.0390
4.0391 %*****
4.0392 function [k,a,error,alpha] = durbin(r,nk)
4.0393 %DURBIN   Durbin's method for time-domain IIR filter design.
4.0394 % [k,a,error,alpha] = durbin(r,nk).
4.0395 % Finds a filter with denominator order nk, (zeros nk at
4.0396 % origin of z-plane), given the autocorrelation row vector r.
4.0397 % Size(r) = nk+1. The LPC filter coefficients are returned
4.0398 % in row vector a (length nk+1), ordered in descending
4.0399 % powers of Z, with reflection coefficients in row vector k
4.0400 % (length nk). See references for error and alpha.
4.0401 %
4.0402 %           1
4.0403 % H(z)=-----
4.0404 %           -1  -2  -3
4.0405 %     1 + a1*z + a2*z + a3*z ....

```

```

4.0406 %
4.0407 % L. Smith 22-2-92      (c) Copyright 1992.
4.0408 %
4.0409 % References:
4.0410 % [1] Papamichalis, Panos E., Practical Approaches to
4.0411 % Speech Coding, Englewood Cliffs, NJ: Prentice - Hall, 1987.
4.0412 % [2] Rabiner, L.R. and R.W. Schafer, Digital Processing
4.0413 % of Speech Signals, Englewood Cliffs, NJ: Prentice - Hall, 1978.
4.0414 % [3] Makhoul, John, "Linear Prediction: A Tutorial Review,
4.0415 % " Proceedings of the IEEE, Volume 63 (April,1975): pp. 561 - 580.
4.0416
4.0417 % DURBIN
4.0418     a(1) = 1.0;
4.0419     anew(1) = 1.0;
4.0420     k(1) = -r(2)/r(1);
4.0421     alpha = r(1)*(1.0-(k(1)^2));
4.0422     a(2) = k(1);
4.0423     for i = 2:nk
4.0424         error = 0.0;
4.0425         for j = 1:i
4.0426             error = error+(a(j)*r(i-j+2));
4.0427         end
4.0428         k(i) = -(error/alpha);
4.0429         alpha = alpha*(1.0-(k(i)^2));
4.0430         a(i+1) = k(i);
4.0431         for j = 2:i
4.0432             anew(j) = a(j)+(k(i)*a(i-j+2));
4.0433         end
4.0434         for j = 2:i
4.0435             a(j) = anew(j);
4.0436         end
4.0437     end
4.0438
4.0439 % END DURBIN
4.0440
4.0441 %*****
4.0442 % Mesh plot of forward and reverse frequency points.
4.0443 % f3d(F,R,n)
4.0444 % n = number of points (200 = full 3sec)
4.0445
4.0446 function f3d(F,R,n)
4.0447 clg;
4.0448 s = [5 5 5];
4.0449 m = [-37.5 80];
4.0450 RF = [F(64:-1:1,1:n);R(:,1:n)];
4.0451 mesh(RF,m,s),title('FORWARD AND REVERSE FREQUENCIES');
4.0452
4.0453 %*****
4.0454 % Converts Laplacian variables to discrete equivalents
4.0455 % to aid in the discrete curve fitting method employed.
4.0456 % var = lap2dis(bseg)
4.0457 % var = a1,a2,a3,b1,b2 (SEE DIFFERENCE EQU)
4.0458
4.0459 function var = lap2dis(s)
4.0460 clg;fs = 66;i = sqrt(-1);
4.0461 u = [1 zeros(1,max(size(s)-1))]; % system input
4.0462 res = lapfit(s);
4.0463 clg;

```

```

4.0464 LTD = res(1); LTE = res(2); LTR = res(3);
4.0465 Ps(1) = -LTR;
4.0466 Ps(2) = -LTE*LTD - i*LTE*sin(acos(LTD));
4.0467 Ps(3) = real(Ps(2)) - i*imag(Ps(2));
4.0468 Pz = exp(Ps/fs);
4.0469 Zz = [0 0 0];
4.0470 [num,den] = zp2tf(Zz',Pz',1);
4.0471 T = filter(num,den,u);
4.0472 gain = max(s)/max(T);
4.0473 num = num*gain;
4.0474 var = [den(2:4) num(1) num(2)];
4.0475 var(:)';
4.0476
4.0477 %T = filter(num,den,u);
4.0478 %subplot(221),title('Impulse response'),plot([s' T']),grid;
4.0479 %subplot(224),title('S-Domain'),s_zpplot([[0 0 0]' Ps'],'c5'),grid;
4.0480 %subplot(222),title('Z-Domain'),z_zpplot([[0 0 0]' Pz'],'c5'),grid;
4.0481
4.0482 %*****
4.0483 % Curve fitting procedure to calculate model parameters of velocity
4.0484 % segment. Laplacian fitting done in the frequency domain.
4.0485 % result = lapfit(segment to analyze);
4.0486 % result = [LTD LTE LTR]';
4.0487 function [result] = lapfit(bseg)
4.0488
4.0489 % points = no of FFT points, fft2 is the two dim. (real+imag) fft
4.0490 echo off
4.0491 clg;
4.0492 points = 512;
4.0493 fs = 66;
4.0494 Bpoints = max(size(bseg));
4.0495 T = (1:Bpoints)/fs;
4.0496 A = 0;B = 0;G = 0;LTD = 0;
4.0497
4.0498 FREQ = fft2(bseg,1,points)/fs;
4.0499 MAG = (sqrt(real(FREQ).*real(FREQ)+imag(FREQ).*imag(FREQ)));
4.0500 MAG = MAG/MAG(1);
4.0501
4.0502 % Note that the the index of the FREQ and MAG arrays must be multiplied by;
4.0503 % 2*pi * (index=fft bin no)*fs/points to get radian freq;
4.0504 % No is the maximum fft bin no of interest, this corresponds to
4.0505 % a frequency of 2*pi*No*fs/points;
4.0506 No = 50;
4.0507 w1 = 0:1:No-1;
4.0508 % Note that the actual frequency (radians) is 2*pi*(fft bin no)*fs/points;
4.0509 w = 2*pi*w1*fs/points;
4.0510 UNIT = ones(1:No);
4.0511
4.0512 % Perform weighted least squares fit in the frequency domain
4.0513 % No of iterations
4.0514 for ITERATION = 0:5
4.0515     if ITERATION == 0
4.0516         Ek = ones(1:No);
4.0517     else
4.0518         for j = 1:No
4.0519             i = 2*pi*j*fs/points;
4.0520             W3 = [i^3 i^2 i 1];
4.0521             Ek(j) = 1/(P*W3)^2;

```

```

4.0522         end;
4.0523     end;
4.0524
4.0525     L0 = UNIT*Ek';
4.0526     L2 = (Ek.*w).*w';
4.0527
4.0528     T1 = (w.*Ek)*(imag(FREQ(1:No)));
4.0529     T3 = (w.*w.*w.*Ek)*(imag(FREQ(1:No)));
4.0530
4.0531     S0 = (UNIT.*Ek)*(real(FREQ(1:No)));
4.0532     S2 = (w.*w.*Ek)*(real(FREQ(1:No)));
4.0533     S4 = (w.*w.*w.*w.*Ek)*(real(FREQ(1:No)));
4.0534
4.0535     U2 = (w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0536     U4 = (w.*w.*w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0537     U6 = (w.*w.*w.*w.*w.*w.*Ek)*(real(FREQ(1:No)).*real(FREQ(1:No))+imag(FREQ(1:No)).*imag(FREQ(1:No)));
4.0538
4.0539     M = [L0 0 T1 S2 -T3; 0 L2 -S2 T3 S4; T1 -S2 U2 0 -U4; S2 T3 0 U4 0; T3 -S4 U4 0 -U6];
4.0540     C = [S0 T1 0 U2 0]';
4.0541     SOL = inv(M)*C;
4.0542
4.0543     P = [SOL(5) SOL(4) SOL(3) 1];
4.0544     poles = roots(P);
4.0545     Z = [SOL(2) SOL(1)];
4.0546     zeros = roots(Z);
4.0547
4.0548     Error = 0;
4.0549     for j = 0:No-1
4.0550         i = j*2*pi*fs/points;
4.0551         W1 = [i 1];
4.0552         W3 = [i^3 i^2 i 1];
4.0553         Error = Error+(MAG(j+1)-(Z*W1)/(P*W3))^2;
4.0554     end;
4.0555     Error = sqrt(Error)/No;
4.0556
4.0557     Best = imag(poles(1));
4.0558     if Best == 0,
4.0559         Gest = -poles(1);
4.0560         Aest = -real(poles(2));
4.0561         Best = imag(poles(2));
4.0562     else
4.0563         Aest = -real(poles(1));
4.0564         Gest = -poles(3);
4.0565     end;
4.0566
4.0567 end;
4.0568
4.0569 H = (Gest*Aest*Aest+Gest*Best*Best)/((Gest-Aest)*(Gest-Aest)+Best*Best);
4.0570 FLOWest = H.*(exp(-Gest.*T) + exp(-Aest.*T)).*(((Gest-Aest)/Best).*sin(Best.*T)-cos(Best.*T));
4.0571 FREQest = fft2(FLOWest,1,points)/fs;
4.0572 MAGest = (sqrt(real(FREQest).*real(FREQest)+imag(FREQest).*imag(FREQest)));
4.0573
4.0574 OUT = [Aest;Best;Gest];
4.0575 LTDest = cos(atan(Best/Aest));
4.0576 LTEest = sqrt(Aest^2+Best^2);
4.0577 LTRest = Gest;
4.0578 OUT2 = [LTDest;LTEest;LTRest];
4.0579 result = OUT2;

```

```

4.0580
4.0581 %echo on
4.0582 % The waveform parameter estimates are:
4.0583 % A B G (Real,imaginary for complex poles, and real pole position resp.)
4.0584 %OUT
4.0585 % LTD,LTE,LTR are
4.0586 %OUT2
4.0587 %Error
4.0588 %echo off
4.0589
4.0590 %*****
4.0591 % Separate "C" file into Xmem and Ymem for MATLAB (386 MATLAB NEEDED)
4.0592 % [Xmem,Ymem] = lrd(name)
4.0593
4.0594 function [Xmem,Ymem] = lrd(name)
4.0595 Xmem = name(1:200,:);
4.0596 Ymem = name(201:400,:);
4.0597 clear name;
4.0598
4.0599 %*****
4.0600 % Separate "C" file into Ffreq and Rfreq for MATLAB (386 MATLAB NEEDED)
4.0601 % [FFreq,RFreq] = lsd(name)
4.0602
4.0603 function [FFreq,RFreq] = lsd(name)
4.0604 FFreq = name(1:200,:);
4.0605 RFreq = name(201:400,64:-1:1);
4.0606 clear name;
4.0607
4.0608 %*****
4.0609 % Prints complex number to string.
4.0610 % str = num_str(num)
4.0611
4.0612 function str = num_str(num)
4.0613
4.0614 str = num2str(real(num));
4.0615 if(imag(num) ~= 0)
4.0616     str_i = num2str(abs(imag(num)));
4.0617     if(sign(imag(num)) == 1)
4.0618         str = [str '+' str_i];
4.0619     else
4.0620         str = [str '-' str_i];
4.0621     end
4.0622 end
4.0623
4.0624 %*****
4.0625 % To determine transfer function of an IIR filter given the impulse
4.0626 % response of the system. Prony's curve fitting is done.
4.0627 % pro(Bseg,nb,na)
4.0628 % Bseg = filtered velocity segment to analyze
4.0629 % nb = numerator order
4.0630 % na = denominator order
4.0631 % NB!! Bseg(1) changed due to prony DIV
4.0632
4.0633 function pro(s,nb,na)
4.0634 hold off;clg;fs = 66;
4.0635 s = s + (max(s)/1000);
4.0636 u = [1 zeros(1,max(size(s)-1))]; % system input
4.0637

```

```

4.0638 % PRONY
4.0639 % NUMBER OF POLES AND ZEROS AS GIVEN (NB! na>nb)
4.0640 xi = 1:(max(size(s)));
4.0641 [b,a] = prony(s,nb,na);
4.0642 [Zz,Pz,k] = tf2zp(b,a);
4.0643 Ps = fs.*log(Pz);
4.0644 Zs = fs.*log(Zz);
4.0645 T = filter(b,a,u);
4.0646
4.0647 if na == 3
4.0648     % PRONY
4.0649     Ps = Ps(:);           % ensure column vector
4.0650     if imag(Ps(1)) == -imag(Ps(2))
4.0651         sys_2 = poly(Ps(1:2));
4.0652         LTR = -real(Ps(3));
4.0653     else
4.0654         if imag(Ps(2)) == -imag(Ps(3))
4.0655             sys_2 = poly(Ps(2:3));
4.0656             LTR = -real(Ps(1));
4.0657         else
4.0658             return;           % ERROR
4.0659         end
4.0660     end
4.0661     LTE = sqrt(sys_2(3));
4.0662     LTD = sys_2(2)/(2*LTE);
4.0663     PI = (max(T)-min(T))/mean(T);
4.0664 end
4.0665
4.0666 subplot(221),title('Impulse Response'),plot([s' T']),grid;
4.0667 subplot(222),title('Z-Domain'),z_zpplot([0.*Zz Pz],'c5'),grid;
4.0668 subplot(224),title('S-Domain'),s_zpplot([0.*Zs Ps],'c5'),grid;
4.0669
4.0670 if na==3
4.0671     xe=0.1;xr=0.2;xd=0.3;xp=0.4;
4.0672     text(xe,0.45,'LTE','sc');
4.0673     text(xr,0.45,'LTD','sc');
4.0674     text(xd,0.45,'LTR','sc');
4.0675     text(xp,0.45,['PI ' '( num2str((max(s)-min(s))/mean(s)) )'],'sc');
4.0676
4.0677     text(0.0,0.415,'Prony','sc');
4.0678     text(xe,0.415,num2str(LTE),'sc');
4.0679     text(xr,0.415,num2str(LTD),'sc');
4.0680     text(xd,0.415,num2str(LTR),'sc');
4.0681     text(xp,0.415,num2str(PI),'sc');
4.0682
4.0683 end
4.0684
4.0685 %*****
4.0686 % Select waveform segment for analysis by disfit() and normalize
4.0687 % the selected segment.
4.0688 % S = ss3(B)
4.0689 % B = blood velocity
4.0690
4.0691 function [S] = ss3(B)
4.0692 clg;
4.0693 B = B(:);
4.0694 subplot(211),plot(B,'r'),title('BLOOD VELOCITY WAVEFORM'),grid,hold on
4.0695 [x,y] = ginput(1);

```

```

4.0696 xb = floor(x);
4.0697 subplot(211),plot(xb,B(xb),'xw');
4.0698 [x,y] = ginput(1);
4.0699 xe = floor(x);
4.0700 subplot(211),plot(xe,B(xe),'xw');
4.0701 x = sort([xb xe]);
4.0702 xb = x(1);xe = x(2);
4.0703 Sp = abs(xe-xb);
4.0704 S = B(xb:xb+Sp);
4.0705 S = S-S(1);
4.0706 S(max(size(S))) = 0; % ss3 difference
4.0707 S = S./max(S);
4.0708 hold off
4.0709 subplot(212),title('SELECTED SEGMENT (NORMALIZED)'),plot(S,'g'),grid
4.0710
4.0711 %*****
4.0712 function s_zpplot(zepo,col)
4.0713 % SZPLOT Plots zeros and poles in S-Domain
4.0714
4.0715 [nzp,nc]=size(zepo);
4.0716 if fix(nc/2)~=nc/2
4.0717 disp('Error: ZEPO should have an even number of columns')
4.0718 return,end
4.0719
4.0720 w = 0;
4.0721 mm = max(abs([zepo(:)])); m=fix(mm+1); r=m*27/20;
4.0722 axis([-r r -m m]);
4.0723 plot(w)
4.0724 hold
4.0725 for k = 1:2:nc
4.0726 iz = find(abs(zepo(:,k))>0);
4.0727 if length(iz) > 0
4.0728 plot(real(zepo(iz,k)),imag(zepo(iz,k)),'o' col)
4.0729 end
4.0730 ip = find(abs(zepo(:,k+1))>0);
4.0731 if length(ip) > 0
4.0732 plot(real(zepo(ip,k+1)),imag(zepo(ip,k+1)),'x' col)
4.0733 end
4.0734 end
4.0735
4.0736 hold
4.0737 axis('normal')
4.0738
4.0739 %*****
4.0740 % Lowpass filter the selected velocity segent using a Yulewalk IIR
4.0741 % filter. NOTE : FiltFilt is used to minimise phase distortion.
4.0742 % v = vfil(s,fcut,N,gr)
4.0743 % s = segment to filter
4.0744 % fcut = cutoff frequency
4.0745 % N = filter order
4.0746 % gr = 0 no graph
4.0747 % = 1 display graph
4.0748
4.0749
4.0750 function [v] = vfil(s,fcut,N,gr)
4.0751 fs = 66;frdiv = 10;n = 256;
4.0752 fc = round(fcut/(fs/(2*frdiv))); %num points where gain will be unity
4.0753 f = [0:1/frdiv:1];

```

```

4.0754 H = [ones(1,fc) zeros(1,frdiv-fc+1)];
4.0755 fhz = f*fs/2;
4.0756 [Bh,Ah] = yulewalk(N,f,H);
4.0757 hh = freqz(Bh,Ah,n);    % compute complex frequency response
4.0758 hy = abs(hh);         % compute magnitude
4.0759 v = filtfilt(Bh,Ah,s);
4.0760
4.0761 ff = fs/(2*n)*(0:n-1);
4.0762 xi = 1:max(size(s));
4.0763 if (gr == 1)
4.0764     clg;
4.0765     subplot(211),plot(xi,s,'r','xi,v','w'),title('Filter Output'),...
4.0766     ylabel('Frequency (Hz)');
4.0767     subplot(212),plot(fhz,H,'r','ff,hy','w'),title('Frequency Response'), ..
4.0768     xlabel('Frequency (Hz)'),ylabel('Magnitude');
4.0769 end
4.0770
4.0771 %*****
4.0772 function z_zpplot(zepo,col)
4.0773 %ZPLOT Plots zeros and poles in Z-Domain.
4.0774 %
4.0775 % zpplot(ZEPO) zpplot([Z,P]);
4.0776 %
4.0777 % The zeros and poles, specified by ZEPO (See ZP for the format) are
4.0778 % plotted, with 'o' denoting zeros and 'x' poles. Poles and zeros
4.0779 % associated with each double column of ZEPO are plotted sequentially.
4.0780
4.0781 [nzp,nc] = size(zepo);
4.0782 if fix(nc/2) ~= nc/2
4.0783     disp('Error: ZEPO should have an even number of columns')
4.0784 return,end
4.0785
4.0786 %
4.0787 om = 2.1*pi*[1:100]/100;
4.0788 w = exp(om*sqrt(-1));
4.0789 mm = max(abs([zepo(:)])); m = fix(mm+1); r = m*27/20;
4.0790 axis([-r r -m m]);
4.0791 plot(w,'w')
4.0792 hold
4.0793 for k = 1:2:nc
4.0794     iz = find(abs(zepo(:,k))>0);
4.0795     if length(iz) > 0
4.0796         plot(real(zepo(iz,k)),imag(zepo(iz,k)),['o' col])
4.0797     end
4.0798     ip = find(abs(zepo(:,k+1))>0);
4.0799     if length(ip) > 0
4.0800         plot(real(zepo(ip,k+1)),imag(zepo(ip,k+1)),['x' col])
4.0801     end
4.0802 end
4.0803
4.0804 hold
4.0805 axis('normal')
4.0806
4.0807 %*****

```

REFERENCES

Alessio, T.D., Di Giuliomaria, C., Sacco, R. and Cavallaro, A.(1983) On the Use of autoregressive time series in the modelling of Doppler signals for diagnostic purposes. *Clin. Phys. Physiol. Meas.*, Vol. 4, No. 4, 395-401.

Atkinson, P., Woodcock, J.P., (1982) Doppler ultrasound and its use in clinical measurement. (Medical Physics Series. Academic Press). London.

Brown, J.M., Nahorski, Z.T., Woodcock, J.P. and Morris, S.J. (1978) Transfer-function modelling of arteries. *Med. & Biol. Eng. & Comput.*, Vol. 16, 161-164.

Bunn, A.E. and Guelke, R.W.(1983) Mechanics of the cochlea model. *Acustica*, 53, Vol. 5, 238-249.

Campbell, W.B., Skidmore, R. and Baird, R.N.(1984) Variability and reproducibility of arterial doppler waveforms. *Ultrasound in Med. & Biol.*, Vol. 10, No. 5, 601-606.

Capper, W.L., Amoores, J.N., Clifford, P.C., Immelman, E.J., and Harries-Jones, E.P.(1986) A system for the rapid analysis of the femoral waveform at the bedside. *Ultrasound in Med. & Biol.*, Vol.12, No. 1, 31-38.

Capper, W.L., Bunn, A.E. and Guelke, R.W.(1991) The estimation of tube wall compliance using acoustic input impedance. *IEEE Trans. on Biomed. Eng.* 38(6), 544-550.

Capper, W.L., Smith, L., Immelmann, E.J.(1994). A system for the acquisition and processing of arterial blood flow velocity waveforms. (In press).

Evans, D.H. and Schlindwein, F.S.(1989) A real-time autoregressive spectrum analyser for Doppler ultrasound signals. *Ultrasound in Med. & Biol.*, Vol. 15, No. 3, 263-272.

Evans, D.H. and Schlindwein, F.S.(1990) Selection of the order of autoregressive models for spectral analysis of Doppler ultrasound signals. *Ultrasound in Med. & Biol.*, Vol. 16, No. 1, 81-91.

Fitzgerald, D.E., Gosling, R.G., Woodcock, J.P.(1971) Grading dynamic capability of arterial collateral circulation. *Lancet*. 1, 66.

Garbini, J.L., Forster, F.K. and Jorgensen, J.E.(1982) Measurement of fluid turbulence based on pulsed ultrasound technique. Part I, II. *J Fluid. Mech.*, 118, 445-470, 471-505.

Giordano, A.A., Frank, M.H.(1985) Least square estimation with application to digital signal processing. Wiley, New York, 40-44, 250-254.

Gosling, R.G. and King, D.H.(1974) Continuous wave ultrasound as an alternative and complement to X-rays in vascular examination. In *CARDIOVASCULAR APPLICATIONS OF ULTRASOUND* (Edited by R.E. Reneman), Chap. 22. North-Holland, Amsterdam.

Guelke, R.W. and Bunn, A.E.(1984) Resonance in theories of hearing. *The J. of Laryngology and Otology*. 98, 1177-1183.

Johnston, K.W., Kassam, M., and Cobbold, R.S.C.(1983) Relationship between doppler pulsatility index and direct femoral pressure measurement in the diagnosis of aorto-iliac occlusive disease. *Ultrasound in Med. & Biol.*, 9, 271-281.

Johnston, K.W., Kassam, M., Koers, J., Cobbold, R.S.C. and MacHattie, D.(1984) Comparative study of four methods for quantifying doppler ultrasound waveforms from the femoral artery. *Ultrasound in Med. & Biol.*, 10, 1-12.

Kaluzynski, K.(1989) Selection of a spectral estimation method for the assessment of velocity distribution based on the spectral distribution of ultrasonic signals. *Med. & Biol. Eng. & Comput.*, 27, 463-469.

Kaluzynski, K.(1989) Order selection in Doppler blood flow signal spectral analysis using autoregressive modelling. *Med. & Biol. Eng. & Comput.*, 27, 89-92.

Law, Y.F., Graham, J.C., Cotton, L.T. and Roberts, V.C. (1984) Validity of the transfer function model of the human arterial system of the lower limb in man. *Med. & Biol. Eng. & Comput.*, 22, 537-542.

Meara, L.A. (1984) Pole-zero extraction by non-linear regression of discrete-time arterial blood-flow waveforms. *Med. & Biol. Eng. & Comput.*, 22, 281-284.

Mendenhall, M., Scheaffer, R.L., (1986) p.621-p.627. *Mathematical Statistics with Applications*. Third Edition (Duxbury Press - Boston).

Skidmore, R. (1979) The use of the transcutaneous ultrasonic flow meter in the dynamic analysis of blood flow. Ph.D. Thesis, University of Bristol.

Skidmore, R. and Woodcock, J.P. (1980) Physiological interpretation of Doppler-shift waveforms-I. *Ultrasound in Med. & Biol.*, 6, 7-10.

Skidmore, R, Baker, J.D. and Cole, S.E.A.(1989) Laplace transform analysis of femoral artery Doppler signals : The state of the art. *Ultrasound in Med. & Biol.*, Vol. 15, No. 1, 13-20.

Wijn, P.F.F., van der Sar, P., Gootzen, T.H.J.M., Tilmans, M.H.J., Skotnicki, S.H. (1987) Value of the spectral broadening index in continuous wave Doppler measurements. *Med. & Biol. Eng. & Comput.*, 22, 281-284.